RA.1217 MAINTENANCE MANUAL

PREFACE

Handbook Guide

This Volume contains detailed information for maintenance purposes. It is assumed that the user also has a copy of the RA.1217 Operators Manual, to which reference should be made for general information concerning installation, operation, and technical principles.

Variants

Chapters 1 to 8 of this Volume refer to the basic RA. 1217 receiver. Details of any variant models will be given in appendices, which will be inserted immediately prior to the illustrations at the rear of the handbook.

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<u>RA. 1217</u>

LIST OF ILLUSTRATIONS

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H.F. Communications Receiver RA. 1217

Fig. No.

At Rear of Book

1 2 3	Wadley System - Block Diagram Electronic Band Selection: Explanatory Block Diagram Block Diagram: RA.1217
4	Circuit: R.F. Module
5	Circuit: 1st V.F.O.
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Frequency Range:	l to 30 MHz
Modes of Reception:	D.S.B. M.C.W., C.W., S.S.B. (U.S.B. or L.S.B.)
Tuning:	Digital presentation in units of kilohertz/ second with interpolation calibration at 200 Hz intervals.
Resetting Accuracy:	±200 Hz
Calibration Accuracy:	±500 Hz (when calibrated to nearest 100 kHz checkpoint)
Calibration:	A 100 kHz signal, derived from the 1 MHz standard crystal oscillator having an accuracy of 5 parts in 10^6 , provides check points at 100 kHz intervals.
Frequency Stability:	 (a) ±50 Hz over an eight hour period at constant ambient temperature and humidity after 2 hours from switching on.
	(b) Better than 50 Hz per degree Centigrade at constant humidity after 2 hours from switching on.
Antenna Input:	(a) Nominal impedance 75 ohms unbalanced
	(b) Wideband, or tuned in five automatically selected bands:
	 (i) 1 to 2 MHz (ii) 2 to 4 MHz (iii) 4 to 8 MHz (iv) 8 to 16 MHz (v) 16 to 30 MHz
Sensitivity:	In 3 kHz b andwidth:
	(a) C. W., S.S.B: l μV for 15 dB signal/ noise ratio.
	(b) M.C.W., D.S.B: 30% modulated at 400 Hz: 3 μV for 15 dB signal/noise ratio.

Selectivity:	Three i.f. bandwidths fitted as standard, although positions available for up to five filters.
	Standard Filters:3 dB Bandwidth6/60 dB Shape Factor8 kHz1:43 kHz1:4200 Hz1:10
	Additional Filters Available:13 kHz1:41.2 kHz1:4500 Hz1:10
Cross Modulation:	For a wanted signal of level up to 1 mV and with appropriate use of Aerial Attenuator (AE ATT) control, an interfering signal 20 kHz removed and modulated 30% must have a level greater than 45 dB above that of the wanted signal to produce a cross modulation of 3%. The ratio of wanted to unwanted signal level is improved at the rate of 2 dB for each one percent up to 10 percent off tune.
Intermodulation:	Better than 80 dB on 1 microvolt for inter- fering signals at least 10% removed from the wanted signal.
Blocking:	For levels of wanted signal up to 1 mV, and with appropriate use of the Aerial Attenuator (AE ATT control) an interfering signal 20 kHz removed must have a level 56 dB greater than the wanted signal to re- duce the signal by 3 dB. The ratio of wanted to unwanted signal level is improved at the rate of 2 dB per one percent up to 10% off tune.
Spurious Response to External Signals (Image, etc):	(a) External signal less than 10% off tune shall be greater than plus 60 dB rela- tive to 1 microvolt to produce a spur- ious signal equivalent to 1 microvolt.
	(b) With tuned aerial (antenna) external signals more than 10% off tune shall be greater than plus 80 dB relative to 1 microvolt to produce a spurious signal equivalent to 1 microvolt.

Automatic Gain Control:	(a)	Time constants (nominal):
	Shor Medi Long		e Discharge 20 mS 250 mS 4 S
	(b)	Output change:	
	will	crease in input of 85dB a produce a change in outpu 4 dB.	
B. F. O:	(a)	Variable: ± 3 kHz centre -3 and -6kHz with respe frequency.	
	(b)	Fixed: ± 1.5 kHz (USB/L controlled.	SB) crystal
A.F. Output:	(a)	Headphone jack on front nominal in 600Ω .	panel: 10 mW
	(b)	10 m W in 600Ω at rear to alternative version prov into 15 ohms for an exten is available to special o	iding one watt rnal loudspeaker
	(c)	l m W in 600Ω 'line' outle level is independent of t control setting.	
Overall A.F. Distortion:	Less	than 5%.	
Overall A.F. Response:	au di 6000	a selectivity handpass of frequency output level f Hz. will not very more f response.	rom 100 Hz to
Meter Indication:	(a)	R.F. Levels.	
	(Ъ)	A.F. level to line.	

Controls:

- (a) Meter Switch (AF level/R.F. level)
- (b) Frequency Readout In-line Digital
- (c) 'Megahertz' Tuning
- (d) 'Kilohertz, Tuning
- (e) System Switch
- (f) DET/B.F.O. Mode
- (g) B.F.O. Tuning
- (h) R.F./I.F. Gain
- (i) A.F. Gain
- (j) A.F. Line Level
- (k) R.F. Tune
- (1) R.F. Bandswitch
- (m) Tuning Locks
- (n) Dimmer (scale illumination)
- (o) Calibrate Fine Tune
- (p) Bandwidth Switch
- (q) R.F. Attenuator
- (r) 2nd V.F.O. INT/EXT switch
- (a) Antenna Input
- (b) I.F. Outputs
- (c) A.F. Line Output
- (d) External L.S. Output
- (e) 2nd V.F.O. Output
- (f) 2nd F.V.O. Input
- (g) 1 MHz Output
- (h) 1 MHz Input
- (i) 1.7 MHz input/output
- (j) L.F. Adaptor Input
- (k) Panoramic Adaptor Output
- (1) A.C. Supply input

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External Connections:

	(m)6 volts output
	(n) Diversity A.G.C.
	(o) Muting Contact
	(p) L.F. Adaptor h.t.
	(q) L.F. Adaptor a.g.c.
Power Supplies:	Alternative power units are available as follows:
	(a) 100-125 or 200-250V, 45-400 Hz, a.c. single phase; or 21-27V d.c.positive earth.
	(b) 9-15V or 18-30V d.c. positive or n egative earth.
Power Consumption:	40VA approx. (with one-watt audio amplifier)
Dimensions:	High Wide Deep
	3.5 in. x 19 in. x $16\frac{1}{2}$ in.
	(8.9 cm) (48.25 cm) (41.9 cm)
Weight:	
Weight: Environmental Conditions:	(8.9 cm) (48.25 cm) (41.9 cm)
C C	 (8.9 cm) (48.25 cm) (41.9 cm) 31 lb. (14.2 kg) approximately. The equipment is, designed to meet certain of the requirements of specification DEF 133 L2, operating within the ambient temperature

CHAPTER 1

GENERAL DESCRIPTION

INTRODUCTION

 The RA.1217 is a versatile high-stability h.f. communications receiver designed for use in a 19-inch (48.25 c.m.) rack mounting under all climatic conditions. Solid state techniques are used throughout. The frequency range is 1 to 30 MHz. Operation down to 200 kHz (but with slightly degraded performance) can be obtained by linking terminals on the rear panel. A comprehensive range of ancillary units can be connected to the receiver.

TECHNICAL INFORMATION

2. This handbook contains installation and operating information with a brief technical description. The circuits at the rear of the book are for information only. For detailed technical descriptions, alignment instructions and component details, the user must refer to the RA. 1217 Maintenance Manual.

RA. 1217 VARIANTS

3. Specification variation is normally confined to a choice of i.f. filter bandwidths and optional audio outputs of either 10 m W 600Ω or one-watt 15Ω. The first production receivers are fitted with the one-watt amplifier but the standard receiver is now the 10 m W type. It should be noted that the one-watt version cannot provide d.c. power for ancillary units such as an i.s.b. adaptor, f.s.k. converter etc. Refer to the Preface which precedes this page for guidance in the identification of the different types of audio amplifier.

ANCILLARY EQUIPMENT

4. The RA.1217 is provided with a number of outlets which permit the attachment of additional units such as :-



RA.1217 TRANSISTORISED H.F. COMMUNICATIONS RECEIVER

- (a) A frequency synthesizer or pre-set channel oscillator which can provide very precise and stable frequency setting of the receiver.
- (b) Panoramic adaptor to provide visual display of a portion of the frequency spectrum.
- (c) L.F. Adaptor which extends the lower frequency limit of the receiver to 3 kHz.
- (d) Diversity switching unit for linking two RA.1217 receivers in dual diversity reception.
- (e) Line amplifier, signal frequency measuring equipment, i.s.b. adaptor etc.

GENERAL INFORMATION

Tuning

5. The receiver is tuned by two principal controls, one control is set to indicate the 'megahertz' content of the signal frequency, the other is set to indicate the 'kilohertz' content. The frequency indication is by a mechanical in-line readout system providing a digital display. Scale markings enable the frequency to be set to an accuracy of 0.2 kHz.

6. The RA.1217 can be tuned over its entire frequency range with no mechanical band-switching. The r.f. input can be set to 'wideband' or tuned circuits can be switched into the r.f. input stage to obtain improved rejection of adjacent channels. An r.f. attenuator control permits selected attenuation of the input signals to minimize the effects of severe interference or excessive signal strength.

7. The Megahertz (MHZ) and Kilohertz (KHZ) tuning controls can be mechanically locked to their selected positions. A calibrate and fine tune control provides a range of tuning adjustment of 8 kHz, this is useful as a clarifier in single sideband reception or as a fine adjustment when the main controls are locked. Calibration signals at 100 kHz intervals, derived from the internal reference crystal oscillator are available for aligning the kilohertz tuning readout.

Audio Outputs

8. The standard receiver provides the following outputs:

- (a) 10 m W 600 Ω at the Phones socket on the front panel.
- (b) $1 \text{ m W } 600\Omega$ (balanced) at the rear outlet.
- (c) 10 m W 600 Ω (unbalanced) at the rear outlet.

Earlier receivers are fitted with a one-watt audio amplifier. Refer to the Preface at the beginning of this chapter.

Construction

9. The receiver is of modular construction, each module being a separate unit connected by plugs and sockets. The main chassis is of cast aluminium to which the individual modules are attached. A range of alternative sub-assemblies is available, providing for the choice of either 100 kHz or 455 kHz output, and for the provision of non-standard bandwidths to meet special requirements.

Individual Modules and Assemblies

10.	(a)	Main Chassis	(g)	37.5 MHz B.P.Filter
	(Ъ)	R.F. Unit	(h)	3rd Mixer
	(c)	lst V. F. O.	(i)	2nd V.F.O.
	(đ)	lst Mixer and 40 MHz Filter	(j)	First I.F. Amplifier Unit
	(e)	2nd Mixer	(k)	Main I.F. Module
	(f)	37.5 MHz Generator 1 MHz Osc. and calibrator	(1)	B. F. O. and Bandwidth Assembly
			(m)	Power Unit
			(n)	Audio Amplifier Board

Power Supply

11. The standard power unit is the PU. 1153 which can operate from the following power sources:

- (a) A.C. 100 125 or 200 250V 45 400 Hz Single phase
- (b) D.C. 21 27 V, positive earth.

An alternative power unit is available, to operate from d.c. only in the supply

range 9 to 30 V with either positive or negative earth. The 10 mW receiver power consumption is 10 VA approximately. The attachment of ancillary units will increase the power consumption by a further 10 VA approximately. If a One-watt audio amplifier board is fitted, the - 16V h.t. load is increased by approximately 200 milliamps.

Optional I.F. Outputs

12. Two types of i.f. unit are available. One provides an i.f. output of 100 kHz, the other of 455 kHz. The customer can check that the correct type is fitted to the receiver by noting the engraving on the i.f. outlet panel which is located at the lower left hand corner when viewing the rear of the receiver. The lower right hand socket is engraved with i.f. output frequency, either 100 kHz OUT, or 455 kHz OUT. Unless otherwise requested, the 100 kHz i.f. unit will be supplied.

HANDBOOK NOTES

Circuit References

13. Receiver construction is based upon a number of sub-assemblies. Each sub-circuit is separately manufactured and has an individual sequence of component numbering. When studying a technical description or component list, note whether the circuit consists of more than one assembly and ensure that the correct portion of the circuit is referred to. In the R. F. unit it has been considered necessary to allocate a special prefix figure to each sub-circuit in the circuit diagram, but in the other modules it is considered that there should be no difficulty in correct identification of a component.

Control References

14. The controls are engraved on the front panel in lower case lettering, but in the handbook such markings will frequently be printed in capital letters (e.g. R.F. TUNE) to add clarity and emphasis to an instruction.

Modifications

15. Before studying this Manual refer to the Handbook Change Information sheet at the beginning of the book and note whether any amendments have been introduced since the book was printed.

CHAPTER 2

INSTALLATION

INTRODUCTION

1. The chapter contains information under three principal headings. "General Information" gives information on the various receiver outlets and connections; "Preparation for use" describes the essential checks prior to operating the receiver for the first time, and "Connections to Ancillary Units" gives instructions for connecting a Synthesizer, L.F. Adaptor and I.S.B. Adaptor. All connections are made at the rear of the receiver.

GENERAL INFORMATION

REMOVAL OF COVERS

Dust Covers

2.

To obtain access to the various modules and chassis components it is necessary to remove the following covers.

- (a) The top cover.
- (b) Left-hand side cover (viewed from the front) for access to the R.F. Module.
- (c) Right-hand side cover for access to the I.F. Module.
- (d) A small cover on the underside for acess to the Audio Amplifier Board.

All covers are secured by screws which are self evident.

Module Covers

3. CAUTION: The various module covers are retained by screws which must be tightened securely to maintain a high standard of screening, but when tightening do not apply excessive force which could cause the screw-hole threads to become damaged.

REAR PANEL CONNECTIONS

4. A brief description is given of each rear panel connection to assist the user in making connections for a particular system or to ancillary units such as i.s.b. adaptor etc. A rear view of the receiver is shown in Fig. 19.

R.F. Input (Antenna)

5. The antenna should be connected using a coaxial connector and 75Ω unbalanced transmission line to the socket marked R.F. Input.

12-way Socket SKT11

NOTE: The free plug and the accessory set for connection to external wiring are: Plug: Plessey MK4 2CZ83302/5 Accessory Set: 508/1/03013/205

6. Pin Function

- A) B) 1 mW audio output (balanced)
- C Screen to pin D.
- D 10 m W 600Ω audio output (unbalanced), alternatively l watt 15Ω. Refer to Preface at beginning of Chapter 1.
- E 16 volts d.c. power supply output for external use.
- F Unregulated d.c. output at 20 to 24 volts.
- G Antenna muting. An earth loop connected to this terminal operates the muting relay, to protect the receiver input when keying an associated high power transmitter.
- H Diversity a.g.c. connection from associated diversity receiver, or a.g.c. from i.s.b. adaptor.
- J Screen to pin H.
- K L.F. H.T. Provides 16V d.c. to L.F. Adaptor Unit.
- L L.F. A.G.C. Provides a.g.c. to L.F. Adaptor Unit.
- M Screen to pin L.

Terminal Block TB1

NOTE: Terminals 1 to 4 on TB1 provide a linking facility to permit receiver operation down to 200 kHz.

7.	Pin No.	Identification	Remarks
	1	H.T. R.F.	For operation down to 200
	2	H.T. L.F.	kHz link pin 1 to pin 2, and and pin 3 to pin 4. Remove
	3	A.G.C. R.F.	the links if an L.F. Adaptor
	4	A.G.C. L.F.	unit is in use. Pins 2 and 4 are in parallel with pins K and L respectively in the 12-way socket SKT11.
	5	DET	Provides a low-level audio output for driving an external amplifier.

Coaxial Outlets

8.

- NOTE 1 : Refer to Table 1 and 2 at the end of this chapter for details of the r.m.s. levels applicable to the following outlets.
 - NOTE 2: For connecting external wiring to the coaxial sockets SKT1 to 3 and SKT12 to 17 use 75 ohm plug Transradio BNC1/7. (British NATO number 5935-99-580-9636). For a 455 kHz I.F. output at SKT2 use a 50 ohm plug type UG/88U.

Identification	Impedance	Function
SKT1 1.6 MHz	$1 \ k\Omega$	1.6 MHz I.F. Output
SKT2 100 KHz OUT or	75Ω	100 kHz I.F. Output
SKT2 455 kHz	50Ω	455 kHz I.F. Output
SKT3 1.7 MHz IN/OUT (100 kHz i.f.)	75Ω	Accepts 1.7 MHz (or 1.5 MHz) input to 100 kHz i.f. converter Remove the 1.7 MHz crystal when the external 1.7 MHz is in use.
or SKT3 1.145 MHz IN/OUT (455 kHz i.f.)	50Ω	Accepts 1. 145 MHz input to 455 kHz i.f. converter, or provides output to slave receiver in diversity.

NOTE: Outlets SKT1 to SKT3 are part of the i.f. module assembly, (Fig. 12).

REAR PANEL CONNECTIONS (Continued)

Identification	Impedance	Function
SKT12 2nd V.F.O. IN	75Ω	Accepts 3.6 to 4.6 MHz from synthesizer
SKT13 2nd V.F.O. OUT	75 Ω	Provides 2nd VFO control of a slave receiver.
SKT14 L.F.	75Ω	Accepts the r.f. signal from anl.f. adaptor.
SKT15 PAN	7 5Ω	Connects the 2 to 3 MHz i.f. spectrum from the 2nd Mixer to a panoramic adaptor unit.
SKT16 1 MHz IN	75Ω	Accepts an external 1 MHz frequency standard. The internal 1 MHz crystal must be removed when this external source is in use.
SKT17 1 MHz OUT	75Ω	Provides a 1 MHz output from the internal crystal oscillator.

Earth (Ground) Terminal

9. A terminal is provided on the rear panel for connection to the earthing system of a cabinet.

Power Module Connections

10.	3-Way Plug P	L1
	Pin A	A.C. connection to line or D.C. +ve
	Pin B	A.C. neutral connection, or D.Cve
	Pin C	A.C. supply earth (ground) connection.
The free sock	et and the acce	ssory set for connecting the external power cable are :

Socket:	Plessey 2 CZ 83283/5
Accessory Set:	508/1/03008/205

PREPARATION FOR USE

GENERAL INSPECTION

11.

- (1) Remove the top cover from the receiver.
 - (2) Check that the receiver assembly is clean and free from damage and that all interconnecting leads and module covers are secure. Note particularly that all gear wheels are clean and free from packing materials and other extraneous matter.
 - (3) Move the lock controls (adjacent to each tuning knob) upward to the "free" position. Check that each tuning control rotates freely over the complete tuning range.
 - (4) Set the 2nd V.F.O. switch to INT.
 - (5) Check that the correct i.f. module is fitted by noting the engravings on the i.f. outlet panel at the rear of the receiver, which will indicate either a 100 kHz, or 455 kHz i.f. output.
 - (6) Check the settings of the A. C. /D. C. switch and the Voltage Selector switch on the power module. If necessary, re-set these switches as follows:

A.C./D.C. Switch

12. A locking plate displays the required type of power input either 'A. C.' or 'D. C.' If the indication is not correct remove the locking plate (one screw) set the switch in the alternative position and replace the locking plate in the reversed position so that the correct indication is displayed.

Voltage Selector Switch

 This switch is in use when the A.C./D.C. switch is set to A.C. It should display the correct a.c. supply voltage range, either
 100-125 or 200-250. If the setting is not correct, remove the locking plate (two screws) re-set the switch and replace the locking plate in the reversed position, so that the correct voltage range is indicated.

Fuses

14. Check that the fuses at the rear of the power module are correct in accordance with the rating engraved on the rear panel. The replacement fuse links must be of the anti-surge type. Beswick Type TDC134 is recommended.

GENERAL INSPECTION (Continued) Power Supply Connection

15. A free connector assembly is normally supplied with the receiver for connection to the 3-way plug on the power module. If not supplied, refer to paragraph 10 for details. Connections should be as follows.

16.	A.C. Supply		
	Pin A		Line
	Pin B		Neutral
	Pin C		Earth (ground)
		or	
17.	D.C. Supply		
	Pin A		+ve
	Pin B		-ve

Phones

 Headphones (600Ω) may be plugged into front panel jack socket. The phones jack plug is an Igranic P50.

Operating Down to 200 kHz

- 19. (1) On the rear panel link the terminal H.T R.F to the terminal H.T L.F.
 - (2) Link the terminal A. G. C. R. F. to the terminal A. G. C. L. F.

Other Connections

20. Other connections will be according to the users requirements. Refer to paragraphs 6 to 8 under General Information.

CONNECTIONS TO ANCILLARY UNITS

SYNTHESIZER CONNECTIONS

General

21. The synthesizer, or channel oscillator, should provide the following inputs to the receiver. Inputs (a) and (b) are essential. Input (c) is required only if the external i.f. (100 kHz or 455 kHz) is to be used.

- (a) 3.6 to 4.6 MHz (variable). Level to be 100 mV e.m.f. from 75 Ω .
- (b) 1 MHz (fixed). Level to be 100 mV e.m.f. from 75Ω .

CONNECTIONS TO ANCILLARY UNITS (Continued)

- (c) An additional fixed frequency at 1.7 MHz is required for use in receivers equipped for 100 kHz i.f. output, or at 1.145 MHz for receivers with 455 kHz i.f. output. The voltage of the applied frequency must be not less than 50 mV in 75 ohms.
- NOTE: The output levels from the Synthesizer Type MA. 350B are 2 volts e.m.f. therefore attenuating connectors must be used with this instrument.

Connecting the Synthesizer

22.

- (1) Set the 2nd V.F.O. switch on the front panel to EXT
 - (2) Remove the receiver cover
 - (3) Remove the cover from the 37.5 MHz Generator Module
 - (4) Remove the 1 MHz crystal from its holder and store it in a safe place. Replace the module cover.
 - NOTE: Operations (5) and (6) are required only if the external i.f. (100 kHz or 455 kHz) is to be used.
 - (5) Remove the cover from the i.f. module.
 - (6) Remove the 1.7 MHz (or 1.145 MHz) crystal, and store in a safe place.
 - (7) Replace the cover on the i.f. module.
 - (8) Replace the receiver cover and make the following connections to the rear panel.
 - (9) Connect the 3.6 to 4.6 MHz output from the synthesizer to the socket SKT13 '2nd V.F.O. IN'. Refer to paragraph 21 (a) regarding drive level.
 - (10) Connect the synthesizer 1 MHz output to socket SKT16 '1 MHz IN'. Refer to paragraph 21 (b) regarding drive level.
 - (11) If the receiver is equipped for an i.f. output of 100 kHz, a frequency of 1.7 MHz should be connected to SKT3 (1.7 MHz IN/OUT). If the receiver has an i.f. output of 455 kHz, the 1.145 MHz from the synthesizer should be connected to the same socket which is marked 1.145 MHz IN/OUT.

L.F. ADAPTOR CONNECTIONS

23. An l.f. adaptor unit such as the RA. 337 must be supplied with -l6V d.c. power,a.g.c. and l MHz from the receiver. The d.c. connections can be taken from the l2-way connector SKT ll or from the terminal block TB1, whichever is more convenient.

- <u>CAUTION:</u> If links have been fitted to the terminal block TBl, for operation down to 200 kHz, these links must be removed prior to operation with an l.f. adaptor.
 - (1) Connect the 1 MHz OUT socket of the receiver to the 1 MHz input of the L. F. Adaptor.
 - (2) Connect pin K of SKT11 (1.f. h.t.) to the h.t. terminal on the L.F. Adaptor unit.
 - (3) Connect pin L of SKT11 (1.f. a.g.c.) to the a.g.c. input terminal of the L.F. Adaptor unit.
 - (4) Connect the antenna feeder to the antenna socket of the L.F. Adaptor unit.
 - (5) Connect the signal output of the L.F. Adaptor unit to the socket marked 'LF' on the RA.1217. Use 75Ω coaxial cable.
 - (6) Turn the MHz tuning control of the RA. 1217 fully anti-clockwise against its stop. The MHz drum should indicate '00'.
 - (7) Refer to the L.F. Adaptor handbook and set switches as required.
 - (8) Tune the RA. 1217 as described in Chapter 3. Do not move the MHz tuning control.

I.S.B. ADAPTOR CONNECTIONS

- 24. The following instructions apply to the I.S.B. Adaptor Type RA. 298 C & D.
 - (1) Remove the 1.7 MHz crystal from the i.f. converter in the i.f. module of the receiver, and store in a safe place.
 - (2) Connect SKT2 (100 kHz OUT) to the 100 kHz input of the I.S.B. Adaptor.
 - (3) Connect SKT3 (1.7 MHz IN/QUT) to the socket "1.7 MHz to receiver" on the I.S.B. Adaptor.
 - (4) Connect pin H (Diversity A. G. C.) of the l2-way socket SKT l1 to the AGC pin of the l2-way outlet PLl on the I. S. B. Adaptor.
 - (5) Connect pin E of SKT11 (-16V HT) to the -16V pin of PLl on the I.S.B. Adaptor.

(6) Connect pin J of SKT11 (screen) to the 0 volt (earth) pin of PL1 on the I.S.B. Adaptor.

Refer to the RA.298 Technical Manual for further details.

(Refer to the next page for data on input and output levels.)

REAR PANEL SOCKETS

CONCISE DATA

TABLE 1

External Inputs

NOTE: All external source impedances are 75 ohms. If the voltage applied to the receiver exceeds 100 mV an attenuating connector must be used.

Facility	Frequency	Socket	Level (Nominal)
1 MHz frequency Reference	l MHz	1 MHz IN	100 mV e.m.f.
Kilohertz tuning (Synthesizer)	3.6 to 4.6 MHz	2nd V.F.O. IN	100 mV e. m.f
Final i.f. Converter Heterodyne	1.7 MHz 1.145 MHz	(1.7 MHZ IN/OUT 1.145 MHz IN/OUT	100 mV e.m.f.

TABLE 2

External Outputs

Facility	Socket	Level r.m.s.	Termination
1.6 MHz I.F. 1.7 MHz oscillator 100 kHz I.F. 1.145 MHz	1.6 MHz OUT 1.7 MHz IN/OUT 100 kHz OUT 1.145 MHz IN/	100 mV 50 mV 270 mV 50 mV	High Ζ 75 Ω 75 Ω 75 Ω
oscillator 455 kHz I.F 1 MHz reference 3.6 - 4.6 MHz	OUT 455 kHz OUT 1 MHz OUT 2nd V.F.O. OUT	220 mV 50 mV 50 mV	50 Ω 75 Ω 75 Ω

CHAPTER 3

OPERATING INSTRUCTIONS

INTRODUCTION

- 1. Before operating the receiver check that it has been prepared for service in accordance with the information in Chapter 2.
- <u>CAUTION:</u> Before switching on for the first time check that the A. C. /D. C. switch and voltage selector switch at the rear of the receiver are correctly set.

FUNCTIONS OF CONTROLS

2. References to the controls and their settings are in capital letters in accordance with the engraving on the front panel.

SYSTEM SWITCH

- This switch provides facilities for power OFF, MAN, CAL, CHECK B.F.O. and three alternative A.G.C. time-constants. Moving the switch from the OFF position to any other setting connects the external power supply to the power module.
- (2) MAN. In the MAN. position the receiver a.g.c. system is not operative, and the receiver gain is controlled manually by the R.F. GAIN control.
- (3) A.G.C. settings: The use of the time-constants SH (short) MED (medium) and LG (long) is dependent upon operating mode and propagation conditions.
- (4) CAL. In the CAL position, harmonic frequencies derived from a crystal source, provide audio marker notes at discrete 100 kHz intervals of the kHz tuning scale. The tuning scale may be corrected by using the CAL-FINE TUNE control.
- (5) Check B.F.O. In this position the B.F.O. frequency may be set to coincide with the i.f. centre frequency by setting the B.F.O. MODE switch to the '0' position and adjusting the B.F.O. TUNE kHz control to provide an audio null.

R.F. RANGE CONTROL (large knob)	This switch provides for the selection of any one of five frequency ranges, or a wideband (WB) position. The selected range (or WB) is displayed in a small aperture adjacent to the switch. The WB position is very suitable for signal search, but if strong interference is present care must be taken to avoid tuning to an adjacent channel instead of the wanted signal.
R.F.TUNE CONTROL (centre knob)	This control tunes the coil selected by the R.F. RANGE switch. Four tuning marks are engraved adjacent to the control, which together with the associated groups of figures, indicate the approximate frequency for each range at the point of rotation. If maximum sensitivity is not required the antenna need not be tuned, but the presence of strong interfering signals anywhere within the spectrum may cause cross-modulation unless the antenna is tuned.
AE ATT CONTROL. (Antenna Attenuator)	This control may be used to reduce the level of all incoming signals, and it must be used when strong interfering signals are present which cannot be entirely eliminated by use of the R.F. TUNE control, or if the wanted signal is sufficiently powerful to cause overloading of the earlier stages. Very strong signals should be reduced by the use of the AE ATT control rather than by turning down the R.F. GAIN or A.F. GAIN controls.
MHz	This control is the electronic band selector which selects the megahertz portion of the required signal frequency. The control should be set so that the required megahertz, digits are displayed on the tuning readout. A locking arm which projects from behind the knob can be moved fully clockwise to mechanically lock the control. By setting the MHz control to indicate '00', microswitches are actuated which connect power and a.g.c. lines for use in the L.F. Adaptor Type RA.337. Alternatively, the '00' setting can be used for reception down to approximately 200 kHz provided the micro- switches have been linked out. Refer to Chapter 2 for information on the links.
kHz	This control selects the last three digits of the required frequency which are displayed on the tuning readout. The final digit in the readout is marked in divisions to permit a setting accuracy of 0.2 kHz. To avoid the necessity to reset the MHz control when tuning through either end of the band the kHz scale has upper and lower extensions of approximately 25 kHz. The when the perfect of the double of the local ty

25 kHz. Thus when the scale is tuned below 000 kHz a minus sign appears in the readout, indicating that the true frequency

kHz (Cont'd)	is one megahertz less than the figure indicated by the MHz display. Conversely, when the kHz tuning is set above 999 a +sign appears indicating that the true frequency is one megahertz higher than the displayed MHz digits. A locking arm, similar to that on the MHz control, can be moved anti-
	clockwise to lock the kHz control.

- 2nd V.F.O.In the INT position the kHz tuning control operates normally,
using the internal 2nd v.f.o. as already described. In the
EXT Position an external 3.6 to 4.6 MHz source such as a
synthesizer or master receiver can be connected to take the
place of the 2nd V.F.O. and the kHz tuning control is not
operative.
- R.F. A.F.This switch connects the meter to monitor either the r.f. signal
level or the a.f. level of the 1 mW audio line. The meter is
calibrated as an 'S' meter in decibels relative to 1 microvolt.MeterA zero setting adjuster (located within the receiver) is operative
in the R.F. position of the Meter switch. (Fig. 18)
- DET-B.F.O.The positions L.S.B. and U.S.B. select off-set crystalMODE andoscillators for single sideband reception. The A.M. positionB.F.O. TUNEprovides reception without use of b.f.o. The positions +6kHzthrough to -6 provide coarse settings of the b.f.o. heterodynenote.In each of the coarse settings a fine adjustment of theb.f.o. note can be made by the adjacent B.F.O. TUNE kHzcontrol, which provides a variation of approximately plus orminus 3 kHz.
- I.F. BW kHz This switch is marked in the i.f. bandwidths which are provided by crystal filters. Each passband is symmetrical about the i.f. centre frequency. A choice of bandwidths is available at customers option, as follows.

Full Range	Standard Range
kHz	kHz
0.2	0.2
1.2	
3.0	3.0
8.0	8.0
13.0	

R.F. and A.F.		These are concentric controls. The R.F. GAIN is the smaller (central) knob. It functions only in the MAN position of the System switch, in which it manually biases the a.g.c. line. In accordance with normal operating practice the R.F. GAIN should be set to the minimum level consistent with the best signal-to-noise ratio.
		The A.F. GAIN is the larger (peripheral) control. It adjusts the audio level to the headphones and the 1 watt audio output at the rear panel.
A.F.	LEVEL	A preset control which is entirely independent of the A.F. GAIN control. It sets the level of the 1 mW audio line output.
DIMM	ER	This control adjusts the brightness of the tuning readout illumination.

CALIBRATION

The kHz readout and the b.f.o. should be calibrated before using the receiver.

kHz Readout Calibration

3.

- (1) Insert a pair of high impedance headphones into the PHONES socket.
- (2) Set the System switch to CAL
- (3) Set the DET-B.F.O. MODE switch to A.M.
- (4) Set the IF BW switch to 3.0 kHz
- (5) Set A.F. GAIN control the the mid-position.
- (6) Turning the kHz tuning control should result in beat notes being heard at each multiple of 100 kHz.
- (7) Set the kHz tuning control accurately to the 100 kHz multiple which is nearest to the operating frequency required, and adjust the CAL-FINE TUNE control to produce a null in the beat note. To retain calibration the CAL-FINE TUNE control should not be moved.

B.F.O. Calibration

4.

- (1) Insert the headphones in the PHONES socket.
 - (2) Set the DET B.F.O. MODE switch to '0'
 - (3) Set the System switch to CHECK B.F.O.
 - (4) Adjust the B.F.O. TUNE kHz control for a null in the beat note, which should occur approximately in the mid position of the control movement.

RECEIVER TUNING

- NOTE: The following procedure is given as a general guide to the correct use of the controls.
- 5. (1) Connect the headphones
 - (2) Set the LOCK controls to 'off' (upwards)
 - (3) Set the System switch to A.G.C. MED
 - (4) Set the R.F. RANGE switch to WB (wideband)
 - (5) Set the AE ATT control to MIN (fully clockwise)
 - (6) Set the DET-B.F.O. MODE as required (refer to (14) below)
 - (7) Set the MHz tuning control to indicate the required frequency in megahertz on the readout indicator, then lock the control
 - (8) Set the kHz, tuning control to indicate the required kilohertz reading on the the readout indicator. The scale divisions on the final figure will facilitate precise tuning.
 - NOTE: When operating with the R.F. RANGE switch set to WB it is possible for strong interfering signals to render station identification difficult. In such conditions the AE ATT control should be turned one or two steps from MIN towards MAX to obtain the best signal-to-noise ratio consistent with the satisfactory elimination of crossmodulation effects.
 - (9) Adjust the kHz tuning to identify the required signal.
 - (10) Set the R.F. RANGE switch to the appropriate frequency range.
 - (11) Adjust the R.F. TUNE control for maximum signal strength. The markings adjacent to the control give a guide to the approximate tuning point.

- (12) Make fine adjustments of kHz tuning, R.F. TUNE and A.F. GAIN controls for optimum clarity and level of signal.
- (13) If signal strength is excessive, set the AE ATT control in an anti-clockwise direction and re-adjust the A.F. GAIN control for a satisfactory audio level.
- (14) Set the DET-B.F.O. MODE switch to A.M. for A2 or A3 reception, or to a setting between +6 and -6 for A1. Adjust the variable B.F.O. TUNE control for the required pitch of beat note on c.w. or f.s.k.
- (15) Lock the kHz tuning control
- (16) The CAL-FINE TUNE control may be used for fine tuning, if necessary.
- NOTE: When the CAL-FINE TUNE control has been used for fine tune purposes it will be necessary to perform the kilohertz calibration procedure (para. 3) whenever precise kHz readout accuracy is again required.

Single Sideband Reception

6.

- Tune the receiver as instructed in paragraph 5 except that in operation (6) the DET-B.F.O. MODE switch must be set to L.S.B. for lower sideband reception, or to U.S.B. for upper sideband reception.
 - (2) Set the I.F. BW switch to 3 kHz.
 - (3) Adjust the kHz tuning control to obtain the strongest and clearest audio signal.
 - (4) Lock the tuning controls and make fine tuning adjustments with the CAL-FINE TUNE control to clarify the signal.

Operating with a Synthesizer

- 7. (1) Check that the synthesizer is correctly connected as instructed in Chapter 2.
 - (2) Check that the 2nd V.F.O. switch on the RA. 1217 is set to EXT.
 - (3) Switch on the synthesizer and allow a suitable warming-up period (minimum 15 minutes; recommended 3 hours.)
 - (4) Tune the receiver as instructed in para.5 of this chapter, but in operations (8) and (12) the kilohertz frequency is set on the

synthesizer dials. The receiver kHz, tuning control is not in use and the CAL-FINE TUNE control cannot be used for fine tune purposes. Refer to the synthesizer handbook for kHz tuning instructions.

Operating with an L.F. Adaptor

8.

- The l.f. adaptor should be connected to the receiver as instructed in Chapter 2. The antenna should be connected to the l.f. adaptor.
 - (2) Set the MHz tuning control on the RA. 1217 fully anticlockwise so that '00' is indicated on the megahertz readout.
 - (3) Set the RA. 1217 System switch to A.G.C. (MED)
 - (4) Tune the RA. 1217 using the kHz control only. The MHz control R.F. RANGE, R.F. TUNE and AE ATT controls are not in use. The remaining controls may be used as in normal operating.

ADDITIONAL OPERATING INFORMATION

Receiving Down to 200 kHz

9. The following modification will give reception down to approx.200 kHz but with degraded performance.

- (1) At the rear of the receiver link the terminal 'h.t., l.f.' to terminal 'h.t., r.f.'
- (2) Link terminal 'a.g.c., l.f.' to terminal a.g.c., r.f.' This bypasses the micro-switches which normally disconnect h.t. and a.g.c. from the receiver input stages when the MHz tuning is set to '00'.
- (3) Set the 'MHz' tuning control to indicate '00' on the MHz readout. Lock the control.
- (4) Tune with the kHz control and all other controls in the normal way.
- NOTE: The links should be removed when an L.F. Adaptor unit is to be connected.

CHAPTER 4

BRIEF TECHNICAL DESCRIPTION

INTRODUCTION

 This chapter describes the basic principles of the receiver design. Only the more significant design features will be referred to. A detailed description of the circuits is given in the Maintenance Handbook. Throughout this chapter reference should be made to the block diagrams, Fig. 2 and Fig. 3 at the rear of the book.

WADLEY SYSTEM

- 2. The main feature of the receiver is the tuning system which is based on a principle often referred to as the 'Wadley system (fig. 2). This system is specifically designed to counteract frequency drift. Its most characteristic features are that the tuning system is separated into MEGAHERTZ and KILOHERTZ tuning, and no mechanical band switching is required, although pre-tuning of the antenna circuit may be provided.
- 3. The 'Megahertz' tuning section selects a 1 MHz band of signal frequencies, which contains both wanted and unwanted signals, and transfers this band of frequencies to the 'Kilohertz' section. The 'Kilohertz' section serves as an interpolation receiver, and selects the wanted signal from within this 1 MHz band. It is essential to grasp this conception of a spectrum of frequencies being transferred, rather than a single frequency as in a conventional superheterodyne system.

MEGAHERTZ TUNING SECTION

Drift Cancellation

4. The 'Megahertz' tuning system is specifically designed to eliminate the effects of drift in the first v.f.o. stage. Any tendency for the receiver to drift off tune due to first v.f.o. instability is automatically corrected by a technique of self-cancelling frequency change.

5. For descriptive purposes the system may be divided into two loops which will be referred to as the 'signal loop' and the 'harmonic mixer loop'.

The first v.f.o. is tuned by the Megahertz tuning control, and its output is supplied to both loops.

6. In the signal loop one of the first v.f.o. outputs is mixed with the incoming r.f. signals to produce an output spectrum slightly more than 1 MHz wide. In the harmonic mixer loop another output from the first v.f.o. is mixed with a harmonic from a 1 MHz crystal oscillator. These separate mixer outputs are filtered and then subtracted in the second mixer stage. Since the first v.f.o. has contributed equally to both loops, any drift in the signal loop is

cancelled by an equivalent drift in the harmonic mixer loop. Thus drift or displacement of the first v.f.o. frequency has no effect on the receiver output. The stability of the Megahertz section is determined by the 1 MHz reference oscillator. This crystal oscillator can be replaced by an external 1 MHz source of even greater stability, if required. The following paragraphs describe the 'Megahertz' tuning in more detail.

Signal Loop

(Fig. 3)

7. This loop comprises the r.f. input stage, the first mixer, the first v.f.o. and the 40 MHz band-pass filter. Incoming r.f. signals between 0.98 MHz and 30 MHz are amplified and fed to the first mixer, together with an output from the first v.f.o. This produces a frequency spectrum (first i.f.) centred on 40 MHz with a bandwidth of ±650 kHz at the output from the 40 MHz band-pass filter. This 1 MHz spectrum contains the wanted signal together with other, unwanted signals.

8. The first v.f.o. can be tuned by the Megahertz (MHZ) control over the range 41.5 MHz (setting 01) to 69.5 MHz (setting 29). At each setting there will be a particular 1 MHz spectrum within the 0.98 to 30 MHz signal range which will produce an output from the first mixer which can pass through the 40 MHz band-pass filter. Thus, by tuning the first v.f.o. the entire signal range of 0.98 to 30 MHz can be selected in discrete 1 MHz steps without mechanical band switching. It should be noted that the passband of the 40 MHz filter exceeds 1 MHz by a tolerance of ±150 kHz. This will be referred to later. A separate 29-channel oscillator unit can be supplied which replaces the first v.f.o., thus providing remotely controlled 'megahertz' tuning.

Harmonic Mixer Loop

9. This loop comprises the 1 MHz crystal oscillator, the harmonic generator, the harmonic mixer and the 37.5 MHz band-pass filter. The 1 MHz oscillator drives the harmonic generator which produces a wide range of harmonics of the 1 MHz crystal. An output from the first v.f.o. is applied to the harmonic mixer together with all the harmonics (up to the 32nd) of the 1 MHz crystal.

10. The output of the harmonic mixer is a band-pass filter with a pass-band of plus or minus 150 kHz centred on 37.5 MHz. Thus, whatever the setting of the first v.f.o. tuning, an output is obtained only when the first v.f.o. frequency minus a harmonic of 1 MHz crystal produces 37.5 MHz. This occurs at discrete 1 MHz steps in the first v.f.o. tuning range, therefore the Megahertz tuning control must always be set in 1 MHz steps.

11. It will be noted that the passband of the 37.5 MHz filter is the same

 $(\pm 150 \text{ kHz})$ as the tolerance in the 40 MHz band-pass filter. This is the permitted allowance for first v.f.o. drift or tuning error, from which it is evident that the setting of the Megahertz tuning control is not critical. The control should, however, always be set to the point of maximum receiver noise, thus ensuring that the mixer output frequencies are centred in their respective filter passbands.

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Second Mixer

12. The signal loop and the harmonic mixer loop are brought together in the second mixer where the 37.5 MHz is subtracted from the 1 MHz spectrum (39.5 MHz to 40.5 MHz) of the signal loop. The resultant output (2nd i.f.) is a 1 MHz spectrum covering the range 2 MHz to 3 MHz. This is amplified, filtered to a more exact 1 MHz bandwidth, and applied to the third mixer, where, by means of the second v.f.o. the 'Kilohertz' tuning is performed. Numerical examples of 'Megahertz' tuning are given in the next paragraph.

Examples of Megahertz Tuning

13.

TABLE 1

(1) MHZ co	ontrol set to		5 MHz
<u>Signal Loop</u>	MHz	<u>Harmonic Loop</u>	MHz
VFO - 1 Signal Frequency	45.5 5 to 6	VFO – 1 Harmonic (8)	45.5 8.0
lst i.f.	39.5 to 40.5	Mixer Output	37.5

2nd Mixer

2nd I.F. = 2 to 3 MHz

(2)	MHZ	control set to	• • • • • • • • • • • • • • • • • • •	29 MHz
<u>Signal</u>	Loop	MHz	Harmonic Loop	MHz
VFO - Signal Freque		69.5 29 to 30	VFO - 1 Harmonic (32)	69.5 32.0
lst i.f.		39.5to40.5	Mixer Output	37.5

2nd Mixer

2nd I.F. = 2 to 3 MHz

KILOHERTZ TUNING

Third Mixer and Second V.F.O.

14. The 2 to 3 MHz second i.f. spectrum contains the wanted signal. This signal frequency is precisely tuned in the third mixer stage by the second v.f.o. (KHZ tuning) which has a tuning range of 3.6 to 4.6 MHz. This second v.f.o. output is mixed with the second i.f. and the difference frequency of 1.6 MHz becomes the third intermediate frequency. Thus the third mixer and second v.f.o. may be regarded as an interpolation receiver operating over the range 2 to 3 MHz. For optimum stability the second v.f.o. may be replaced by an external frequency synthesizer or controlled by an automatic frequency stabilizer such as the "Racalator".

4 - 3

Frequency Inversion

15. It should be noted that the 1 MHz spectrum in the antenna circuit is transferred and inverted to a 3 to 2 MHz spectrum in the second i.f. stage. For example a 15.45 MHz signal becomes 2.55 MHz in the second i.f. stage, and a 15.75 MHz r.f. signal becomes 2.25 MHz. This inversion is of no significance to the user unless the design or suitability of an 1.f. adaptor or synthesizer is under consideration. The inversion arises because the first v.f. o. frequency is above the signal frequency whereas the harmonic mixer output (37.5 MHz) is below the first i.f. A frequency from the harmonic mixer which is higher than the first i.f. cannot be employed as it would require for its generation a harmonic of zero order when tuned to the 2 to 3 MHz signal band. (1st V.F.O. tuned to 42.5 MHz).

Calibration

16. The setting accuracy of the kilohertz tuning control can be checked at 100 kHz intervals by means of a calibrator circuit when the system switch is set to CAL. The calibrator circuit generates a 100 kHz fundamental obtained from a divider circuit driven by the 1 MHz crystal reference source. The 100 kHz is fed to the third mixer where the harmonics beat with the second v.f. o. frequency. The kHz control is set to indicate a multiple of 100 kHz and the CAL-FINE TUNE control on the front panel is then adjusted to shift the 2nd v.f. o. frequency which produces a zero beat frequency in the phones.

17. The CAL-FINE TUNE control is a variable resistor which adjusts the bias on a varactor diode in the second v.f.o. tuned circuit. This control can also be used to make fine adjustments of signal tuning when, for example, the main tuning controls are locked. The frequency range of the fine tune adjustment varies over the kilohertz tuning range but should be not less than 8 kHz.

L.F. Adaptor Unit

18. To receive frequencies lower than 0.98 MHz an external l.f. adaptor is required, the output from which is a 1 MHz spectrum in the band 2 to 3 MHz. This signal spectrum is fed into the second i.f. stage in lieu of the spectrum from the second mixer. When the l.f. adaptor is in use the receiver r.f. stage, first v.f.o., first mixer and harmonic mixer are not required and the h.t. supply to these stages is disconnected. Terminals on the rear panel of the receiver provide h.t. and a.g.c. connections for an l.f. adaptor unit. The h.t. and a.g.c. are automatically switched through to the l.f. adaptor when the receiver MHZ tuning control is set to '00'.

Panoramic Adaptor Unit

19. The 2 to 3 MHz second i.f. spectrum can be displayed on a panor-amic adaptor unit, thus permitting visual display of the activity in any selected 1 MHz band of signals between 980 kHz and 30 MHz. (Down to 3 kHz if the 1.f. Adaptor RA. 337 is connected). The signal input to the Pan

Adaptor is taken from the coaxial outlet PAN at the rear of the receiver. The -16V and -24V power supplies for the Pan. Adaptor may be obtained from the RA. 1217 receiver via the appropriate pins on the 12-way outlet.

TABLE 2

OVERALL TUNING EXAMPLE

RA. 1217 Tuned to a Signal of 15. 48 MHz

Dial Settings	
Megahertz control set to indicate Kilohertz control set to indicate	
Signal Loop	MHz
VFO - 1 Tuning (MHZ) set to VFO - 1 Frequency (40.5 +15) 1st i.f. spectrum (55.5 minus 15 to 16)	15 55.5
Wanted signal within the spectrum is 55.5 minus 15.48	40.02
Harmonic Mixer Loop	
VFO - 1 frequency VFO - 1 minus 18th harmonic of 1 MHz	
2nd Mixer	
lst i.f. spectrum minus 37.5 MHz provides 2nd i.f. Wanted signal within the 2nd i.f. spectrum is 40.02 minus 37.5 MHz	2 to 3
3rd Mixer and 2nd VFO	
VFO - 2 tuning (KHZ) set to 480 which tunes VFO - 2 to required setting (2.52 +1.6) 3rd i.f. is (4.12 minus 2.52) which is	
100 kHz i.f. Converter	
Converter oscillator (1.7 MHz) minus 3rd i.f. (1.6 MHz)	100 kHz
455 kHz i.f. Converter	
1.6 MHz minus converter osc.	

1.6 MHz minus converter osc.	
(1.145 MHz)	
FINAL I.F. STAGES

20. The final intermediate frequency used within the receiver is 1.6 MHz. Four amplifiers at this frequency provide the greater part of the receiver gain. Crystal band-pass filters are included to provide the requisite selectivity. An additional intermediate frequency is available for external use with ancillary equipment such as an i.s.b. adaptor, f.s.k. converter, etc. The particular external frequency available (either 100 kHz or 455 kHz) is dependent upon the type of i.f. module fitted to the receiver and should be specified by the purchaser. (Refer also to para. 32).

A.G.C. CIRCUITS

21.

A.G.C. is obtained by detecting an output from the 1.6 MHz i.f. Three time-constants are available, selected by the System Switch. An interconnection via the I.F. Bandwidth switch ensures that the short timeconstant is not operative whenever the I.F. Bandwidth switch is set to either 0.2 or 1.2 kHz. If the System switch is set to A.G.C. SH (short) the Bandwidth switch inserts A.G.C. medium time-constant whenever either of these narrow bandwidths are selected. The receiver gain can be controlled from an external a.g. c. source such as a companion diversity receiver, or an i.s.b. adaptor etc.

R.F. GAIN Control

22. In the MAN position of the System switch the a.g.c. line is disconnected from the a.g.c. source and a manually controlled bias derived from -16V via the R.F. GAIN control is substituted. In all other settings of the System switch the R.F. GAIN control has no function

DETECTOR AND B. F. O. CIRCUITS

Detector Circuits

23. The detector circuit has two separate functions:

- (a) As a product detector in all positions of the B.F.O. switch except A.M.
- As an a.m. demodulator in the A.M. position of the B.F.O. (b) switch.
- 24. Product Detector. The output from the i.f. amplifier stage is fed to the product detector together with one of the following frequencies, dependent upon the setting of the B. F. O. switch.
 - (a)1601.5 kHz (B. F. O. switch at L.S. B.) or
 - 1598.5 kHz (B. F. O. switch at U.S. B.) (b)
 - A 1.6 MHz frequency which is variable by (c) means of the B.F.O. TUNE control. (B.F.O. switch at +6 through -6).

25. <u>A. M. Detector.</u> In the A. M. position of the B. F. O. switch the 1.6 MHz i.f. signal is demodulated. The B. F. O. frequencies referred to in the previous paragraph are switched out.

B.F.O. Circuits

26. <u>CW (Al) Reception</u>. The local heterodyne oscillator provides a 600 kHz nominal frequency which can be varied through a range of plus or minus 9 kHz by suitable combinations of the switch settings +6 to -6, and variable adjustment of the B. F. O. TUNE control.

B.F.O. Circuits

27. The variable 600 kHz output is fed via a buffer amplifier to a mixer on the detector board in the i.f. module. A fixed frequency of 1 MHz from a crystal oscillator (the same source as that which drives the harmonic generator) is supplied to the same mixer. The combined output at 1.6 MHz (plus or minus any variation applied to the 600 kHz via the B.F.O. TUNE control) is supplied to the product detector stage to heterodyne the 1.6 MHz intermediate frequency.

28. Single-Sideband Reception. The positions L.S.B. and U.S.B. on the B.F.O. switch provide electronic switching of two oscillator frequencies on the detector board. In the L.S.B. setting a crystal controlled frequency of 1601. 5 kHz is generated. This is fed to the product detector stage to serve as a stable inserted carrier 1.5 kHz above the 1.6 MHz i.f. thus permitting lower sideband reception. In the U.S.B. setting of the B.F.O. switch a 1598. 5 kHz crystal is substituted in the oscillator circuit which then provides carrier insertion 1.5 kHz below the 1.6 MHz i.f. thus facilitating upper sideband reception. The 3 kHz i.f. bandwidth setting should be used so that the inserted carrier is placed at one edge of the filter thus allowing the required sideband to occupy the whole of the passband.

29. It may be noted that the mixer to the 1 MHz and 600 kHz frequencies (para. 27) becomes an amplifier when the B. F. O. switch is set to L. S. B. or U. S. B. The 600 kHz b. f. o. is switched off, thus there is no variable b. f. o. facility available when the L. S. B. or U. S. B. positions are selected.

AUDIO STAGES

30.

The output from the a.m. detector or the product detector is fed via a buffer stage to an audio amplifier board located on the

underside of the receiver. The standard amplifier supplies the following outputs.

- (a) 1 mW in 600 ohms for line use.
- (b) 10 m W into 600 ohms.
- (c) 10 mW in 600Ω to the headphones outlet on the front panel.

The level of the 1 m W output is adjusted by the preset A.F. LEVEL control. The 10 m W output level is adjusted by the variable A.F. GAIN control. To special order an audio amplifier is available which provides outputs of one watt and one milliwatt. Refer to the Preface at the beginning of Chapter 1.

I.F. CONVERTERS

31. Two alternative types of converter board are available for fitting into the i.f. module. The particular type fitted is determined whether the intermediate frequency output is required to be 100 kHz or 455 kHz.

32. The 1.6 MHz intermediate frequency is supplied to the converter board where it is mixed with the output from a crystal controlled oscillator, the difference frequency is the required final i.f. Thus if the final i.f. is to be 100 kHz the converter oscillator crystal is required to be 1.7 MHz. For an i.f. of 455 kHz a crystal frequency of 1.145 MHz is required. These crystal frequencies can be injected from an external source such as a synthesizer. When an external source is employed the internal crystal should be removed.

POWER SUPPLY MODULE

33. The standard power module Type PU. 1153 is a plug-in unit which can operate from either a.c. or d.c. supplies. It provides outputs at -16V (regulated) and 20 to 24V nominal (unregulated) capable of supplying the power requirements of the receiver and also ancillary units such as an 1.f. adaptor, i.s.b. adaptor etc. Refer to the Technical Specification at the front of this handbook for performance data.

34. The A. C. / D. C. and Voltage Selector switched must be correctly set before operating the receiver, as instructed in Chapter 2. Two fuses are provided; the 250 mA fuse protects the input when an a.c. supply is in use. The h.t. fuse, which is in circuit with both a.c. and d.c. power inputs, protects the -16V and -24V outputs.

35. The receiver has no independent power switch. The input to the power supply is connected via a pair of microswitches which are actuated when the System switch is moved from the OFF position.

CHAPTER 5

INITIAL FAULT LOCATION

INTRODUCTION

1. The advice in this chapter provides a simple test procedure which will assist the location of an elementary fault. It is assumed that the only instrument available is a universal testmeter, and that the receiver is not connected to a synthesizer, 1. f. adaptor or other external unit. If the fault cannot be located it will be necessary to refer to the more detailed test procedures described in the RA. 1217 maintenance handbook.

PRELIMINARY CHECKS

2. If the receiver is newly installed check the following items:

- (a) AC/DC selector switch correctly set. (rear panel).
- (b) 2nd V. F. O. switch set to INT (front panel).
- (c) MHz tuning control not set to '00'
- (d) Power connection: Note that the polarity of a d.c. supply must be correct. (Refer to Chapter 2).
- (e) Antenna connected

INITIAL FAULT LOCATION PROCEDURE

Controls

3.

Set the receiver as follows and check for signals or noise.

- (1) System switch to MAN
- (2) DET-BFO switch to A. M.
- (3) Meter switch to R. F.
- (4) A. F. GAIN to maximum (clockwise)
- (5) I.F. BW switch to 3 kHz
- (6) R. F. GAIN to maximum (clockwise).
- (7) AE ATT control to the MINIMUM position (MIN)
- (8) R.F. RANGE switch to 'WB'.
- (9) Ensure that the MHz tuning control is not set to '00'

General Diagnosis

4. The most useful indication in elementary fault diagnosis is receiver noise, or 'mush'. The controls should be set as listed in paragraph 3 and the receiver tuned over a suitable portion of the h.f. band. At each step of the MHz tuning control make a fine adjustment and listen for a rise in receiver noise level. If no noise can be heard, check that the phones are serviceable and, if possible, listen at an alternative audio outlet as well as at the phones jack socket.

POWER CHECK

- 5. If the receiver appears dead (no noise) and the dial lights are not illuminated, check the h.t. voltage in the receiver as follows:
 - (1)Connect the testmeter (30 volt d.c. range, or higher) to the terminal 'RF. HT' on terminal block TBl at the rear of the receiver. The reading should be -16 volts relative to chassis. If no reading is obtained at this point it is probable that a fault exists in the power unit or power supply. If connected to a d.c. supply check that the polarity of the connections is correct. (Refer to Chapter 2).

I.F. MODULE CHECK

- 6. If noise can be heard, vary the setting of the I.F. BW switch. A change of noise level at each switch movement indicates that the i.f. module is serviceable.
- 7.

If no noise is heard, switch on the b.f.o. (DET-BFO switch to

+3 or -3 kHz). The meter (R.F. position) should show a reading and b.f.o. noise should be heard. If a meter reading is obtained but no noise is heard, the fault may be in the detector board of the i.f. module, or in the audio amplifier and its connections. If b.f.o. noise is heard the fault may be in the i.f. amplifier stages of the i.f. module, or in various stages of the receiver prior to the i.f. module. Carry out a front end check.

FRONT END CHECK

8.

Set the DET-BFO switch to A. M. Listen intently and slowly rotate the MHz tuning control. If a very slight rise in noise level can be heard as the MHz tuning passes through each resonant point it suggests that the 37.5 MHz loop is functioning and therefore the fault is more likely to be in the antenna circuit, R.F. Module or 1st Mixer. Make the check in the WB setting of the R.F. RANGE switch as well as in the tuned antenna condition (adjust RF TUNE control). Thoroughly check all front end connections as follows:

- (1)Check antenna.
- (2)Check continuity through the 500 mA fuselink in the R.F. Module.
- (3) Check that the antenna spark gap is not short circuited.
- (4) Check the flying lead connections between the R. F. Module and the 1st Mixer, and between 1st VFO and 1st Mixer.

If, when tuning the MHz control as described in the previous

paragraph, no noise can be heard, the 1st VFO or its connections may be faulty. Check connections from 1st V.F.O. to 1st Mixer and 37.5 MHz Generator at the bracket on the underside of the main chassis.

RF HT Check

10. On the terminal block TBl at the rear of the receiver connect the terminal HT RF to the terminal HT LF. If the receiver then functions correctly the microswitch ISB (Fig. 16, Fig. 18) should be checked.

1 MHz Check

11. Set the System switch to CAL and tune the KHz control to the 100 kHz calibration check frequencies. If the calibration whistle is heard at each point it indicates that the 1 MHz oscillator is functioning. If no calibration whistles are heard, turn the System switch to CHECK BFO and set the DET-BFO switch to +6, +3, -3 and -6 kHz in turn. If, again, no heterodyne whistles are heard, it indicates a faulty 1 MHz crystal oscillator. Check that the crystal is correctly fitted in the upper deck of the 37.5 MHz Generator Module.

A.G.C. FAULT

If the receiver operates satisfactorily with manual r.f. gain control (System switch to MAN) but overloads on strong signals in the a.g. c. settings of the System Switch check as follows:

- Tune the receiver to a strong signal. Set the System switch to AGC Med and the Meter switch to R. F. If the meter indicates a reading appreciably greater than 1 microvolt the a.g.c. board in the I. F. Module is serviceable. If no reading is obtained the fault is probably in the I. F. Module.
- (2) If the meter reading is satisfactory, connect the test meter negative lead to the terminal AGC RF on the rear panel (positive lead to chassis). As the receiver is tuned through a powerful signal the a.g.c. level should change from -4V (weak signal) to approximately 0 volts (strong signal). If no reading is obtained check the microswitch ISA adjacent to the MHz tuning shaft. (Fig. 16 Fig. 18)
- NOTE: The levels quoted in para. 12, cannot be given exactly because the level of a strong signal is not defined.

9.



281/2

Wadley System - Block Diagram

Fig.1

Electronic Band Selection — Explanatory Block Diagram Fig. 2



273

Block Diagram + RA1217

F



Circuit: R.E. Mo'dule

19





Circuit: 1st. V. F. O

Fig 5



Circuit : IMHz Amplifiers, Oscillator and Calibrator





Circuits:Harmonic Generator Mixer and 37.5MHz Filter (37.5MHz Generator)





Circuits:Harmonic Generator Mixer and 37.5MHz Filter (37.5MHz Generator)

Fig.7















30 39456 28170

Fig. II





DC 38560 281/12





Circuit : 1.F. Module

DC 38560 281/12

Fig. 12



0C 38450 281/13



DC 38450	280/45



Fig.14b





DC. 384 50/A8CE 281/16



Interconnections : RA. 1217







201/17



Chassis Upperside RA 1217

Fig.18



RA. 1217 MAINTENANCE MANUAL

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Frequency Range:	l to 30 MHz
Modes of Reception:	D.S.B. M.C.W., C.W., S.S.B. (U.S.B. or L.S.B.)
Tuning:	Digital presentation in units of kilohertz/ second with interpolation calibration at 200 Hz intervals.
Resetting Accuracy:	±200 Hz
Calibration Accuracy:	±500 Hz (when calibrated to nearest 100 kHz checkpoint)
Calibration:	A 100 kHz signal, derived from the 1 MHz standard crystal oscillator having an accuracy of 5 parts in 10^6 , provides check points at 100 kHz intervals.
Frequency Stability:	 (a) ±50 Hz over an eight hour period at constant ambient temperature and humidity after 2 hours from switching on.
	(b) Better than 50 Hz per degree Centigrade at constant humidity after 2 hours from switching on.
Antenna Input:	(a) Nominal impedance 75 ohms unbalanced
	(b) Wideband, or tuned in five automatically selected bands:
	 (i) 1 to 2 MHz (ii) 2 to 4 MHz (iii) 4 to 8 MHz (iv) 8 to 16 MHz (v) 16 to 30 MHz
Sensitivity:	In 3 kHz b andwidth:
	(a) C. W., S.S.B: 1 μV for 15 dB signal/ noise ratio.
	(b) M.C.W., D.S.B: 30% modulated at 400 Hz: 3 μV for 15 dB signal/noise ratio.

Selectivity:	Three i.f. bandwidths fitted as standard, although positions available for up to five filters.		
	Standard Filters:3 dB Bandwidth6/60 dB Shape Factor8 kHz1:43 kHz1:4200 Hz1:10		
	Additional Filters Available:13 kHz1:41.2 kHz1:4500 Hz1:10		
Cross Modulation:	For a wanted signal of level up to 1 mV and with appropriate use of Aerial Attenuator (AE ATT) control, an interfering signal 20 kHz removed and modulated 30% must have a level greater than 45 dB above that of the wanted signal to produce a cross modulation of 3%. The ratio of wanted to unwanted signal level is improved at the rate of 2 dB for each one percent up to 10 percent off tune.		
Intermodulation:	Better than 80 dB on 1 microvolt for inter- fering signals at least 10% removed from the wanted signal.		
Blocking:	For levels of wanted signal up to 1 mV, and with appropriate use of the Aerial Attenuator (AE ATT control) an interfering signal 20 kHz removed must have a level 56 dB greater than the wanted signal to re- duce the signal by 3 dB. The ratio of wanted to unwanted signal level is improved at the rate of 2 dB per one percent up to 10% off tune.		
Spurious Response to External Signals (Image, etc):	(a) External signal less than 10% off tune shall be greater than plus 60 dB rela- tive to 1 microvolt to produce a spur- ious signal equivalent to 1 microvolt.		
	(b) With tuned aerial (antenna) external signals more than 10% off tune shall be greater than plus 80 dB relative to 1 microvolt to produce a spurious signal equivalent to 1 microvolt.		
Automatic Gain Control:	(a)	Time constants (nominal):	
--------------------------	---	--	-------------------------------------
	Shor Medi Long		Discharge 20 mS 250 mS 4 S
	(b)	Output change:	
	will	An increase in input of 85dB above 2 microvolts will produce a change in output level of less than 4 dB.	
B. F. O:	(a)	Variable: ± 3 kHz centred -3 and -6kHz with respect frequency.	
	(b)	Fixed: ± 1.5 kHz (USB/LS) controlled.	B) crystal
A.F. Output:	(a)	Headphone jack on front panel: 10 mW nominal in 600 Ω .	
	(b)	l0 m W in 600Ω at rear terminals. An alternative version providing one watt into 15 ohms for an external loudspeaker is available to special order.	
	(c)	l m W in 600Ω 'line' outlet. level is independent of the control setting.	
Overall A.F. Distortion:	Less	than 5%.	
Overall A.F. Response:	With a selectivity handpass of 13 kHz the audio frequency output level from 100 Hz to 6000 Hz. will not very more than 4dB from peak response.		
Meter Indication:	(a)	R.F. Levels.	
	(Ъ)	A.F. level to line.	

Controls:

- (a) Meter Switch (AF level/R.F. level)
- (b) Frequency Readout In-line Digital
- (c) 'Megahertz' Tuning
- (d) 'Kilohertz, Tuning
- (e) System Switch
- (f) DET/B.F.O. Mode
- (g) B.F.O. Tuning
- (h) R.F./I.F. Gain
- (i) A.F. Gain
- (j) A.F. Line Level
- (k) R.F. Tune
- (1) R.F. Bandswitch
- (m) Tuning Locks
- (n) Dimmer (scale illumination)
- (o) Calibrate Fine Tune
- (p) Bandwidth Switch
- (q) R.F. Attenuator
- (r) 2nd V.F.O. INT/EXT switch
- (a) Antenna Input
- (b) I.F. Outputs
- (c) A.F. Line Output
- (d) External L.S. Output
- (e) 2nd V.F.O. Output
- (f) 2nd F.V.O. Input
- (g) 1 MHz Output
- (h) 1 MHz Input
- (i) 1.7 MHz input/output
- (j) L.F. Adaptor Input
- (k) Panoramic Adaptor Output
- (1) A.C. Supply input

Page 4 Tech. Spec.

External Connections:

	(m)6 volts output		
	(n) Diversity A.G.C.		
	(o) Muting Contact		
	(p) L.F. Adaptor h.t.		
	(q) L.F. Adaptor a.g.c.		
Power Supplies:	Alternative power units are available as follows:		
	(a) 100-125 or 200-250V, 45-400 Hz, a.c. single phase; or 21-27V d.c.positive earth.		
	(b) 9-15V or 18-30V d.c. positive or n egative earth.		
Power Consumption:	40VA approx. (with one-watt audio amplifier)		
Dimensions:	High Wide Deep		
	3.5 in. x 19 in. x $16\frac{1}{2}$ in. (8.9 cm) (48.25 cm) (41.9 cm)		
Weight:			
Weight: Environmental Conditions:	(8.9 cm) (48.25 cm) (41.9 cm)		
C C	 (8.9 cm) (48.25 cm) (41.9 cm) 31 lb. (14.2 kg) approximately. The equipment is, designed to meet certain of the requirements of specification DEF 133 L2, operating within the ambient temperature 		

RA.1217 MAINTENANCE MANUAL

PREFACE

Handbook Guide

This Volume contains detailed information for maintenance purposes. It is assumed that the user also has a copy of the RA.1217 Operators Manual, to which reference should be made for general information concerning installation, operation, and technical principles.

Variants

Chapters 1 to 8 of this Volume refer to the basic RA.1217 receiver. Details of any variant models will be given in appendices, which will be inserted immediately prior to the illustrations at the rear of the handbook.

CHAPTER 1

DETAILED CIRCUIT DESCRIPTION

INTRODUCTION

 The RA. 1217 is constructed on a modular system, the separate modules themselves being composed of smaller assembly units.
 This type of construction lends itself to a servicing system based on replacement units and simplifies the incorporation of improved designs or special facilities. Many of the board assemblies contain further coil or transformers sub-assemblies etc., but it is strongly recommended that the user should regard the printed circuit board assembly as the smallest item suitable for unit replacement. In accordance with this principle the receiver can be dismantled into the following main items. Each principal heading indicates a module, with the printed circuit boards contained in that module listed beneath.

MODULE ASSEMBLIES

2.	Units and Sub-Assemblies		Racal Part Number
R.F	. Module Assembly	Fig. 4	CA.38538/A
	Aerial Filter Coil and Switch Assembly R.F. Amplifier Board H.T. Filter (not located in the R.F. Module)		AA. 28188 BA. 34082 BA. 28185 AA. 28179/B
lst	Mixer and 40 MHz Filter Assembly	Fig. 8	BA.28211
1.	lst Mixer Board		BA.28215
2.	40 MHz Filter		AA.28197
2nd	Mixer Assembly	Fig. 9	BA.3095 9
1.	Mixer Board		BA.28177
3rd	Mixer Assembly	Fig. 11	BA.35970
1.	Mixer and Filter Board		BA.35966
lst	V.F.O. Assembly	Fig. 5	CA.28120
1.	Oscillator Board		BA.35195
2.	Buffer Amp (to 1st Mixer)		BA. 28128
3.	Buffer Amp (to Harmonic Mixer)		BA.32535

<u>Units</u> and	Sub-Assemblies		Racal Part Number
2nd V. F. O. Assembly	y_	Fig. 10	CA.28101/B
 Oscillator Board Buffer Amplifiers 			BA. 35808 BA. 35807
1 MHz Amps, Osc. a	nd Calibrator	Fig. 6	(Part of CA.28276/B)
 1 MHz Amplifie 2. 1 MHz Osc. and 		·	BA. 32858 BA. 32860
37.5 MHz Generator		Fig. 7	(Part of CA. 28276)
 H.T. Filter Harmonic Gener 	rator Board		BA. 28284 BA. 32854
$\left(\begin{array}{c} 3.\\ 4.\end{array}\right)$ Harmonic Filter	r		BA.35836
5. Harmonic Mixer 6. 37.5 MHz Bulfe			BA. 37894 BA. 32850
** The overall module (CA. 28276) is usually referred to as the '37.5 MHz Generator' module although it also contains the 1 MHz and calibrator section.			
37.5 MHz Band-Pass	Filter	Fig. 7	AC. 28192
I.F. Module Assemb	ly (455 kHz i.f.) (100 kHz i.f.)	Fig. 12	DA. 38580/A DA. 38580/B
 I. F. Amplifier Detector Board A. G. C. Board Convertor Oscil Convertor Amplifier 	lator Board (455 kHz) (100 kHz)		BA. 30533 BA. 28236 BA. 31466/B BA. 38568A BA. 38568B BA. 34783A BA. 34783B
B.F.O.		Fig. 13	3
 600 kHz Oscilla Buffer Amplifie 			BA.30540 BA.30542
Power Unit PU. 1153 (a. c. /d. c.) Fig. 15a			5a
 Complete Assen Component Boar 	nbly (with Componen rd Assembly	t board)	CA. 38640 BA. 37549

MAIN CHASSIS GENERAL DETAILS

Before commencing a detailed description of each module a few points concerning the main chassis may be noted. The most suitable illustrations to refer to are the Interconnections diagrams Fig. 16 and 17. The location of modules in the main chassis is shown in Fig. 18.

H.T. DISTRIBUTION

4. The -16V h.t. supply from the Power Unit is supplied, without switching, to the 3rd Mixer, 1 MHz oscillator, I.F. Module and B.F.O. The -16V supply to those stages prior to the 3rd Mixer, which are the R.F. Module, 1st Mixer, 2nd Mixer, 1st V.F.O. and 37.5 MHz
Generator circuit, is controlled by the System switch. The switched h.t. to these units is routed via the microswitch SB and the terminal block TB1 which enables the switched h.t. to be connected to an external l.f. adaptor unit, if required.

5. The microswitch SB connects h.t. to the terminal 'h.t. r.f.' in all settings of the MHz tuning control except '00'. When the MHz control is turned to display '00' on the scale, a cam moves the microswitch SB to the opposite contact, thereby transferring the h.t. to the terminal 'h.t. l.f.' on TB1. For operation below l MHz the terminals 'h.t. r.f.' and 'h.t. l.f. can be linked. This link should be removed when the Racal l.f. adaptor is connected.

System Switch SE

Fig. 16

6. In the OFF position the external supply to the power unit is disconnected by the opening of the switch contacts SCl and SC2 which are mounted on an extension of the shaft of the system switch SE. The function of each wafer of switch SE is described as follows:

7. <u>Wafer SE2R</u>: This wafer controls the distribution of -16V h.t. to certain circuits as follows: In positions CHECK B.F.O. and CAL, -16V is supplied to the calibrator circuit and disconnected from the units prior to the 2nd Mixer. In the MAN and A.G.C. positions of the switch the -16V is connected to the circuits prior to the 2nd Mixer, and disconnected from the calibrator circuit.

8. <u>Wafer SEIR</u>: In all settings except MAN, this wafer connects the a.g.c. line to the a.g.c. amplifiers in the i.f. module. In MAN. the a.g.c. line is connected to the slider of the R.F. Gain control.

9. <u>Wafer SE3F</u>: This wafer provides an earth connection to the b.f. o. switch (SB1F). In the CAL position the earth is dis-

connected in order to disable the b.f.o. during the calibration procedure.

10. <u>Wafer SEIR</u>: An earth is provided only in the three a.g. c. positions of the system switch, thus completing the charge path of the appropriate time-constant capacitor in the I.F. Module. The

Main Chassis (Continued)

contact 'a.g.c. short' is connected to the Bandwidth switch, in order to prevent the use of short time-constant a.g.c. whenever the Bandwidth switch is set to bandwidths of 1.2 kHz, or less.

Meter Switch

11. In the R. F. position the + terminal of the meter is connected to the a.g.c. output in the I. F. Module and the negative terminal to earth via the slider of the set-zero potentiometer 3RV4. With nil signal input and with the R. F. GAIN control at fully clockwise the a.g.c. line is at approximately 4 volts negative to earth, the meter can be set to zero by adjusting 3RV4. Any subsequent a.g.c. output then provides a meter indication of signal strength. In the A. F. position the meter is connected across the output of the rectifier bridge 3D1-3D4 and indicates the level of the 1 mW 600Ω audio line.

Meter Zero Setting

12. The potentiometer 3RV4 is mounted on a small component board located within the receiver, on the left hand side. (Fig. 18). The setting procedure requires the injection of a 1 μ V e.m.f. at the antenna socket, with the system switch at AGC Med. and the Meter switch to R.F. The potentiometer is then adjusted to give an 'S' Meter reading of exactly 1 microvolt.

MHz Tuning Control

13. This control rotates the variable capacitor of the lst v.f.o. thus providing the electronic band selection described in the Operators Manual.

14. A further function of the MHz control is that when set to indicate '00' a pair of microswitches (ISA and ISB on Fig. 16) are actuated, which disconnect the a.g. c. and h.t. from the r.f. unit, first mixer, second mixer, first v.f. o. and 37.5 MHz generator and transfers these voltages to alternative terminals on the rear panel marked 'h.t. l.f.' and 'a.g.c. l.f.' from which they may be connected to an l.f. adaptor unit.

KHz Tuning Control

15. This control rotates the variable capacitor of the 2nd v.f.o. and has no auxiliary functions.

R.F. Gain A.F. Gain, Line Level

16. These controls are described in the paragraphs headed 'I.F. MODULE'. The circuit connections are illustrated in Fig. 12 and Fig. 16.

Main Chassis (Continued)

Cal-Fine Tune Control 1RV3

17. Refer to the paragraphs headed '2nd V. F. O. ' and Fig. 10.

2nd V, F. O. Switch

18. Refer to the paragraph headed '2nd V.F.O.' and Fig. 10. (Page 1.14)

Plugs and Sockets

19. Several of the modules are connected to the chassis wiring via Cannon connectors which contain both d. c. and r. f.
 connections. The part of the connector attached to the module is fitted with d. c. pins and coaxial sockets and is described as the plug (PL1). The mating socket (SKT) is attached to the chassis wiring and carries d. c. sockets and coaxial pins. Diagrams of these connectors are shown in Fig. 16 Interconnections.

Wiring Identification

20. In the main chassis cableform, interconnections can be traced by reference to marker sleeves for coaxial and screened wiring, and colour coding for d. c. wiring. Refer to Fig. L-16 for information. The wire numbering for the Crystal Filter Unit is shown in Fig. 18.

R.F. MODULE

21. The R.F. Module provides filtering, pre-tuning and amplification of the r.f. signal, with delayed a.g.c. The module consists of three assemblies through which the signal passes in sequence, referring to Fig. 4 they are:-

(a)	From the Antenna input via the Muting relay
	contact RLA1 to the 0 to 30 MHz low-pass filter.

- (b) Antenna attenuation and pre-tuning stage.
- (c) The r.f. amplifier board with a.g.c.

Wiring Connections

22. Wiring connections to the module are made via pins on the side. In later versions of the module these connections are made

via a 5-way connector at the forward end, with only the r.f. outlet to the 1st Mixer remaining on the side.

R.F. Module (Continued)

Antenna Input and Filter

23.

The antenna is connected to the rear panel socket R.F. INPUT, thence via RLA1 and a 500 mA fuse to the antenna filter circuit. A discharge path is provided to protect the antenna circuit against excessive accumulation of static charges, and the 500 mA fuse is a protection against a heavy surge such as might occur if the receiver is inadvertently tuned through the carrier frequency of an adjacent high-powered transmitter. The muting relay RLA can be wired, via a pin on the 12-way outlet on the rear panel, to open the antenna circuit when an associated transmitter is keyed. The filter circuit Ll, L2, L3 and Cl to C5 has a passband of 0 to 30 MHz which is designed to prevent any break-through at the 40 MHz intermediate frequency (1st i.f.) or at the image frequency (80 MHz). The filter also prevents radiation of the first v.f.o. frequency from the antenna.

Attenuator Circuit

24. The switch SB which is operated by the front panel control marked AE ATT introduces attenuation into the signal path in approximately 10 dB steps. In the minimum attenuation position of SB the switch is fully clockwise and the signal passes from IL3 via SB2F to the switch 2SA without attenuation. As the switch is moved anti-clockwise the resistor network R4, R5 and R6 is connected, giving 10 dB attenuation. Subsequent switch settings introduce more resistance into the signal path, thereby providing attenuation levels of -20 dE, -30 dB and -40 dB. To maintain specification for measurements such as cross-modulation, intermodulation etc. at levels above 10 microvolts it is essential to use the AE ATT control.

R.F. Tuning and Range Selection

25. This circuit comprises the following: -

- (a) R. F. TUNE ganged variable capacitors 2C6a and 2C6b
- (b) The switched r.f. range filters 2Ll to 2L5
- (c) The range switch 2SA which selects either the required range filter or the wideband (WB) path.

26. Range Selection The range selection is in octave steps, 1 to 2 MHz, 2 to 4, 4 to 8, 8 to 16 and 16 to 32 MHz. Wafer 2SA4F selects the primary and 2SA1F the secondary of the filter. Wafers 2SA3B and 2SA2B short circuit all the filters except the one in use. The signal path from the r.f. tuning selection is via 2SA1F to transistor 3VT2 on the r.f.

27. R.F. Tune and W.B. The ganged variable capacitors 2C6a and 2C6b provide tuning of the selected r.f. filter (2L1 to 2L5)

under the control of the R.F. TUNE control on the front panel. In the WB setting of the Range switch the input signal by-passes the range

amplifier board.

R.F. Module (Continued)

selection and r.f. tuning circuits and is fed via a 20 dB resistive attenuating network R1, R2 and R3 into the base of transistor 3VT2. The diodes 3D11 and 3D12 provide overload protection against excessive r.f. voltages. (up to 15 volts e.m.f.)

R.F. Amplifier Board

28. The r.f. amplifier consists of two similar stages 3VT2 and 3VT3 each feeding into a low-pass filter. Considerable attention has been given to filtering and the amplifiers follow conventional practice, except for the method of applying automatic gain control which will be described in detail.

29. <u>A.G.C. Action</u> Consider the amplifier 3VT2. The gain of the amplifier can be varied according to the amount of capacitive by pass applied to the emitter resistor 3R9. The emitter by-pass is via capacitor 3C4 the junction of diodes 3D3 and 3D4 and capacitors 3C2 and 3C3 to earth.

30. The by-pass impedance is determined by the conductivity of the diodes 3D3 and 3D4 and this in turn can be controlled by the amount of d. c. bias applied by the emitter level of the control stage 3VT1. The current through 3VT1 and hence the emitter voltage, is controlled by the a.g.c. voltage applied to the base of 3VT1.

31. Under conditions of minimum a.g. c. action (maximum amplifier gain) the voltage on the a.g. c. line is -4V. This is applied to the base of 3VTl via 3R3. This causes 3VTl to conduct heavily and draw the maximum current through the emitter path formed by the chain of diodes 3Dl to 3D4. Under these conditions the impedance of the by-pass from the emitter of 3VT2 is a minimum and amplifier gain is therefore at a maximum. It is essential that the diode chain has a low forward resistance. A high resistance diode could cause a reduction in the maximum gain of the amplifier.

32. Increasing signal strength may cause the a.g. c. voltage applied to the base of 3VT1 to become less negative, as a result the current in 3VT1 decreases and the impedance of the diode path increases, thereby reducing the gain of 3VT2. Under conditions of maximum a.g. c. the level at the emitter of 3VT1 may reach 0 volts but the diodes 3D5 and 3D6 (with 3R5) will ensure that the junction of 3D5 and 3D4 remains at approximately 1.5 volts negative, thus ensuring that the diode chain 3D1 to 3D4 is completely cut off, thereby interrupting the emitter by-pass circuit of 3VT2 and reducing the amplifier gain to a minimum.

33. <u>Temperature Compensation</u> The thermistor 3THI in parallel with 3R2 provides temperature compensation. With rising temperature the current in 3VTl tends to increase. This is counteracted by a decrease in the resistance of 3THI which has the effect of taking the

R.F. Module (Continued)

bias on 3VT1 towards a more positive level, thereby checking the increase in emitter current due to temperature rise.

34.

Setting-Up 4RV1. The potentiometer 4RV1 which is mounted

on the H.T. Filter board on the underside of the receiver (Fig. 18) is provided to allow for variations in diode characteristics, thus avoiding the necessity for selection or matching of components. Adjustment should be necessary only when putting a new amplifier board into service or following component changes. The setting is quite critical and adjustment should be made strictly according to the procedure given in the alignment chapter. The general principles of the adjustment procedure are as follows:

35. The system switch is set to Manual and the R.F. GAIN control is set to the maximum gain position, thereby causing the a.g.c. line to acquire a level of -4 volts to chassis. A d.c. voltmeter (AVO 8) is connected to the collector of 3VTl (-ve) on the r.f. amplifier board.

36. Potentiometer 4RV1, should be adjusted so that the collector voltage of 3VT1 decreases (becomes less negative) as the transistor is brought towards saturation. The correct setting is the <u>exact</u> point where the collector voltage just ceases to change, indicating that the transistor has 'bottomed'. The actual reading on the voltmeter at which this occurs will probably be between 3 and 4V negative. A 'bottom' reading greater than -4 volts indicates that the diode chain has a high forward resistance which will be detrimental to amplifier gain. The forward resistance of any diode in the chain should not exceed 25Ω when measured on the 'ohms x 100' range of the AVO Model 8 test meter. (Diodes type HS1842)

FIRST MIXER

FIRST_MIXER (BA. 28211)

Fig. 8

37. A balanced mixer circuit is used to reduce the possibility of the second harmonic of a 40 MHz signal entering the 40 MHz filter. The incoming signal from the r.f. unit is fed into transformer T1 and drives the emitters of VT1 and VT2 in push-pull. The first v.f.o. frequency is applied via PL3 and C4 to the bases of VT1 and VT2. The circuit is balanced by the potentiometer RV1.

38. The mixer output appears in the inductor Ll which is mounted in a sub-assembly with the remaining inductors of the 40 MHz filter, L2 to L8. Each coil has an adjustable core which combined with the critical spacing of the coils determines the response of the filter which has a passband 650 kHz either side of 40 MHz. It is essential that the filter should have a sharp cut-off, particularly on the high frequency side, to prevent the first v. f. o. frequency entering the filter when the v. f. o. is operating at the lower end of its frequency range. FIRST V.F.O.

Fig. 5

39. The first v.f.o. consists of three sub-circuits mounted on individual boards. The oscillator is a conventional Hartley circuit tuned by the Megahertz tuning capacitor Cl which is not mounted on the board. The oscillator frequency range is 40.5 MHz to 69.5 MHz. Normally the lowest frequency used is 41.5 MHz when the Megahertz tuning control is set to 01. The oscillator output is taken from a tapping near the earthy end of inductor Ll and fed to two buffer amplifiers in parallel.

40. The two buffer amplifiers are identical circuits, the only difference being that one feeds out via plug PL2 to the first mixer stage and the other via PL1 to the harmonic mixer (37.5 MHz generator). A cascode circuit is employed to obtain adequate buffering, and the circuits are entirely conventional. When removing the 1st V. F. O. module from the chassis the h.t. lead (violet) must be disconnected from the small terminal block beside the module. Some earlier receivers do not have this terminal block, in which case the violet wire must be unsoldered from the H. T. Filter board on the underside of the receiver. (Fig. 18).

37.5 MHZ GENERATOR MODULE

NOTE: This module contains the following two main assemblies:-

- (a) The 1 MHz Oscillator, Amplifier and Calibrator
- (b) The 37.5 MHz generator assembly consisting of a harmonic generator mixer and amplifier. Refer to instructions in Chapter 7 for obtaining access to this assembly.

The 1 MHz oscillator and the calibrator assembly will be described first.

<u>1 MHZ OSCILLATOR AMPLIFIER AND CALIBRATOR</u> Fig. 6

41. This assembly consists of two sub-circuits. One board contains the 1 MHz crystal oscillator and calibrator circuit, the other board contains the amplifier stages which provide buffering in the 1 MHz input and output circuits. The boards are mounted side-by-side on the upper (hinged) deck of the 37.5 MHz Generator Module. The calibrator is easily identified by the 1 MHz crystal mounted on the board.

1 MHz Oscillator and Calibrator

42. Transistor VTl is contained in a conventional Pierce type of crystal-controlled circuit. The output is taken from the emitter via the capacitive divider C4 and C5 to provide the correct level at the base circuits of transistors VT2 and VT3 on the amplifier board. The 1 MHz crystal XLl is contained in a holder and must be removed if an external 1 MHz reference source (synthesizer) is connected to the receiver.

43. The calibrator circuit is a regenerative divider which receives an input at 1 MHz from the amplifier VT2 and supplies 100 kHz with harmonics via connector A2 to the 3rd mixer board. The divider circuit comprises transistors VT2 and VT3, and the diode mixer circuit D1 to D4.

44. The action of the divider is as follows: The 1 MHz input is

applied via R9 to the centre tap on transformer L2. The circuit L2 and C12 is designed to resonate at 100 kHz and any response at this frequency is amplified by VT3 whose collector circuit, L3 and C14, is tuned to 100 kHz. The 100 kHz in the secondary winding of L3 is fed back to the base of VT2 whose collector circuit L1, C10 is tuned to 900 kHz. Thus 900 kHz and 1 MHz are mixed in the diode ring circuit, the difference frequency (100 kHz) is transferred via T1 to the base of VT3 and the divider action becomes self sustaining. The output is taken via diodes D5, D6 and connector A2, to the third mixer module where the required range of harmonics is injected for calibration of the receiver kHz scale.

1 MHz Amplifier

45.

Transistors VT2 and VT3 on the amplifier board can be driven by the 1 MHz oscillator or they can be supplied with 1 MHz (via VT1) from an external source such as a synthesizer. The output from VT2 is fed from the junction of R9 and R10 to the connector A3, thence to the rear panel socket '1 MHz OUT'. The 1 MHz from VT2 also goes to the Detector board in the i.f. module to mix with the 600 kHz b.f.o. frequency.

46. Amplifier VT3 has a collector circuit which is modified by

R15, C12 and R17 in order to pre-shape the output waveform to obtain a suitable drive for the harmonic generator. The effect of the pre-shaping is to convert one half-cycle of the sine wave to a peaked waveform capable of generating a wide range of harmonics. (see Chapter 4. paragraph 18). Amplifier VT1 buffers the 1 MHz input whenever a synthesizer or external frequency source is connected.

47. The calibrator circuit functions only when the system switch (SE2R Fig. 16) is set to CAL or CHECK B. F. O. In other switch positions the h.t. supply to the calibrator is disconnected. When the calibrator is switched off it is essential that there shall be no leakage of 1 MHz into the third mixer stage. The leakage is prevented by diode D1 in the amplifier board which is reverse-biased in all settings of the System switch except CAL and CHECK B.F.O. In the two latter settings of the System switch the -ve voltage forward-biases Dl which allows 1 MHz to pass to the calibrator. Additional isolation is provided by diode D6 which similarly gates the calibrator output.

RA. 1217

HARMONIC GENERATOR AND MIXER (37.5 MHZ GENERATOR) Fig. 7

48. This section consists of six small sub-assemblies which are numbered 1 to 6, the same numbering also being shown on the circuit diagram. The section is contained in the same module as the 1 MHz amplifier and calibrator. The combined module is commonly referred to as the '37.5 MHz Generator'.

Harmonic Generator (Board 2)

49. The function of this stage is to produce a wide range of harmonics of 1 MHz, which it does by virtue of the pre-shaped 1 MHz input and the special characteristics of the diode Dl. The capacitive property of the diode has the effect of producing a very fast edge possessing a high harmonic content which is applied to the base of VT1. The entire range of harmonics (up to at least the 32nd) must be amplified by VT1 without discrimination in favour of any particular frequency. The pre-set capacitor C1 can be adjusted to provide a constant amplitude over the harmonic range, measured at the output (pin 4). The diode D2 is provided as a safeguard against excessive base voltages. A similar diode connection is seen in the 37.5. MHz amplifier (D1 and D2).

Harmonic Filter (Boards 3 and 4)

50. This is a low-pass filter designed to pass all harmonic frequencies between 3 and 32 MHz but with a sharp cut-off immediately above 32 MHz. The output of the filter is applied to the base of VT1 on the harmonic mixer board.

Harmonic Mixer (Board 5)

51. A balanced type of mixer circuit is used so that the harmonic spectrum and the input from the first v.f.o. will tend to cancel each other at the output. The harmonic spectrum input is supplied via pin 2 to the base of VT1. The 1st v.f.o. output is supplied via pin 4 to the base of VT2. Mixing takes place in the common collector circuit L1, C7, C8 which is tuned to 37.5 MHz. The output via pin 5 is fed to the cascode amplifier board. The circuit is balanced by potentiometer RV1 by which the bias on each transistor can be adjusted so that the emitter currents are equal. This adjustment is described in Chapter 3.

37.5 MHz Amplifier (Board 6)

52. This is a buffer stage in a cascode circuit mounted on board number 6. A cascode circuit is used to provide adequate buffering with a low noise figure. The capacitor C7 provides neutralizing feedback. The collector circuit of VT2 (T1, C4) is tuned to 37.5 MHz and the secondary of T1, is connected to socket SKT1, which mates with plug PL1, on the 37.5 MHz filter unit.

> 1 - 11 Volume 2

RA. 1217

37.5 MHz Filter Unit

53. This is a separate unit mounted on the receiver chassis which is connected to the output from the 37.5 MHz amplifier. The filter is designed and set-up to provide a passband of plus or minus
150 kHz centred on 37.5 MHz. The correct alignment of this filter is vital to the satisfactory functioning of the Wadley system. The user is advised not to attempt any adjustment of the filter alignment.

SECOND MIXER

Fig. 9

54. The second mixer produces the 2nd i. f. by mixing the 40 MHz i.f. spectrum with the 37.5 MHz injection, and selecting the 2 to 3 MHz difference frequency. The stage VT1 and VT2 is a cascode buffer amplifier similar to the buffer circuits in the 37.5 MHz generator module. The 37.5 MHz is supplied via connector A2 and pin 1; the test point TP1 is provided to check the injection level. The stage is neutralized by capacitor C19 and the collector circuit T1 and C3 is tuned to 37.5 MHz. The secondary of T1 forms part of the emitter circuit of the mixer stage VT3. The capacitor C3 is wired to the rear of the printed circuit board.

55. The 40 MHz i.f. is supplied via A3 to the base of VT3. The resistor IR4 terminates the 40 MHz filter. Test points TP3 and TP2 are provided to check the signal and injection levels, respectively. The collector circuit L1 and C8 form part of the 2-3 MHz band-pass filter and is tuned to the difference frequency, which is in fact a spectrum of signals extending from 2 MHz to 3 MHz. The output is coupled via C9 to the remainder of the 2-3 MHz band-pass filter. Signals via an 1.f. adaptor unit, which have been translated to the 2 to 3 MHz band, can be fed in via the LF socket on the receiver rear panel via connector A1 and pin 5 to the input of the 2-3 MHz band-pass filter. A panoramic adaptor can be connected to the rear panel socket PAN. This unit examines the spectrum of signals at the collector of VT2. The band-pass filter is terminated by a 39Ω resistor in the 3rd mixer stage.

THIRD MIXER

THIRD MIXER (BA. 35970)

Fig. 11

56. In the third mixer the 1 MHz spectrum of signal frequencies from the 2nd mixer (2-3 MHz) is mixed with a frequency.

(3.6 - 4.6 MHz from the second v.f.o. The difference frequency contains the required signal intelligence at a frequency of 1.6 MHz. A 100 kHz input from the calibrator which is supplied via the connector A3 provides harmonics to which the 2nd v.f.o. (kHz) tuning scale can be aligned.

Signal Input

57. The preceding 2 to 3 MHz band-pass filter (Fig. 9) is connected via socket A4 to the low-pass filter, formed by IL1, 1C1 and IC2, which has a cut-off at approximately 5 MHz. The function of the filter is to give additional protection against 6 MHz and 37.5 MHz.

58. The filtered signal spectrum is fed via 1C3 and 1L2 to pin 1 on the mixer board thence via the blocking capacitor C1 to transformer T1 of the diode mixer circuit.

Calibrator Input

59. Also applied to Tl is the harmonic input from the calibrator, via A3 and the diode 1D1. This input is switched off except when the System switch is at CAL or CHECK BFO but in order to ensure that there shall be no leakage of the fundamental 1 MHz, even though the calibrator is switched off, a suitable negative voltage is developed at the junction of R1 and R2. This reverse biases the diode in the calibrator output.

Input From Second V.F.O.

60. The variable 3.6 to 4.6 MHz from the second v.f.o. is supplied via the coaxial connector A2 to the band-pass filter formed by L7, L6, L4 and L2 and associated capacitors. Termination is provided by R7 in parallel with the primary of T2. The filter passband should extend from 3.6 MHz to 4.6 MHz with a fairly sharp cut-off above and below these limits. The filter is inserted to meet the following requirements:

- (a) The second v. f. o. is a wideband source and it is necessary to exclude the noise generated at the signal frequencies between 2 and 3 MHz, at 1.6 MHz, and also at the image frequencies of 5.2 MHz and 6.2 MHz.
- (b) If two receivers are connected in a master-slave relationship using a common 2nd v. f. o., the filter will prevent cross-talk arising from coupling between the respective 2 to 3 MHz circuits which could occur via the 2nd v. f. o.

61. The input from the second v.f.o. is applied via T2 to the bases of the balanced amplifier stage VT1 and VT2. Note the test point TP2. The collectors of VT1 and VT2 are connected via R6 and R9 to the transformers T1 and T3 respectively in the diode mixer circuit.

Mixer Circuit

62. The mixer circuit consists of the diode ring Dl to D4 together with transformers Tl and T3. This type of mixer is selected for its linearity which cancels the 'odd order' mixing which tends to occur where the input spectrum has a fairly wide bandwidth (in this instance 2-3 MHz) and the i.f. output (1.6 MHz) almost comes within the input spectrum.

63. The sum and difference frequencies from the mixer are fed via T3 into the filter formed by L3 and L5 with C3, C5 and C6. This is a wideband filter centred on 1.6 MHz. Note the test point TP1 at the input to the filter. Associated with this test point is provision for connecting a 68Ω resistor for test purposes only. The 68Ω resistor is connected when aligning the filter to ensure that the coupling factor between

64. The amplifier VT3 provides the output required to drive the i.f. unit. The 1.6 MHz output from the collector is taken via C12 and the coaxial connector A1 to the bandpass filter FL1, thence to the 1st i.f. amplifier unit. The resistor R14 matches the input impedance of the filter.

L3 and L5 is less than unity. This alignment is done at the factory and

<u>IST I. F. AMPLIFIER UNIT</u>

IST I.F. AMPLIFIER UNIT

should not normally be attempted by the user.

65. This is a small unit containing a single stage serving as a buffer between the crystal bandpass filter FLl and the subsequent crystal bandpass filter selected by the bandwidth switch. The circuit is illustrated on the right hand side of Fig. 11. The wideband tuned inductor Ll is tapped to provide separate outputs having impedances of 100Ω and 1 kΩ respectively. The 1 kΩ output feeds the subsequent filters (Fig. 16) and the 100Ω output is used for reception in the widest setting of the I.F. BW switch without further filtering. The core of Ll is adjustable via a hole in the side of the container.

SECOND V.F.O.

General

66. The second v.f.o. assembly consists of an oscillator driving two wideband buffer stages. The oscillator can be tuned over the range 3.6 to 4.6 MHz by the KH₂ tuning control of the receiver, also, fine variations of tuning can be made by the FINE TUNE control. Two outputs are provided, one to the 3rd mixer and the other for external use. (2nd V.F.O. OUT). The oscillator stage can be switched off by setting the front panel 2nd V.F.O. switch to EXT which permits the receiver kilohertz tuning to be determined externally by either a synthesizer or the 2nd v.f.o. of a master receiver.

Fig. 11

Fig. 10

Oscillator Stage

67. The oscillator VT1 works into a tuned collector circuit comprising the inductor L2, and the KHz tuning capacitor C1 with its associated preset trimming capacitor. These tuning components are mounted on a separate sub-assembly. The oscillator tuning can also be shifted by the variable capacitance diodes D1 and D2. The junction of the diodes is connected via PL1 pin 1 to the CAL-FINE TUNE potentiometer 1RV3, which provides a variable -ve voltage which can shift the 2nd v.f. o. frequency by approximately 8 kHz over the full range of the control.

68. The remainder of the oscillator circuit is conventional. The diodes D1-D4 provide d. c. stabilization against possible variations in h.t. voltage arising from extreme temperature changes affecting the power unit. The oscillator output is taken from the junction of the coupling network R7 and R8 which minimizes any loading effects on the oscillator tuned circuit.

69. The oscillator h.t. supply is connected to pin 5 via the 2nd V.F.O. switch which is mounted on the front panel of the receiver. When this switch is set to INT the 2nd v.f.o. is in operation and an output is available for external use if required. When the switch is set to EXT the oscillator h.t. supply is disconnected and the 2nd v.f.o. frequency for the receiver must be fed in from an external source, such as a synthesizer or master receiver. The external frequency is switched by the gating diodes on the buffer amplifier board.

Buffer Amplifier Board

70. The diodes D2 and D3 enable the user to switch from internal 2nd v.f.o. to an external source, and vice versa, without the changing or removal of cable connections.

71. When the oscillator stage is running, the diode D2 on the amplifier board is forward-biased by the negative supply which is connected through the 2nd V.F.O. switch (INT position) thence via pin 5 on PL1 and R2 to the diode. This allows the oscillator output to pass via D2 to the amplifier VT1. At the same time the diode D3 is reverse-biased, thus isolating the external input.

72. When the 2nd V.F.O. switch is set to EXT the negative supply is disconnected from pin 5 and connected instead to pin 4 of PLl whence it is applied via R4 as a forward bias to diode D3. The external input can now pass via A1, C2 and D3 to the amplifier VT1. The resistor R6 (75Ω) terminates the input connector. In this condition diode D2 is reverse-biased thus isolating the oscillator circuit from the amplifier board.

73. The transistors VT1, VT2 and VT3 are wideband buffer stages which provide suitable output levels from the collector circuits.
 The output from the collector of VT2 is connected via C6 and socket A2 to the 3rd mixer, the resistor R18 provides the required 1 kΩ source impedance

to the bandpass filter on the mixer board. The external output is taken via C9 from the collector of VT3, the resistor R22 providing the required 75 ohm source impedance for the connector.

74.

The B.F.O. assembly consists of the DET - B.F.O. switch

assembly and variable tuning capacitor, together with a 600 kHz oscillator board and a buffer amplifier. The b.f.o. frequency can be adjusted in fixed steps by the positions +6 to -6 of the DET-B.F.O. switch SB and varied by the fine tune capacitor C2 which is controlled by the central knob of the B.F.O. TUNE control. Switch SB has certain other functions which are described in a later paragraph.

600 kHz Oscillator Board

75. The oscillator VT1 is tuned by the inductor L1 and the variable capacitor C2. In parallel with C2 are four preset capacitors C2, C3, C5 and C6 which are connected in the +3, 0, -3 and -6 positions respectively of switch SB2F, each one providing a progressive reduction of oscillator frequency in 3 kHz steps. The oscillator is initially set-up with the DET B. F. O. switch set to position '+6' and the variable control C2 set to its mid-travel position. The core of L1 is then adjusted to provide an output at 606 kHz. In each of the subsequent switch positions (+3, 0, -3, and -6) the appropriate capacitor C2. C3, C5 or C6 is adjusted to give the required frequency. The oscillator output is taken from the junction of R1 and R2 which provides a low-level input to the buffer amplifier.

 The negative h.t. supply to the amplifier and oscillator is permanently connected, but the positive (earth) side of the supply
 to the 600 kHz oscillator stage is connected via pin 5 and the +6 to -6 positions
 of switch wafer SB1F thence via a wafer on the System switch to earth. In
 the switch positions LSB, USB and A. M. the earth is disconnected from the
 b.f.o. thus switching off the 600 kHz oscillator.

DET B. F. O. Switch SB

77. The function of each wafer will be described briefly.

78. <u>SB1F</u>: The wiper contact (tag 9) is connected to earth in all positions of the system switch <u>except</u> CAL. (Thus ensuring that the b.f.o. is switched off during the calibration procedure). The LSB and USB positions of SB1F connect the earth to the detector board (i.f. module) where it serves to connect the appropriate crystal for the s.s.b. off-set oscillator.

79. SB1R: In the A. M. position of the DET B. F. O. switch an earth is made to pin 2 of the Detector board (i.f. module) in order to connect the a.m. detector. 80. <u>SB2F</u>: Contacts 2 to 5 which correspond to switch settings +3 to -6 connect the preset capacitors C2, C3, C5 and C6 respectively across the h f o turned circuit

the b.f.o. tuned circuit.

81. <u>SB2R</u>: In every switch position except A. M. an earth is made to pin 3 of the Detector board (i.f. module) in order to connect the product detector circuit.

B.F.O. Buffer Amplifier Board

82. This amplifier is designed mainly to buffer the b.f.o. from the Detector board circuits. The output is taken from the secondary winding of T1 which provides the low impedance required by the mixer in the Detector board. The adjustable core of T1 together with C2 tunes the output, and R5 ensures sufficient bandwidth to accept the plus or minus 8 kHz variation of b.f.o. frequency. Note that the screen of the cable from T1 secondary, which feeds the 600 kHz b.f.o. frequency to the i.f. module must not be earthed, because it constitutes a part of the base-emitter circuit of transistor VT5 in the Detector board.

I.F. BANDWIDTH SWITCHING

83. Provision is made for the fitting of five crystal bandpass filters, FL1 to FL5. The actual number and specification of the filters fitted is stated in the Technical Specification at the front of this Manual. The filter switching is illustrated in Fig. 16. The widest i.f. bandwidth is determined by FL1. The standard receiver also incorporates FL3 and FL4: the additional filters FL2 and FL5 can be incorporated at customers option. The bandwidths are switched by the I.F. B.W. control which operates the switch SA1F to SA4R illustrated in Fig. 16. This switch also selects the a.g.c. time constants to suit the bandwidth selected.

Bandwidth Switch

Fig. 16

84. The input to the 1st I.F. Amplifier is filtered to the widest available bandwidth by the filter FL1. (Fig. 11). Two outputs are taken from the 1st I.F. amplifier. The 100 ohm output via 2R1 (Fig. 16) goes to switch wafer SA3F without further filtering. The 1000 ohm output goes via SA1F and SA1R to the selected filter, FL2 to FL5, then to the wafers SA3F and SA3R. The filters are terminated by resistor 2R4. The 1.6 MHz output is fed via plug 1PL3 to the socket SKT4 on the main I.F. Module (Fig. 12).

85. <u>A. G. C. Over-ride</u>. It is the function of the wafer SA2 (Bandwidth switch Fig. 16) to ensure that the short time-constant a.g. c. cannot be used whenever the I.F. B. W. control is set to select the filters FL1 or FL2. If the System switch is set to 'A.G.C. Sh' the Bandwidth switch over-rides this setting and substitutes "medium time-constant', as follows:

86. The required a.g. c. time-constants are selected by the connection of an earth to the appropriate time-constant capacitor in the

a, g. c. board of the i.f. module. This earth connection is made by contacts on the System switch wafer SEIF. The short time-constant path is connected to the System switch (tag 2) via the wafer SA2R of the Bandwidth switch. When the filters FLl or FL2 are selected, the "a.g. c. short" earth line is connected via SA2F tags 1 or 2, to the medium time-constant capacitor in the a.g. c. board. In the remaining positions of the Bandwidth switch the the 'a.g. c. short' connections are normal. To sum up; 'long' and 'medium' a.g. c. is available in all settings of the I.F. B.W. control, but a.g. c. short is confined to the settings 3 kHz and above.

NOTE: The above description refers to the standard bandwidths. Any modification to suit a customers special requirements will be described in an Appendix to the handbook.

I.F. MODULE

- 87. The I.F. Module contains the following printed circuit boards, illustrated in Fig. 12.
 - (a) I.F. Amplifier Board.
 - (b) Detector Board.
 - (c) A.G.C. Board.
 - (d) I.F. Converter Oscillator Board.
 - (e) I.F. Converter Amplifier Board.

CONNECTIONS

88. The 1.6 MHz i.f. from the bandwidth switch is fed into the i.f. module via the coaxial connector SKT4 at the forward end of the module. All other connections to the receiver are via the 37-way connector 1SKT1. Details of the 37-way connections are shown in Fig. 16. External connections are made via the coaxial sockets, SKT1, SKT2 and SKT3 and the terminal block TB1 mounted on the module and accessible at the rear of the receiver. These outlets are shown in Fig. 12.

I.F. AMPLIFIER BOARD

Fig. 12

89. The three amplifiers VT1, VT3 and VT4 employ conventional pretuned collector circuits with damping resistors R4, R16 and R20 respectively, to ensure a wide bandwidth. A moderate amount of stabilizing feedback is applied to each stage by an un-bypassed resistor in each emitter circuit. The transistor VT2 together with the diode D1 and associated components form part of the a.g.c. system and will be described in more detail.

90. The transistor VT2 is connected in the emitter circuit of VT3 and is in effect a variable feedback device which determines the gain of VT3 under the control of the a.g.c. voltage. The a.g.c. voltage is fed in at pin 3 and via R9 to the base of VT2. Consider first the state when the signal is weak. The a.g.c. line will be at its maximum negative level. This causes VT2 to saturate and offer a low impedance, thus reducing the amount of un-bypassed resistance in the emitter circuit of VT3, resulting in higher gain.

91. Increasing signal strength causes the a.g.c. level to become less negative, and VT2 conducts less heavily, thus increasing the resistance in the emitter circuit of VT3 which results in lower gain. The combined effect of D1 with R7, R8 and R9 is to modify the characteristic of VT3 so that the curve of amplifier gain plotted against change of a.g.c. volts is less abrupt, thus improving the stability of the circuit.

92. The 1.6 MHz output from VT3 is taken from two tappings on L2. One output is taken via C11 to the buffer amplifier VT4. This stage is similar to the preceding amplifiers but is provided with an additional hum-filtering capacitor C14. The other output from L2 goes to a 1.6 MHz amplifier on the A.G.C. Board. The circuit description will continue by tracing the signal path as it leaves the i.f. amplifier via C16 and pin 5 enroute to the Detector Board.

DETECTOR BOARD

93. The Detector board contains the a.m. and product detectors, also the offset crystal oscillator for s.s.b. reception. The a.m. detector operates only in the A. M. position of the DET-B. F. O. switch and the product detector operates in all the remaining positions. The switching of these circuits depends upon the biasing of various diodes.

A.M. Detector

94. In the A. M. position of the DET-B. F. O. switch +ve h.t. (earth) is applied via pin 2 and R8 to resistor R3 and diode D2. This forward biases the diode thus completing the collector circuit of VT2. At the same time the current through R3 saturates VT1 thus completing the emitter circuit of VT2, via R6, VT1 and R5. The diode D1 is reverse biased by the negative rail connection via the path R4, R11, R19 and R18. Hence VT2 acts as a conventional 1.6 MHz amplifier with the collector circuit tuned by the inductance of T1 with capacitor C1.

95. The 1.6 MHz signal appears in the secondary windings of T1. The signal across pins 1 and 6 of T1 is detected by the diode D4 the load being formed by R11, R19 and R18. The detected signal is filtered by C6, L2 and C8, and passed via C9 to the audio emitter-follower VT3. Thence via C7, pin 8 and PL1 pin 7 to the A.F. GAIN potentiometer. A detector output is taken from pin 8 on the Detector board to pin 5 of the

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terminal block TB1 at the rear of the receiver to drive a loudspeaker amplifier or tone converter etc.

96. It will be noted that the output circuit of the b.f.o. amplifier VT4 is connected to the emitter circuit of VT2. Whenever the a.m. detector is in operation, a contact on the DET-B.F.O. switch cuts off VT4 to ensure that the signal-to-noise ratio is not degraded by noise injected from this source.

Product Detector

97. The product detector utilizes the circuit of VT1 and VT2 but with certain changes achieved by diode biasing. Due to the setting of the DET-B.F.O. switch (all positions except A.M.), pin 2 is open circuited and pin 3 is connected to earth. A -ve voltage is applied via R9 and R8 which cuts off VT1 via R3, and reverse biases D2. Thus, with VT1 cut off, the emitter path of VT2 is through R5, the secondary winding of T2, R7 and R6. The earth on pin 3 supplies h.t. + through R11, R4 and D1 to complete the supply to VT2, at the same time it disables the a.m. detector by a reverse bias on D4. The output from the b.f.o. amplifier VT4 via the secondary of T2 is supplied to the emitter circuit of VT2, and the 1.6 MHz signal from the i.f. amplifier board is supplied to the base of VT2. The product detector output is taken from the junction of R4 and R11. This audio output then follows the path described for the a.m. detector.

S.S.B. Offset Oscillator

98. Transistor VT6 on the Detector board is a crystal oscillator stage which operates in the L.S.B. and U.S.B. positions of the DET-B.F.O. switch, provided the system switch is not set to CAL. (In the CAL position all b.f.o. circuits are disabled to prevent interference with the calibration procedure).

The oscillator frequency is determined by the crystal XL1 (DET-99. B.F.O. switch set to L.S.B.) or by XL2 (DET-B.F.O. switch set to U.S.B.) according to whether the oscillator is to run 1.5 kHz above or 1.5 kHz below, the 1.6 MHz i.f. Each crystal has a preset capacitor in parallel (C21 and C22) for alignment purposes.

100. When the DET-B.F.O. switch is set to U.S.B. and the System switch is set to any position except CAL, +ve h.t. is connected via plug PLI to pin 15 on the Detector Board and through R32 to the diode D7. This forward biases D7 which conducts, thereby completing the crystal circuit for VT6 which oscillates at the frequency of XL2. Crystal XL1 remains open circuited by the reverse bias on diode D6 due to the negative rail connection. In the L.S.B. position of the DET-B.F.O. switch the forward bias is removed from D7 and applied to D6 via R33, thus connecting XL1. Diode D7 reverts to the reverse biased condition. The oscillator output is taken from the emitter of VT6 and fed from the capacitive divider C18 and C19 which via R24 provides the correct impedance into the base of the mixer VT5.

101. This stage operates as an amplifier when the DET-B.F.O. switch is set to the L.S.B. and U.S.B. positions, and as a mixer in all positions from +6 to -6. The function of the mixer is to combine the 1 MHz from the receiver crystal source with the variable 600 kHz from the b.f.o.

102. <u>Mixer Action</u>. The emitter of VT5 is connected via plug PL1 to the output winding of the 600 kHz transformer in the b.f.o.

assembly. Thus, whenever the DET-B.F.O. switch is set to any of the positions from +6 to -6, the b.f.o. frequency is supplied to VT5. It should be noted that the screen of this connection is the return path of the emitter-tobase circuit of VT5 and must <u>not</u> be earthed. The base of VT5 is connected via C17 and plug PL1 to the receiver 1 MHz crystal source. The output from VT5, which is 1.6 MHz plus or minus any variation applied by the B.F.O. fine tuning control is taken via the tapping on L3 to the amplifier VT4.

103. <u>Amplifier Action</u>. When the DET-B. F. O. switch is set to L. S. B. or U. S. B. the 600 kHz b. f. o. frequency is removed and VT5 amplifies the crystal frequency 1601.50 or 1598.50 kHz fed in from VT6 and passes it via L3 to VT4.

Amplifier VT4

104. This is a conventional amplifier with stabilizing feedback provided by R16. The collector circuit is tuned by T2 and C12, and damped by R17 to ensure sufficient bandwidth. The output from T2 secondary is connected to the emitter circuit of VT2. When the DET-B.F.O. switch is set to A. M. it is necessary to cut off VT4 to prevent noise reaching the emitter circuit of VT2; this is done by disconnecting R19 from earth at the b.f.o. switch SB2R (Fig. 13)

A.G.C. BOARD

105. The a.g. c. board contains two amplifiers operating at 1.6 MHz, followed by the a.g. c. detector and d. c. amplifier feeding into the time-constant circuits. The remainder of the board is concerned with providing d. c. amplification and manual gain control facilities etc. Provision is made for connecting the a.g. c. circuit to a companion receiver in dual diversity operation.

Amplifier VT1

106. This is a 1.6 MHz amplifier tuned by L1 in the collector circuit. The input at 1.6 MHz is applied via C1 to the base of VT1 and

separate outputs are taken from tappings on L1. The output via C5 is at a level of 100 mV for external use and is connected, via pin 3 to the 1.6 MHz OUT socket at the rear panel of the receiver. The output via C4 is fed to the base of VT2 and also to the converter board via pin 4. It is reduced in level by the capacitive divider formed by C4 with C3. This arrangement is necessitated by the requirement to obtain two different levels from the one output circuit.

Amplifier VT2

107. Transistor VT2 is a 1.6 MHz amplifier with a collector circuit tuned by L2 with C9. The inductor L2 is connected as a step-up transformer so that the stage presents a high impedance to the a.g.c. detector.

A.G.C. Detector

108. The amplified 1.6 MHz from VT2 is coupled to the a.g.c. detector D1 via C10. The components R13, R14, C12 and C13 form an r.f. filter between D1 and the emitter-follower VT3.

A.G.C. Output

109. Under weak signal conditions current flows in VT3, causing a voltage drop across R15 which reverse-biases the diodes D2, and D3 via the time-constant network C14, C15, C16, R16, R17 and RV1. The a.g. c. output voltage at pins 14, 15 and 16 will be at its maximum negative level.

110. An increase in signal strength will cause the voltage at the base of VT3 to become more negative and less current will flow in R15. The diodes D2 and D3 become forward-biased, leading to a fall in current through VT4, therefore the voltage across R21 will be less negative. This causes a change of current in VT5. The output of VT5 is coupled to the emitter-follower VT6 via the System switch wafer SE1R (all settings except MAN). In the MAN setting of the System switch the a.g. c. level from VT5 is disconnected, and a variable -ve voltage, obtained via the R.F. Gain potentiometer 2RV2, is used to manually bias the a.g. c. line which controls the i.f. and r.f. amplifier stages.

111. The final a.g.c. output is taken via three parallel paths from the emitter of VT6. The output at pin 15 is used within the i.f. unit to control the 1.6 MHz amplifier. The output from pin 14 goes via PL1/1SKT1 pin 34, to the microswitch 1SA (Fig. 16) and from pin 16 the output is fed via the Meter switch (R.F. setting) to the 'S' meter on the receiver front panel.

112. The a.g. c. line voltage from an associated dual-diversity receiver or i.s.b. adaptor can be connected to the junction of diodes D2

and D3 via, pin H on the 12-way socket at the rear of the receiver, thence to pin 6 on the board.

I.F. Converter

113. Although the i.f. converter circuitry is mounted on two separate boards it should be regarded functionally as one circuit. Two

types of converter are available but only one is fitted to a particular receiver, one produces a 455 kHz output and the other a 100 kHz output. The circuit differences are confined to the oscillator crystal frequency, coils L1 and L2 on the oscillator board and the values of components R12 and C9. Resistor

R10 on the amplifier is not fitted to the 455 kHz version. Refer to NOTE 1 on Fig. 12.

114. The output of the crystal oscillator VT1 on the Converter board is fed via C3 and pin 3 to the amplifier VT1 on the amplifier board. The output from the amplifier returns via pin 4 to the emitter of the mixer stage VT2. The signal input at 1.6 MHz is applied via pin 5 to the base of VT2. The circuit L1, C8, C9, C10 and L2 provides filtering at the required i.f. Resistor 12 and capacitor C9 determine the passband, the values shown are for the 100 kHz version.

115. The mixer output is taken from a tap on L2, and via pin 7 to the amplifier board where it feeds into the base of VT2 which is part of the d.c. coupled two-stage amplifier VT2 and VT3. The direct coupled amplifier has overall feedback via R9 which gives a very low impedance at the input to V2 and contributes to the excellent thermal stability. The amplifier also has good linearity which minimises problems which might otherwise arise from intermodulation products.

116. The output circuit of VT3 (L1 and C8) is pre-tuned to the appropriate i.f. (455 kHz or 100 kHz). The resistor R10 is inserted only on the 100 kHz converter in order to off-set the higher 'Q' of L1 at the lower frequency. From a tap on L1 the output goes to the coaxial socket SKT2 on the rear panel. This socket will be engraved either 455 kHz or 100 kHz, according to the type of converter board which is fitted to the i.f. unit. It should be noted that the required termination is 50Ω for the 455 kHz and 75Ω for the 100 kHz converter.

117. If the RA. 1217 is connected to a suitable frequency synthesizer, channel oscillator or master receiver, the production of the final i.f. can be locked to the external source by a frequency injected in lieu of the crystal XL1. When an external source is connected, the crystal XL1 must be removed from its socket. For the 100 kHz i.f. a frequency of 1.7 MHz may be supplied to the 1.7 MHz IN/OUT socket.

118. The RA. 1217 may be used in a master-slave relationship with another receiver. For this purpose the crystal oscillator frequency is fed to the slave receiver via the socket 1.7 MHz IN/OUT (455 kHz IN/OUT).

С

AUDIO AMPLIFIER BOARDS

Alternatives

119. The standard RA. 1217 receiver is fitted with a 10 mW audio amplifier board, but a one-watt board is available at customers option. It should be noted that there are restrictions on the amount of -16 volt power available from the power unit for external use when the one-watt board is fitted.

10 mW Audio Amplifier Board

Fig. 14a

120. The amplifier board contains two separate amplifiers. The circuits are conventional and will therefore be described only briefly. The a.f. output at the Detector board is taken via the slider of the A.F. GAIN control to the base of the driver transistor VT1. Input capacitors C2 and C3 are in parallel so that 1.f. attenuation can be provided by the removal of C3, if desired. Negative feedback is provided by R3. The push-pullstage VT2 and VT3 operates in class A in order to eliminate the need for any setting-up adjustments; negative feedback is applied by R9 and R10. The 10 mW 600 ohm output is supplied to the PHONES jack socket on the front panel and also to the outlets at the rear panel.

121. The line amplifier VT4 is supplied with an a.f. signal via the slider of the A.F. Level pre-set control. The additional input capacitor C10 may be removed if attenuation of the a.f. input is required. The 1 mW output is taken from the secondary winding of transformer T3 to the meter diodes and to a pair of pins at the rear panel outlet.

One-Watt Amplifier Board

Fig. 14b

122. The one-watt board contains two separate amplifiers:

- (a) A one-watt push-pull section comprising transistors. VT1
 to VT5 suitable for driving an external 15 ohm loudspeaker.
- (b) A low powered section (VT6 to VT8) which provides 1 mW into the transformer 1T1 for connection to an external 600 ohm line.

123. The one-watt amplifier is a transformerless push-pull circuit employing the principle of "complementary symmetry". The audio input from the signal detector is applied via the A.F. Gain potentiometer to the base of transistor VT1, which serves as a driver to the transistors VT2 and VT3. The bases of VT2 and VT3 are directly coupled to the collector circuit of VT1 and current variations in the circuit RV2, D1, D2, R1 and R2 will vary the base potentials of VT3 and VT2.

RA.1217

124. Transistor VT2 is an n.p.n. transistor which functions as an emitter-follower for positive-going inputs, while at the same time VT3 is cut off. On a negative-going input VT2 will cut off and the p.n.p. transistor VT3 will conduct, thus giving push-pull operation. The output transistors VT5 and VT4 are d.c. coupled to VT3 and VT2 respectively. It should be noted that transistors VT4 and VT5 are mounted in holders adjacent to the board and must be carefully released when removing the board from the chassis.

125. In the static condition the junction of R7 and R8 is at a potential which is half the supply voltage (i.e. at -8V). A variation in the current in VT4 and VT5 will produce a varying voltage at this junction, which via capacitor 1C12 drives the loudspeaker. Under static conditions the amplifier is balanced by means of potentiometer RV1 which permits accurate adjustment of the half-supply voltage at the juntion of R7 and R8 (pin 5).

126. A.F. negative feedback is provided by capacitor Cl and to limit the frequency response a low value capacitor C3 is connected between collector and base of VT1. The diodes Dl and D2 provide temperature compensation to balance the effects of temperature change on the base-emitter junctions of VT2 and VT3. The potentiometer RV2 permits adjustment of the quiescent bias conditions of VT2 and VT3.

127. <u>Adjustments</u>. With no audio input connect a d. c. voltmeter to pin 5 (-ve) and chassis (+ve). Adjust RV1 for a reading of -8 volts. To adjust RV2 the -16V (violet) wire must be unsoldered from pin 6 and a milliammeter connected in series. RV2 is then adjusted for a reading of 55 milliamps. Re-connect the -16V lead and then re-check the -8 volt reading at pin 5. Re-adjust RV1 if necessary.

1 mW Amplifier

128. The 1 mW section follows conventional practice. The input from the signal detector is applied via the A.F. Level potentiometer and pin 7 to amplifier VT6 which is coupled via an emitter-follower VT7 to the output transistor VT8. The capacitor C7 is inserted to limit the frequency response.

POWER SUPPLY MODULES

129. The standard type of power module is the PU.1153. The unit consists of a main assembly to which is attached a printed circuit board containing the stabilizer circuitry. Referring to Fig. 15a, the components mounted on the assembly are prefixed '1'. Those references without a prefix are located on the printed circuit board.

INPUTS

130.	The PU.1153 can operate from the following supplies:			
	100-125 volt a.c. 45-400 Hz single phase			
	200-250 volt a.c. 45-400 Hz single phase			
	21- 27 volt d.c. with positive earth			
OUTPUTS				

131. -16 volt (regulated) 400 milliamps
-20 volt (unregulated) 200 milliamps. Generally, this level will be approximately 24 volts.

SWITCHES

AC/DC Switch

132. A locking plate desplays the type of power input, either 'A. C.' or 'D. C.' If the indication is not correct, remove the locking plate (one screw) set the switch in the alternative position and replace the locking plate in the reversed position so that the correct indication is displayed. This switch must be correctly set before switching on the receiver.

Voltage Selector Switch

133. This switch is in use when the A. C. /D. C. switch is set to A. C. It should display the correct a.c. supply voltage range, either 115 or 225. If the setting is not correct remove the locking plate, reset the switch and replace the locking plate in the reversed position so that the correct voltage range is indicated.

POWER SUPPLY MODULES (Continued)

FUSELINKS

134. Two fuse links are provided. The 250 mA MAINS fuse is in circuit whenever the A. C. /D. C. switch is set to A. C. The 1 amp H. T. fuse is in circuit on both A. C. and D. C. supplies. Replacement fuses must be of the anti-surge type.

Replacement Fuselinks

135.	MAINS	2 50 mA	Beswick	TDC 134
	Н.Т.	l amp	Beswick	TDC 134

POWER CONNECTIONS

136. The external power supply is connected via a 3-core power cable to which a suitable 3-pin connector should be fitted: for an a.c. supply connect the red lead to line; black to neutral and green to the earth (ground) pin. For a d.c. supply connect the red lead to +ve and the black lead to -ve. The free socket for connecting the power cable to the receiver plug is a Plessey MK.4 2CZ83283/5 with accessory set 508/1/03008/205.

ADJUSTMENTS

-16 Volt Output

137. The potentiometer 1RV1 on the power unit should be adjusted to provide an output of -16 volts relative to chassis. A suitable measuring point is the terminal H.T. R.F. on the rear panel of the receiver.

Dimmer Resistance: Dial Lamps

138. The setting of the dimmer potentiometer 1RV2, which is accessible at the front panel of the receiver may be set by the user to provide a suitable level of illumination.

CIRCUIT DESCRIPTION PU.1153

139. The external power supply which enters the module via the 3-pin connector is routed via the receiver ON/OFF switch and returns to the module via pins 15 and 13 of the 15-way connector 1SKT1.

140. In the A.C. setting of switch ISB, the a.c. power is fed to the primary windings of transformer 1T1. The switch ISA connects the windings in parallel for 100-125 volt input, or in series for 200-250 volt input. In the D.C. setting of switch ISB the +ve d.c. input is connected to the 0 volt output, and the -ve input to the regulator transistor and to the unregulated -20 volt output.

POWER SUPPLY MODULE (Continued)

141 The fuse IFS1 (Mains) protects the input to the transformer, the other fuse 1FS2 (H.T.) is common to both a.c. and d.c. inputs to the bridge rectifiers D1 and D2. The rectified output from D1 is fed to the stabilizer board and the output from D2, which is normally in the range 20 to 24 volts, is fed without stabilization to a pin on the 12-way outlet at the rear of the receiver.

142. The capacitors 1Cl and 1C2 serve as an r.f. filter. The resistor 1Rl is provided to discharge the capacitors when the power is switched off.

Stabilized D. C. Circuit

143. The output from the bridge rectifier Dl is fed to the filter capacitor 1C3. The transistor 1VT1 completes the connection to the -16 volt output line, and stabilizes the output by acting as a series regulator under the control of the d. c. amplifiers VT1 and VT2. The output can be set to -16 volts by adjustment of the potentiometer 1RV1.

144. The emitter of VT2 is held at a constant voltage by the zener diode D3. The level at the base of VT2 is determined by the setting of IRV1 and also by any change of voltage occurring in the chain R4, IRV1 and R5 due to a change in output load. For example, if an increase in output loading causes the voltage at IRV1 to go less negative and the collector current in VT2 decreases. This causes a change in the base current of the beta multiplier VT2 which in turn increases the output of IVT1 so as to maintain a constant output voltage.

145. To allow the stabilizer to operate from a wide range of input voltages the network R1, R2 and R3 together with the zener diode D4 provide input regulation. The diode D4 holds the junction of R1 and R2 at a constant 4.7 volts, thus clamping the base of VT1 at a constant level despite changes in input voltage. The electrolytic capacitor C1 eliminates ripple.

CHAPTER 2

TEST AND MAINTENANCE EQUIPMENT

1. The following items of test equipment are required for the procedures in Chapter 3 and 4.

R.F. Signal Generator

NOTE: A good quality signal generator, with a frequency range up to 8 MHz and a source impedance of 75 ohms is suitable for many of the tests, but to meet all requirements two signal generators of the following specification are required:

Frequency Range10 kHz to 70 MHzOutput Impedance75 ohmsA. M. Modulation distortion less than 10%Example:Marconi TF. 144H, with 50 to 75 ohms adaptor
and 20 dB pad for each instrument.

Multimeter AC/DC

Sensitivity Range Accuracy Example 20 000 ohms/volt 0 to 300 volts 2% of full scale AVO Model 8

Electronic Voltmeter (RF Voltmeter)

Input impedance	No
Range $(0 \text{ dB} = 0.775 \text{ volts})$	M
Frequency Range	Up
Example	Ai

Not less than 1 megohm Minus 50 dB to plus 10 dB Up to 70 MHz Airmec Type 301

Electronic Voltmeter (L. F.)

Input Impedance Range (0 dB = 0.775 volts) Example Not less than 1 megohm Minus 50 dB to plus 10 dB Advance Advac VM77

Digital Frequency Meter (Counter)

Example

Frequency Range Accuracy 1 Hz to 70 MHz Internal reference frequency 1 part in 10⁶ plus or minus 1 count. Racal Type SA.550 with probe for high input impedance.

Output Power Meter			
Frequency Range	100 Hz to 6000 Hz		
Input Impedance	15 ohms and 600 ohms		
Example	Marconi TF. 893A		
Noise Generator			
Output impedance	75 ohms		
Example	Marconi TF. 1106		
Oscilloscope			
Frequency Range	0 to 30 MHz with dual		
	trace.		
Example	Tektronix 545A		
-			
Waveform Analyzer			
Frequency Range	100 Hz to 10 kHz; capable		
	of measuring to 40 dB minimum.		
Example	Wayne Kerr A321		
Audio Signal Generator			
Frequency Range	100 Hz to 8000 Hz		
Example	Advance Type J2		
Example	Advance Type J2		
Telephone Headset	600 ohm im ped ance		
Tools			
Spanners: ¹ / ₂ inch AF (6 5 mm) an	d 10/32 UNE Her		

Spanners: $\frac{1}{4}$ inch AF (6.5 mm) and 10/32 UNF Hex. Screwdrivers: Various Hexagonal wrench (Allen keys): various sizes Soldering iron.

Terminating Resistors

100 o hms	¼ watt
75 ohms	¼ watt
330 ohms	$\frac{1}{4}$ watt
27 ohms	¹ ₄ watt

Heat Shunt

Required when soldering certain coil assemblies to the printed circuit board. Refer to the next page for details.

REPAIR DATA

IMPORTANT NOTICE

 When soldering certain types of coil assembly to the printed circuit board, the heat can cause serious distortion of the coil former. The types of coil most liable to this distortion are those wound on a "Neosid" former, and the following precautions must be observed with these particular coils. The procedure is recommended for all coil soldering

- (1) Remove the adjustable core from the replacement coil assembly.
- (2) Insert a brass 'dummy core' which acts as a heat shunt during the soldering operation.
- (3) When soldering is completed, remove the brass heat shunt and insert the normal core.

HEAT SHUNT

2.

The illustration below, shows the data required to make a suitable heat shunt for the above operation (4 mm core).



Material: Brass

RA. 1217/RA. 1218/RA. 217

CHAPTER 3

PERFORMANCE CHECKS

INTRODUCTION

 The instructions in this chapter provide a series of checks on receiver performance suitable for use subsequent to an overhaul, or in the event of adverse reports on receiver performance. Suitable test equipment is listed in Chapter 2. The tests should be performed in the order given. A satisfactory result must be obtained from the test being made before continuing with the next.

2. The performance requirements which are stated for each test refer to a newly-manufactured factory-aligned receiver, and should not be applied too rigidly to a receiver which has been in use for a considerable time. Furthermore, the user should assess the accuracy of his own test equipment when evaluating test results. Do not attempt to improve the receiver performance by adjusting any preset trimmer or core etc. other than in an approved test procedure.

NOTE: Signal generator levels are given as e.m.f. unless stated otherwise R.F. output voltages are r.m.s. unless otherwise stated.

CAUTION: While making test connections to the 12-way outlet on the rear panel, the receiver should be switched off to avoid an accidental short-circuit on the 1 watt output which might damage transistors.

MECHANICAL INSPECTION

- 3. (1) Check that all modules are inserted and made secure, with covers in position.
 - (2) Check all plug and socket connections.
 - (3) Check correct setting of the following switches at the rear of the receiver.
 - (a) A.C./D.C. selector switch as appropriate
 - (b) Mains Voltage Selector (225 or 115 as appropriate).
 - (4) The 2nd V.F.O. switch on the front panel should be set to INT.
 - (5) Check that fuses are of correct value and secure in their holders.
 - (6) Set the LOCK controls to OFF.
 - (7) Check all controls for smooth operation.
 - (8) Connect power supply to receiver.
POWER CHECK

- 4.
- (1) Connect the d.c. voltmeter -ve lead, (25 volt range) to the terminal H.T. R.F. on terminal block TB1 (Fig. 19).
 - (2) Remove the receiver cover.
 - (3) Set the System switch to MAN.
 - (4) Check that the voltmeter indicates -16 volts relative to chassis. If necessary, adjust potentiometer RVl on the power module. (Fig. 18).
 - (5) Adjust the DIMMER potentiometer for a suitable level of dial illumination.
 - (6) Remove the test equipment.

SYSTEM CHECK

- 5. (1) Connect high impedance headphones to PHONES socket.
 - (2) Connect the signal generator (impedance 75 ohms) to the R.F. INPUT (Antenna) socket.
 - (3) Set the signal generator to 3.5 MHz at 15 microvolts
 e.m.f. 30% modulation at 400 Hz.
 - (4) Set the receiver controls as follows:
 - (a) MHz control to 03.
 - (b) KHz control to 500.
 - (c) R.F. RANGE to WB.
 - (d) AE ATT to MIN.
 - (e) I. F. BW KHz to 3.
 - (f) R.F. GAIN fully clockwise (maximum gain).
 - (g) A. F. GAIN mid-position.
 - (h) CAL-FINE TUNE control to mid-position.
 - (j) DET-B.F.O. switch to A.M.
 - (5) Set the System switch to CAL.
 - (6) Adjust KHz tuning control around the 500 kHz indication until a beat note is heard in the phones. If necessary, calibrate the tuning scale by setting KHz control to indicate exactly 500 and adjust the CAL-FINE TUNE control to give a zero beat note.

- (7) Set the System switch to CHECK B. F. O. Alter the KHz tuning by approximately 10 kHz. Switch the B. F. O. TUNE control through positions +6 to -6. Note that correct change of beat note occurs at each change of setting.
- (8) Set System switch to MAN.
- (9) Reset KHz control to indicate 500.
- (10) If necessary, adjust signal generator frequency to the frequency of the receiver tuning.
- (11) Set the RA. 1217 meter switch to A. F. position.
- (12) Adjust the preset A. F. LEVEL control (on front panel) and note that the meter indication changes with variation in setting.
- (13) Connect the -ve lead of a d.c. voltmeter, set to the 10 volt range, to the terminal A.G.C. R.F. on the rear panel.
- (14) Set the RA. 1217 meter switch to R. F.
- (15) Switch off the signal generator.
- (16) Set the System switch to the A.G.C. positions Lg. Med and Sh. in turn. Check that -ve 4 volts is indicated on the d.c. voltmeter in each position.
- (17) Switch on the signal generator and adjust the e.m.f. to $1 \mu V$.
- (18) Set the R.F. RANGE to 2-4 and adjust R.F. TUNE for maximum signal.
- (19) Set the System switch to A. G. C. Med, then adjust the 'Meter Set Zero' preset control (see Fig. 18), so that $1 \mu V$ is indicated on the RA. 1217 meter.
- (20) Note that the d.c. voltmeter indicates approximately -ve 4 volts (not less than 3.8v) at the terminal 'A.G.C. R.F.'
- (21) Increase signal generator output in 10 dB steps up to +40 dB, check that the a.g.c. level as shown by the d.c. voltmeter becomes less negative at each 10dB step. Also check that the RA. 1217 meter indicates approximately the appropriate 'S' level at each 10 dB step. At the 12-way rear panel outlet (SKT11) briefly earth pin G (Antenna Muting) and note that the 'S' Meter indication changes from 40 dB down to less than 10 dB when the earth is applied.
- (22) Reduce the signal generator output level until the indication on the receiver front panel meter is $1 \mu V$.

- (23) Set the MHz tuning control to '00'. Note that the d.c. voltmeter a g.c. reading is removed.
- (24) Transfer the d.c. voltmeter -ve lead to the terminal H.T. R.F. Turn the MHz control clockwise. Note a reading of -16 volts. Reset the MHz control to '00' and check that the voltmeter reading is removed.
- (25) Set the MHz control to 03 and the System switch to OFF.
- (26) Transfer the -ve lead of the d.c. voltmeter to pin H(A.G.C. DIV) of the 12 way outlet SKT11.
- (27) Check that the signal generator is set to 3.5 MHz: μV e.m.f; 30% modulation: 400 Hz.
- (28) Set the System switch to A.G.C. Med. The d.c. voltmeter should indicate 9.5V plus or minus 1 volt.
- (29) Connect an l.f. electronic voltmeter to the terminal DET on TB1; a reading of 200 mV should be obtained.
- (30) Set the System switch to OFF.
- (31) Transfer the electronic voltmeter to the coaxial socket '1.6 MHz OUT'.
- (32) Transfer the signal generator output to the coaxial socket 'L. F. '
- (33) Set the signal generator to 2.5 MHz and an e.m.f. of $200 \mu V$.
- (34) Set the System switch to A.G.C. Med and check that the electronic voltmeter indicates not less than 60 mV.
- (35) Set the System switch to OFF.
- (36) Connect an electronic voltmeter to the socket 2nd V.F.O. OUT at the rear of the receiver. (unterminated)
- (37) Set the System switch to MAN.
- (38) Set the 2nd V.F.O. switch to INT. The electronic voltmeter should indicate 100 mV, approximately.
- (39) Set the 2nd V.F.O. switch to EXT. Check that the indication noted in (38) has been removed.
- (40) Reset the 2nd V.F.O. switch to INT. Remove all test equipment.

Crystal Frequency Check

- NOTE 1: Receivers supplying a 100 kHz i.f. output are fitted with a 1.7 MHz crystal; if the i.f. output is 455 kHz the corresponding crystal frequency is 1.145 MHz.
- NOTE 2: It will be necessary to remove the covers from the 37.5 MHz Generator Module and the I.F. module to make the adjustments.
- 6. (1) Connect a digital counter to the rear panel socket 1.7 MHz (or 1.145 MHz)
 - (2) Set the System switch to MAN.
 - (3) Check that the digital counter indicates 1.7 MHz (or 1.145 MHz) ±2 Hz.
 - (4) If necessary adjust the trimmer capacitor C4 in the Converter board of the I.F. Unit to achieve the 1.7 MHz indication. (See Fig. L12).
 - (5) Transfer the digital counter to the rear panel socket 1 MHz OUT.
 - (6) Adjust the trimmer C7 located on the 1 MHz board in the 37.5 MHz Generator Module (See Fig. L6, C7 is beside the crystal base).
 - (7) The digital counter should indicate 1 MHz ± 2 Hz.

Auxiliary Inputs and Outputs.

- NOTE: The receiver will have either a 1.7 MHz crystal (100 kHz i.f.) or a 1.145 MHz crystal (455 kHz i.f.).
 - (1) Terminate with 75 ohms each of the following sockets in turn and connect an electronic voltmeter to the terminated outlet. Set the System switch to MAN. The outputs should be as follows:

TABLE 1

5	Socket	Required Output	
(a)	1. MHz OUT) 50 mV in 75 ohms	3
(b)	1.7 MHz OUT or	30 mV in 75 ohme	3
	1. 145 MHz OUT	30 mV in 75 ohms	3
(c)	2nd V.F.O. OUT	50 mV in 75 ohms	3.

- (2) Set the System switch to OFF.
- (3) Disconnect the power from the receiver.
- (4) Remove the 1 MHz crystal from the 37.5 MHz Generator Module.
- (5) Set the 2nd V. F. O. INT/EXT switch to EXT.
- (6) Re-connect the power to the receiver. Set the System switch to MAN.
- (7) Connect a signal generator sequentially to the sockets listed in Table 2 below. The signal generator output to be as stated in each case. Connect an electronic voltmeter and measure the output at the sockets listed in the right hand column of Table 2. The output in each is to be not less than 50 mV into 75 ohms.

TABLE 2

Connect Sig. Gen to these Sockets	Sig. Gen. Frequency	Sig. Gen. Output (75Ω Source)	Measure Output at the Sockets
l MHz IN	l MHz	100 mV e.m.f.	l MHz OUT
2nd V.F.O. IN	4 MHz	100 mV e.m.f.	2nd V.F.O, OUT

(10) Set the System switch to OFF. Replace the 1 MHz crystal Replace the module covers.

KHz Tuning (2nd V.F.O.) Calibration Check

8.

(1) Set the kHz tuning to '000' and the System switch to CAL.

- (2) Adjust the CAL FINE TUNE control for an audio null in the phones.
- (3) Set the kHz tuning initially to 100 and carefully tune with KHz control to obtain an audio null, which should occur within plus or minus 500 Hz of the 100 kHz calibration point.
- (4) Set the kHz scale to exactly 100.
- (5) Adjust the CAL-FINE TUNE control for an audio null.
- (6) Measure the error at 000 kHz by tuning to the audio null in the vicinity of 000 kHz. The null should be within 500 Hz of the 000 kHz scale setting.
- (7) Set the KHz tuning to exactly 200 kHz and adjust the CAL-FINE TUNE control for an audio null.
- (8) Tune to the null and measure the error at the adjacent 100 kHz and 300 kHz points which should be within 500 Hz of the scale settings.

- (9) Repeat this procedure at each 100 kHz point on the kHz scale.
- (10) Calibrate the kHz tuning at the '500' setting. Then check the kHz re-setting accuracy by de-tuning and re-setting to this check point, first from a lower frequency than from a higher. The re-setting accuracy should be within plus or minus 200 Hz.
- (11) Set the kHz scale to 500. Set the CAL-FINE TUNE control full clockwise and tune the kHz control for an audio null. Note the scale reading.
- (12) Set the CAL-FINE TUNE control fully anti-clockwise and tune the KHz control for an audio null. Note the scale reading.
- (13) The difference between the two scale readings of (11) and (12) should be not less than 7 kHz. This is the range of the CAL-FINE TUNE control.

MHz Tuning (1st V. F. O.) Calibration Check

- 9.
- (1) Connect a digital counter to either of the free coaxial plugs of the 1st V.F.O. module. These are located on a bracket on the underside of the receiver. (Fig. 18).
 - (2) Set the System switch to MAN.
 - (3) Set the MHz tuning scale to indicate 01.
 - (4) The digital counter should display a frequency of 41.5 MHz plus or minus 20 kHz. If necessary make a careful adjustment of the MHz control setting to obtain this reading.
 - (5) Adjust the MHz tuning control over its full range of free movement but without changing the scale indication. The digital counter should indicate a variation of plus or minus 0. 12 MHz approximately.
 - (6) Set the MHz tuning control to 29 and adjust carefully.
 - (7) The digital counter should display 69.5 MHz plus or minus 20 kHz.
 - (8) Set the MHz tuning successively in steps of 1 MHz from
 01 to 20 on the scale. The digital counter should display
 41.5 MHz at 01 increasing by 1 MHz at each step. All

frequencies should be plus or minus 20 kHz. At each MHz point, free movement of the MHz control should produce a frequency variation of approximately plus or minus 0.12 MHz on the digital counter display.

(9) Remove all test equipment. Re-connect the 1st V.F.O. coaxial plug to its chassis socket.

B.F.O. Calibration

10.

- (1) Remove the cover from the I.F. Module.
 - (2) Connect a digital counter to test point TPl on the detector board in the i.f. module. (On the collector of VTl, Fig. 12).
 - (3) Set the System switch to CHECK B. F. O. and adjust the B. F. O. TUNE control for an audio null.
 - (4) Operate the DET-B. F.O. control through all positions. Check that the digital counter indicates the frequency in accordance with Table 3.

TABLE 3

B.F.O. CHECK

Det - B.F.O. Control Setting	Measured Frequency at TPl on Detector Board.	
L.S.B.	1 601 500 Hz ±2 Hz	
U.S.B.	1 598 500 Hz ±2 Hz	
+6	1 606 000 Hz nominal	
+3	1 603 000 Hznominal	
-3	1 597 000 Hz nominal	
- 6	1 594 000 Hz nominal	

- (5) In the '0' position of the DET-B.F.O. switch check that the B.F.O. TUNE control gives a frequency shift of not less than plus or minus 3 kHz.
- (6) Remove the test equipment. Replace the cover on the module.

CAUTION: This chapter refers to audio power meter measurements on both the 10 mW 600Ω and the 1-watt 15Ω versions of the RA.1217. The user must ensure that the correct termination is used for the particular version of the receiver under test. If in doubt, refer to the Preface at the beginning of Chapter 1 for advice on audio amplifier identification.

Overall Receiver Sensitivity Check

- NOTE: The audio power meter must be terminated in 15 ohms for a 1-watt receiver or in 600 ohms for a 10 mW receiver.
 - Terminate the power meter in 600 ohms (10 mW output) or
 15 ohms (1-watt output) and connect it to the audio output at
 pin D and pin C (screen) of the 12-way outlet 1SKT11.
 - (2) Connect the H.F. electronic voltmeter to the unterminated
 1.6 MHz OUT socket. The length of cable between socket
 and voltmeter must not exceed 12 inches. (30 cms).
 - (3) Connect the Signal generator to the R.F. INPUT (Antenna) socket. Set the generator to 3.5 MHz modulated 30% at 400 Hz. Output level 1 μ V e.m.f. from 75 Ω source.
 - (4) Set the receiver controls as follows:-
 - (a) AE ATT to MIN.
 - (b) R.F. RANGE to 2-4 MHz.
 - (c) R.F. TUNE tuned to 3.5 MHz.
 - (d) DET B.F.O. switch to A.M.
 - (e) R.F. GAIN fully clockwise (maximum gain).
 - (f) $I_{\bullet}F_{\bullet}B_{\bullet}W_{\bullet}$ switch to 3.
 - (g) System switch to MAN.
 - (5) Tune receiver MHz and KHz controls to 3.5 MHz and make fine adjustments to obtain maximum output on the electronic voltmeter. Peak the R.F. TUNE control.
 - (6) Set the System switch to A.G.C. Med.
 - (7) Observe the maximum level indicated on the electronic voltmeter which should be not less than 60 mV.
 - (8) Adjust the A.F. GAIN control for maximum output on the power meter. Note the indicated level which should be not less than 10 mW into 600Ω or 1 watt into 15Ω .
 - (9) Transfer the electronic voltmeter to the 100 kHz OUT socket on the rear panel, the output to be terminated in 75 ohms. Note the output level as follows:-
 - (a) For 100 kHz output not less than 230 mV.

(10) Set the System switch to MAN.

Single-Signal Selectivity

- 12. (1) Set the RA. 1217 controls as follows:-
 - (a) AE ATT to MIN.
 - (b) R.F. RANGE to 2-4 MHz.
 - (c) R.F. TUNE tuned to 3.5 MHz.
 - (d) I.F. B.W. switch to .2 (200 Hz).
 - (e) R.F. GAIN fully clockwise (maximum gain).
 - (f) DET-B.F.O. to A.M.
 - (g) System switch to MAN.
 - (2) Set the signal generator to 3.5 MHz, c.w. $1 \mu V e.m.f.$
 - (3) Connect the digital frequency meter directly to the signal generator output and accurately tune the generator to 3.5 MHz. Remove the digital frequency meter and connect the signal generator to the receiver R.F. INPUT (Antenna) socket.
 - (4) Connect the H.F. electronic voltmeter to the 1.6 MHz OUT socket of the receiver.
 - (5) Connect the digital frequency meter in parallel with the electronic voltmeter using a sensitive probe.
 - (6) Tune the receiver to obtain the maximum indication of the electronic voltmeter. Note the output level obtained, as a reference.
 - (7) Decrease the frequency of the signal generator until the indication of the electronic voltmeter is 3 dB down relative to the reference level noted in (6). Note the frequency on the digital frequency meter.
 - (8) Increase the frequency of the signal generator above 3.5 MHz until the electronic voltmeter again indicates 3 dB down relative to the reference level noted in (6). Note the frequency on the digital frequency meter.
 - (9) Calculate the bandwidth by subtracting the frequency noted in (7) from that noted in (8). This shall be 200 Hz plus or minus 50 Hz.

(10) Repeat the procedure outlined in operations (2) to (9) for each setting of the I.F. BW. switch, setting the signal generator frequency accordingly. The limits shall be as specified below.

I.F. BW Setting	Minus 3 dB Bandwidth
3 kHz	$3 \text{ kHz} \pm 300 \text{ Hz}$
8 kHz	$8 \text{ kHz} \pm 800 \text{ Hz}$
1.2 kHz)	$1.2 \text{ kHz} \pm 120 \text{ Hz}$
6.0 kHz) if fitted	$6.0 \text{ kHz} \pm 600 \text{ Hz}$
13 kHz)	13 kHz ±1300 Hz

(11) Disconnect the electronic voltmeter and digital frequency meter.

Signal-to-Noise Ration (C. W. and S.S.B.)

- (1) Connect the power meter (terminated in 600Ω for 10 mW or 15 Ω for 1-watt) to the audio output (ISKT11 pins D and C).
 - (2) Connect the signal generator to the R.F. INPUT (Antenna) socket. Set the output to $1 \mu V e.m.f.$ and the frequency to 3.5 MHz.
 - (3) Set the receiver controls as follows:-
 - (a) R.F. RANGE to 2-4 MHz.
 - (b) AE ATT to MIN.
 - (c) I.F. BW to 3 kHz.
 - (d) DET-B.F.O. to USB.
 - (e) R.F. GAIN fully clockwise (max. gain).
 - (f) A.F. GAIN fully clockwise (max. gain).
 - (g) Set the System switch to MAN.
 - (4) Tune the receiver (including the R.F. TUNE) to obtain maximum a.f. output on the power meter.
 - (5) Set the System switch to A.G.C. "Med" and adjust the A.F. GAIN control until the power meter indicates a reference level of 10 mW into 600Ω or 1 watt into 15Ω according to the type of audio board fitted.
 - (6) Set the System switch to MAN and adjust the R.F. GAIN control to restore the power meter reading to the reference level noted in (5).
 - (7) Set the signal generator to CARRIER OFF.
 - (8) Note the power meter reading which should be not less than 15 dB below the reference level noted in (5).

- (9) Set the R.F. RANGE switch to WB.
- (10) Set the signal generator to CARRIER ON and increase the generator output to $15 \mu V$.
- (11) Repeat the procedures (6) to (9) inclusive.

Signal-to-Noise Ratio (M.C.W.)

- 14. On completion of the C. W. Signal-to-Noise check continue as follows:-
 - (1) Set the DET-B.F.O. switch to A.M.
 - (2) Set the System switch to MAN.
 - (3) Set the signal generator output to 30% modulation at 400 Hz at an e.m.f. of 3 μ V.
 - (4) Set the R.F. RANGE to 2-4 and tune for a maximum output on the power meter.
 - (5) Set the System switch to A. G. C. "Med" and adjust the A. F. GAIN control until the power meter indicates a reference level of 10 mW into 600Ω or 1 watt into 15Ω according to the type of audio board fitted.
 - (6) Switch off the modulation at the signal generator. Note the power meter reading which should be at least 15 dB below the reference level noted in (5).
 - (7) Disconnect the power meter.

Gain/Frequency Characteristic

- 15. (1) Connect a signal generator to the R.F. INPUT (Antenna) socket. Set the signal generator to 1 MHz at 1 μV e.m.f.
 C. W. Maintain the generator e.m.f. at 1 μV throughout the tests.
 - (2) Connect the H.F. electronic voltmeter to the 1.6 MHz OUT socket using short leads.
 - (3) Set the receiver controls as follows:-
 - (a) AE ATT to MIN.
 - (b) R.F. RANGE to 1-2 MHz.
 - (c) R.F. GAIN fully clockwise (maximum gain).
 - (d) I.F. BW to 3.
 - (e) DET-B.F.O. to L.S.B.
 - (f) System switch to MAN.
 - (g) Meter switch to R.F.

- (4) Tune the receiver (including R.F. TUNE) for a maximum output on the electronic voltmeter of not less than 60 mV. Note the actual reading obtained.
- (5) Set the signal generator successively to the following frequencies (at 1 μV) and repeat operation (4). Record the readings obtained:1.5 MHz 2.0 MHz 3.0 MHz 4.0 MHz 6.0 MHz 12.0 MHz
 16.0 MHz 24.0 MHz and 29.999 MHz.
 Each output should be within a 12 dB range over the frequency band 1-30 MHz. The minimum level shall be not less than 35 mV and the maximum level not greater than 250 mV.

A.G.C. Characteristic Check

- 16. (1) Connect the signal generator to the R.F. INPUT (Antenna) socket.
 - (2) Set the signal generator to 10.5 MHz, modulation 30% at 400 Hz, e.m.f. $2 \mu V$.
 - (3) Connect the power meter (terminated in 600Ω for 10 mW or 15Ω for 1 watt) to the audio output (ISKT11 pins D and C).
 - (4) Set the receiver controls as follows:-
 - (a) R.F. RANGE to 8-16 MHz.
 - (b) AE ATT to MIN.
 - (c) I.F. BW to 3.
 - (d) DET-B.F.O. to A.M.
 - (e) R.F. GAIN fully clockwise (maximum gain).
 - (f) System switch to MAN.
 - (5) Tune the receiver (including R.F. TUNE) to 10.5 MHz and adjust for maximum output on the power meter.
 - (6) Set the System switch to AGC Med.
 - (7) Adjust the A.F. GAIN control to provide a reference reading on the power meter of either 1 mW into $600\Omega \text{ or}$ 10 mW into 15Ω .
 - (8) Increase the signal generator output to plus 85 dB relative to 2 μ V. Check that the power meter indication does not increase by more than +4 dB.
 - (9) Reset the signal generator to $2 \mu V$ and then repeat operation (8) in the AGC 'sh' and AGC 'lg' settings of the System switch.
 - (10) The power meter should remain connected for the Noise Factor check.

Noise Factor Check

17.

(1) Connect the noise generator (759 source) to the R.F. INPUT (Antenna) socket. Do not switch on the noise generator.

- (2) Connect the power meter (terminated in 600Ω for 10 mW or 15Ω for 1 watt) to the audio output (ISKT11 pins D and C).
- (3) Set the receiver controls as follows:-
 - (a) AE ATT to MIN.
 - (b) System switch to MAN.
 - (c) DET-B.F.O. to U.S.B.
 - (d) I.F. BW to 3.
 - (e) R.F. GAIN fully clockwise (maximum).
 - (f) The R.F. RANGE and R.F. TUNE should be set to the appropriate settings for peak tuning at each test.
- (4) Check that the noise generator is switched off. Set R.F. RANGE to 1-2 and tune the receiver (including R.F. TUNE) to 01.000 MHz. Carefully adjust the MHz control for maximum indication on the power meter and peak the R.F. TUNE control.
- (5) Adjust the A.F. GAIN control for 100 mW on the power meter.
- (6) Switch on the noise generator and increase its output until the power meter reading is increased by +3 dB. The increase in noise generator output (noise factor) to achieve this increase should not exceed 10 dB.
- (7) Repeat operations (4), (5) and (6) at the following frequencies:
 1.5 MHz 2.0 MHz 3.0 MHz 4.0 MHz 6.0 MHz 12.0 MHz, tuning the receiver to the appropriate range in each case.
- (8) With the R.F. RANGE switch set to '16-32' repeat the procedures of (4), (5) and (6) checking that the noise factor does not exceed 12 dB at 16.0, 24.0 and 29.999 MHz.
- (9) Disconnect the Noise Generator.

First Mixer Balance Check

18. CAUTION: Do not remove the cover from the 1st mixer and 40 MHz filter module. A suitable alignment hole is in the cover.

- (1) Connect the signal generator to the R.F. INPUT (Antenna).
- (2) Set the signal generator to 3.5 MHz, C. W., e.m.f. $15 \mu V$.

- (3) Connect the power meter (terminated in 600Ω for 10 mW or 15Ω for 1 watt) to the audio output (ISKT11 pins D and C).
- (4) Connect the electronic voltmeter to the socket 1.6 MHz OUT.
- (5) Set the receiver controls as follows:-
 - (a) AE ATT to MIN.
 - (b) R.F. RANGE to WB.
 - (c) DET-B.F.O. to U.S.B.
 - (d) I.F. BW to 3.
 - (e) R.F. GAIN fully clockwise (maximum gain).
 - (f) MHz tuning to 3 and KHz tuning to 500.
 - (g) Meter switch to A.F.
 - (h) System switch to MAN.
- (6) Tune the receiver to a maximum reading on the power meter and set the A.F. GAIN control to obtain 100 mW on the power meter as a reference level.
- (7) Set the signal generator frequency to 20 MHz.
- (8) Increase the signal generator output by plus 60 dB relative to $15 \mu V$.
- (9) Vary the signal generator tuning around the 20 MHz point until a maximum reading is obtained on the power meter.
- (10) Balance the mixer by adjusting the potentiometer RV1 on the 1st Mixer board (Fig. L8) to obtain a <u>minimum</u> output on the power meter. Check that this minimum reading is not less than 60 dB down relative to the reference level noted in (6).
- (11) Remove the signal generator.

Spurious Response to Internal Signals

- CAUTION: Spurious responses can occur if module securing screws are slack or if covers are loose.
 - (1) Ensure that all module covers are secure also the top and bottom covers of the receiver.
 - (2) Connect the power meter (terminated in 600Ω for 10 mW or 15Ω for 1 watt) to the audio output (ISKT11 pins D and C).
 - Disconnect and switch off the signal generator and terminate the R.F. INPUT (Antenna) socket with 75 ohms.

- (4) Set the Receiver controls as follows:-
 - (a) AE ATT to MIN.
 - (b) R.F. RANGE to WB.
 - (c) DET-B.F.O. to U.S.B.
 - (d) R.F. GAIN fully clockwise (Maximum).
 - (e) I_*F_* BW to 3 kHz.
 - (f) System switch to MAN.
 - (g) Set the MHz and KHz tuning to 01.000 initially.
- (5) Adjust the MHz tuning carefully for maximum noise level on the power meter.
- (6) Adjust the A.F. GAIN control to obtain a noise level on the power meter of 1 mW into 600Ω or 100 mW into 15Ω .
- (7) Turn the KHz tuning control slowly and carefully through its range from 000 to 999. When a spurious response is heard in the phones, offset the KHz tuning until the response is no longer audible and then adjust the A.F. GAIN control to restore the reference level noted in (6).
- (8) Return the Khz tuning to the spurious response and carefully tune to obtain a peak reading on the power meter.
- (9) Note the increase in the power meter reading relative to the reference level obtained in (7). This increase should not exceed 3 dB.
- (10) Repeat operations (5) to (9) at each setting of the MHz control from 02 up to 29.
- (11) Repeat operations (5) to (9), with the R.F. RANGE set to the appropriate range, and the R.F. TUNE control adjusted to maximum noise setting at each MHz setting from 02 up to 29.
- (12) On completion remove the 75Ω termination from the R.F. INPUT socket.

Spurious Response to External Signals

- 20. (1) Connect the signal generator to the R.F. INPUT (Antenna) socket.
 - (2) Set the signal generator to 3.5 MHz, C. W. $1 \mu V e.m.f.$
 - (3) Connect the power meter (terminated in 600Ω for 10 mW or 15Ω for 1 watt) to the audio output (ISKT11 pins D and C).

- (4) Set the receiver controls as follows:-
 - (a) AE ATT to MIN.
 - (b) R.F. RANGE to 2-4 MHz.
 - (c) DET-B.F.O. to L.S.B.
 - (d) R.F. GAIN fully clockwise.
 - (e) I.F. BW to 3 kHz.
 - (f) System Switch to MAN.
- (5) Tune the receiver (including R.F. TUNE) to the signal generator frequency for a maximum power output.
- (6) Adjust the A.F. GAIN control to obtain a reference reading on the power meter of either 1 mW for the 10 mW receiver or 100 mW for the 1 watt receiver.
- (7) Set the signal generator to 3.55 kHz and increase the signal generator output level to +70 dB. Make a very fine adjustment of the MHz control until a spurious frequency is heard. Lock the tuning controls at this point. (If locks are fitted).
- (8) Tune the signal generator carefully from 3 MHz up to 4 MHz. If a spurious frequency is found, adjust the signal generator output to restore the appropriate reference level noted in (6). Check that such generator level settings are not less than 60 dB above 1 μV to provide the appropriate reference level.
- (9) If a spurious response is located, which results in a signal generator level of less than 60 dB for the reference level reading on the power meter, proceed as follows:-
 - (i) Rémove the cover from the 37.5 MHz Generator module.
 - Make a small adjustment of the potentiometer RVI on the Harmonic Mixer board (Fig. L-7) to reduce the spurious response, as shown by a fall in the power meter reading.
 - NOTE: For access to the Harmonic Mixer board remove the cover from the 37.5 MHz Generator module.
 - (iii) Adjust the signal generator output level to obtain the reference level on the power meter (either 1 mW or 100 mW as appropriate). The signal generator level should be not less than 60 dB, in accordance with operation (8). If necessary make a further small adjustment of RVI and repeat the test.

- (iv) Replace the cover on the 37.5 MHz Generator module and the receiver cover.
- (v) Repeat operation (8).
- (10) Disconnect the signal generator and power meter.

CHAPTER 4

ALIGNMENT PROCEDURES

INTRODUCTION

- 1. The purpose of this chapter is to enable the modules of a receiver to be tested to a serviceable standard. The tests are designed to be as independent as is practicable, so that the checks on a particular module rely as little as possible on the correct functioning of another module, thereby providing a úseful aid to the fault location chapter. If the user wishes to check the overall performance of the receiver, reference should be made to Chapter 3.
- NOTE: Throughout this chapter the signal generator output level in each test is given as e.m.f. unless otherwise stated. All r.f. voltages are r.m.s. unless otherwise stated.
- CAUTION: Under normal conditions the receiver will maintain the factory alignment over a long period of time, consequently any other causes of trouble should be eliminated before re-alignment is undertaken. If it becomes necessary to re-align any part of the receiver, only small angular adjustments of any trimmers or tuning slugs should be necessary. When replacing access covers, module shields, etc., ensure that all screws are firmly secured to prevent any spurious signals from affecting the receiver, but do not overtighten, to the extent that screw-hole threads become damaged.

PROCEDURES

2.	Audio Amplifier Board	4 - 2
	I.F. Module (including B.F.O. check)	4 - 3
	Third Mixer	4 - 6
	Second Mixer	4 - 8
	1 MHz and 37.5 MHz Module	4 - 10
	First Mixer	4 - 13
	Filters	4 - 13
	Second VFO	4 - 14
	First VFO	4 - 14
	R.F. Module	4 - 14

Page

AUDIO AMPLIFIER BOARD (1 watt Version)

Test Equipment

3. Multimeter: 20 000 ohms/volt. (AVO 8). Soldering Iron.

Initial Control Settings

4. R.F. GAIN and A.F. GAIN fully anti-clockwise (minimum).

Setting Procedure

5.

- (1) Remove the metal panel from the underside of the receiver.
 - Set the multimeter to the 25 volt d.c. range:
 Connect the -ve lead to pin 5 on the audio amplifier board and the +ve lead to chassis.
 - (3) Set the System switch to MAN.
 - (4) Adjust the potentiometer RV1 on the audio board for a reading of -8 volts on the multimeter.
 - (5) Set the System switch to OFF and unsolder the h.t. (violet) lead from pin 6 of the board.
 - (6) Set the multimeter to the 100 mA d.c. range and connect in series with pin 6 and the disconnected h.t. lead. (+ve lead to pin 6).
 - (7) Set the System switch to MAN and adjust potentiometer RV2 on the board for a reading of 55 mA on the multimeter.
 - (8) Set the System switch to OFF.
 - (9) Remove the multimeter and re-solder the h.t. lead to pin 6.
 - (10) Repeat operations (2), (3) and (4), adjusting RV1 if necessary.
 - (11) Remove the test equipment and replace the metal panel.

AUDIO AMPLIFIER BOARD (10 mW Version)

There are no adjustments in the 10 mW audio amplifier board. Refer to Chapter 3 paragraph 11 for a check procedure. If the 10 mW level cannot be obtained in operation (8) of that paragraph it will be necessary to employ basic servicing methods to find the cause of low output.

Test Equipment

6. Signal Voltmeter.
D. C. Voltmeter.
Electronic Voltmeter.
Digital Frequency Meter.
Power Meter: (600Ω for 10 mW or 15Ω for 1 watt measurements).
0.1 μF Capacitor 30V rating.

Initial Control Settings

R.F. GAIN - Fully clockwise.
A.F. GAIN - Fully clockwise.
System switch - MAN.
DET - B.F.O. switch - A.M.
BANDWIDTH - 3 kHz.

Alignment Procedure

Fig. L-12 Fig. 12 Fig. 18

- <u>CAUTION 1</u>: Coaxial Test connections to the 1.6 MHz OUT socket at the rear panel must be short. (Not longer than 12 inches, 30 cm).
- CAUTION 2:Note the alternative terminations for the 10 mW (600Ω) and1 watt (15Ω) audio outputs. The user must determine which
one is appropriate for the particular receiver under test.
- NOTE: To make tests and adjustments within the I.F. Module the module cover and the receiver sideplate must be removed. Refer to Chapter 7 page 7-2 for removal instructions.
- 8.
- (1) Disconnect the coaxial input lead from the socket SKT4 at the forward end of the i.f. module.
- (2) Connect the electronic voltmeter to the rear panel socket
 1.6 MHz OUT using a short lead.
- (3) Connect the D.C. voltmeter (10V range) to pin 3 on the I.F. Amplifier board (-ve lead).
- (4) Adjust potentiometer RV1 on the A.G.C. board to produce a reading of -4 volts on the voltmeter.
- (5) Set the signal generator to 1.6 MHz ±10 Hz at an e.m.f. of 40 microvolts. Connect the generator to the i.f. input, SKT4 at the forward end of the I.F. Module.
- (6) Note the 1.6 MHz output level on the electronic voltmeter which should be between 80 and 150 mV. If below 80 mV, align the coils in the following order for maximum output on the electronic voltmeter.

L1 on the A.G.C. board. L2 and L1 on the I.F. Amplifier board.

I.F. Module (Continued)

- (7) Terminate the audio output (SKT11 pins D and C) in 600Ω (10 mV) or 15Ω (1 watt) and connect the power meter.
- (8) Set the signal generator to 1.6 MHz modulated 30% at 1000 Hz, and check that the audio output is not less than 10 mW (or 1 watt). (A.F. GAIN fully clockwise).
- (9) If the 10 mW (or the 1 watt) indication is not obtained, adjust the cores L3 on I.F. Amplifier board and T1 on the Detector board to obtain such an indication on the Power meter.
- (10) Switch off the modulation and set the DET B.F.O. switch to L.S.B. and U.S.B. in turn. Note that the Power meter reads at least 10 mW (or 1 watt) in each setting.
- (11) Set the DET B.F.O. switch successively to the positions
 +6 through to -6 and note that the 10 mW (or 1 watt) output is obtained in each setting.
- (12) Set the DET B.F.O. switch to A.M. and the System switch to A.G.C. 'Sh'.
- (13) Increase the signal generator output level by +36 dB. Observe the change of reading on the electronic voltmeter which should not exceed +3 dB. If the indication is satisfactory, omit the next operation.
- (14) If, in operations (13) the electronic voltmeter reading showed an increase of more than 3 dB adjust L2 in the A. G. C. board to produce a minimum level in the electronic voltmeter reading.
- (15) Re-set the signal generator output according to operation(5).
- (16) Terminate the 100 kHz OUT socket in 75 ohms. Connect the electronic voltmeter across the termination. For a 1.6 MHz 40 microvolt input to the i.f. unit the output should be not less than 230 mV into 75 ohms at 100 kHz. If the output is low adjust L1 on the converter amplifier board. If necessary adjust L2 and L1 on the converter oscillator board.
- CAUTION: Peaking of the converter board inductors can adversely affect the selectivity characteristics. Set the I.F. BW switch to the widest bandwidth and tune the signal generator through the receiver passband; note that the response is symmetrical.

I.F. Module (Continued)

- (18) Connect the digital counter to the 1.7 MHz IN/OUT socket (1.145 MHz OUT). Check the appropriate frequency. If necessary adjust C4 on the oscillator board to obtain the required frequency ±2 Hz.
- (19) Remove the counter and connect the electronic voltmeter in its place. A reading of not less than 100 mV should be obtained.

B.F.O. Check

- NOTE: Refer to Chapter 3 page 3-8 para. 10 for calibration check. If necessary align as follows:-
 - (1) Connect a digital counter to test point TP1 on the detector board in the I.F. Module via a 0.1 μ V capacitor.
 - (2) Set the System switch to MAN.
 - (3) Set the B. F. O. TUNE control to the mid-point of its movement. (Capacitor vanes at half-mesh).
 - (4) Set the B. F. O. KHz switch to L. S. B. The counter should read 1601.50 kHz ±2 Hz. If necessary adjust the Trimmer capacitor C21 on the Detector board to obtain the correct frequency.
 - (5) Set the DET B. F. O. switch to U. S. B. The counter should read 1598.50 kHz ±2 Hz. If necessary adjust Trimmer capacitor C22 on the Detector board to obtain this frequency.
 - (6) Set the DET B.F.O. switch to +6. Refer to Fig. L-13 and Fig. 18 to identify B.F.O. board. Adjust coil Ll in the B.F.O. board for a frequency of 1.606 000 Hz, plus or minus 100 Hz.
 - Set the DET B. F. O. switch and adjust the appropriate capacitor on the 600 kHz B. F. O. board, to an accuracy of ±100 Hz as follows:

B.F.O. Switch Setting	Capacitor to Adjust	Frequency (Hz)
+3	C2	1 603 000
0	C3	1 600 000
-3	C5	1 597 000
-6	C6	1 594 000

- (8) Remove all test equipment.
- (9) Leave the coaxial input lead to the i.f. unit (PL3) disconnected for the 3rd Mixer check which follows:-

3RD MIXER

Test Equipment

10. Two Signal Generators with frequency ranges up to 6 MHz and
 4 MHz respectively.
 Electronic Voltmeter.

Initial Control Settings

System Switch - MAN
 2nd V. F. O. Switch - EXT
 MHz Tuning - set to 00 MHz
 CAL-FINE TUNE Control - Mid-position of its travel

Alignment Procedure

Fig. L-11, Fig. 11

- 12. (1) Connect the signal generator to the 2nd V.F.O. IN socket on the rear panel.
 - (2) Connect the electronic voltmeter to test point TP2 on the 3rd Mixer board.
 - (3) Set the Signal Generator to 5.6 MHz and set the attenuator on the generator for a convenient reading on the electronic voltmeter (50 mV).
 - (4) Tune coil L4 for a minimum on the electronic voltmeter.
 - (5) Set the generator to 3 MHz and tune L6 for minimum on the electronic voltmeter.
 - (6) Set the generator to 3.6 MHz and tune L7 for maximum on the R. F. voltmeter.
 - (7) Set the generator to 4.6 MHz and tune L2 for maximum on the electronic voltmeter.
 - (8) Repeat the above procedures until the response is flat within 3 dB from 3.6 to 4.6 MHz.
 - (9) Transfer the signal generator to the socket L.F. on the rear panel.
 - (10) Connect the electronic voltmeter to pin 1 on the 3rd Mixer board.
 - (11) Set the generator to 3 MHz and set the generator output for a 50 mV reading on the electronic voltmeter.
 - (12) Tune coil 1Ll on the 3rd Mixer for maximum reading on the electronic voltmeter.
 - (13) Tune the generator from 2 MHz to 3 MHz and check that the output remains constant within 2 dB.

3rd Mixer (Continued)

- (14) Transfer the signal generator from the L.F. socket to pin 3 on the 2nd MIXER board. Set the generator accurately to 2.4 MHz ±10 Hz. Adjust the generator output level to a p.d. of 10 mV measured at pin 3 on the 2nd Mixer board.
- (15) Connect a second signal generator to the 2nd V. F. O. IN socket on the rear panel. Set this generator accurately to 4 MHz ±10 c/s and an e.m.f. of 100 mV.
- (16) With the input lead at the forward end of the I.F. Unit
 (PL3) disconnected, terminate the lead in 100 ohms.
 Connect the electronic voltmeter across the termination.
- (17) On the 3rd Mixer board tune coils L5 and L3 for maximum reading on the electronic voltmeter.
- (18) Check that a stage gain of not less than unity is obtained (calculated from the p.d. at pin 3 on the 2nd Mixer to the p.d. at the 100 ohm termination). Normally the gain is approximately +3 dB.
- (19) Remove the test equipment. Reset the 2nd V.F.O. switch to INT.

2ND MIXER

Test Equipment

13. Two Signal Generators each with a frequency range up to 40 MHz. Electronic Voltmeter.

Initial Control Settings

14. System switch - MAN MHz tuning - Set to indicate 00 MHz to disable 1st V.F.O. etc.

Alignment Procedure

Fig. L-9 Fig. 9

- 15.
- Remove the I MHz crystal from the I MHz module. (Fig. L-6) Connect the R.F. voltmeter to TP2 on the 2nd Mixer board. Connect the signal generator to pin 1.
 - (2) Set the signal generator to 37.5 MHz and an e.m.f. of 10 mV.
 - (3) Tune coil Tl for maximum on voltmeter. The output should be approximately 100 mV.
 - (4) Connect the electronic voltmeter to pin 3.
 - (5) Connect the signal generator to the socket L.F. on the rear panel.
 - (6) Set the signal generator to 3.66 MHz and tune coil L2 for a minimum.
 - (7) Set the signal generator to 2.6 MHz and tune L3 for a minimum. The signal generator attenuator may have to be increased as the minimum is approached to obtain a reading on the voltmeter.
 - (8) Set the signal generator to 1.3 MHz and tune L6 for a minimum.
 - (9) Set the signal generator to 4.5 MHz and tune L5 for a minimum.
 - (10) Set the signal generator to 2.5 MHz and tune coils Ll and T2 for maximum.
 - (11) Set the signal generator to 2.1 MHz and tune L4 for a maximum.
 - (12) Repeat operations (6) to (11) until a response is obtained which is flat within 3 dB from 2 to 3 MHz.
 - (13) Disconnect the signal generator.

RA. 1217

- NOTE: Two signal generators are required for the following Mixer test.
- (14) Set signal generator No. 1 to 37.5 MHz at an e.m.f. of 10 mV and connect it to pin 1 of the 2nd Mixer board.
- (15) Set signal generator No. 2 to 40 MHz and connect it to pin 6 of the 2nd Mixer board. Adjust this generator for a p.d. of 10 mV measured at pin 6.
- (16) Connect the electronic voltmeter to pin 3 (output) of the mixer board and check that a reading of 10 mV ±3 dB is obtained.
- (17) Disconnect all test equipment.

37.5 MHz GENERATOR MODULE

1 MHz AMP, OSCILLATOR AND CALIBRATOR SECTION

<u>Test Equipment</u>

I6. Signal Generator with a frequency range at least to 37.5 MHz.
 Electronic Voltmeter
 Digital Frequency Meter (Counter)
 Oscilloscope

Initial Control Settings

17. System Switch - MAN

Alignment Procedure

18.

Fig. L-6 Fig. 6

- NOTE: The 1 MHz and Calibrator Section is on the upper deck of the module. Refer to the illustration in Fig. L-7.
 - (1) Check that the 1 MHz crystal is secure in its holder on the oscillator board.
 - (2) Connect the electronic voltmeter to the rear panel socket '1 MHz OUT' and check for a reading of approximately 100 mV.
 - (3) Connect the counter to the '1 MHz OUT' socket. The frequency should be 1 MHz ±2 Hz. If necessary adjust capacitor C7 on the oscillator board to achieve the required frequency.
 - (4) Connect the oscilloscope to the junction of C12 and R17 on the amplifier board.
 - (5) Adjust the coil L1 on the amplifier board for maximum amplitude which should be approximately 17 volts. The approximate waveform is shown below

- (6) Remove the 1 MHz Crystal from its holder on the oscillator board.
- (7) Connect the signal generator to the '1 MHz IN' socket.
 Set the signal generator to 1 MHz and 100 mV e. m. f.

- (8) Transfer the electronic voltmeter to the junction of C4 and C5 on the oscillator board and check for a reading of approximately 100 mV.
- (9) Replace the crystal. Disconnect the signal generator and electronic voltmeter.
- (10) Connect the oscillator to the collector of VT2 on the oscillator board. Adjust coils L1, L2 and L3 for the correct display as shown below.

- (11) Transfer the oscilloscope to the -ve side of D6, (pin A2 in the corner of the board). The display should be 300 mV peak-to-peak.
- (12) Remove the test equipment and proceed with a check of the lower deck of the module. Switch off the receiver.

HARMONIC GENERATOR, MIXER AND 37.5 MHz AMPLIFIER

- 19.
- Refer to Fig. L-7 and to Chapter 7 for access instructions.
 - (1) Disconnect the Harmonic Filter from the Harmonic Generator by unsoldering the lead from pin 1 on the Harmonic Filter.
 - (2) Connect the electronic voltmeter to pin 2 on the Harmonic Mixer board.
 - (3) Connect a signal generator, source impedance 390 ohms to pin 1 of the Harmonic Filter. See diagram below.



- (4) Set the signal generator output to 2 volts e.m.f. Explore the passband of the Harmonic Filter from 1 MHz to 32 MHz. The ripple should not exceed 3 dB. If necessary adjust capacitors C2, C4, C6 and C8 in the Harmonic Filter for minimum ripple.
- (5) Set the signal generator to 33 MHz and check that the output is 8 dB down relative to the level at 32 MHz.
- (6) Disconnect the signal generator and electronic voltmeter. Reconnect the lead unsoldered in (1). Switch on the receiver.
- (7) Connect the oscilloscope to pin 1 of the Harmonic Generator board. The waveform and amplitude should be as measured in para. 18 operation (5) (approx. 17V p-p).
- (8) Transfer the oscilloscope to pin 2 of the Harmonic Mixer board. The amplitude should be approx. 1.2V peak-to-peak.
- (9) Transfer the electronic voltmeter to pin 5 on the Harmonic Mixer board.
- (10) Connect the signal generator to pin 4 on the Harmonic Mixer board.
- (11) Set the signal generator to 37.5 MHz at 2 mV e.m.f.
- (12) Tune coil Ll in the Mixer to a maximum reading on the electronic voltmeter which should indicate approximately 10 mV.
- (13) Transfer the electronic voltmeter to pin 4 on the 37.5 MHz Amplifier and adjust T1 on the amplifier for maximum indication. A level of approximately 100 mV should be obtained.
- (14) At the bracket on the underside of the receiver disconnect the lead which goes to the 37.5 MHz Generator module (Fig. 18). Connect the signal generator output to this lead and inject 37.5 MHz at an e.m.f. of 20 mV.
- (15) Connect the electronic voltmeter to pin 5 on the Harmonic Mixer board. Adjust RV1 on the Mixer board for a <u>minimum</u> reading on the electronic voltmeter.

1ST_MIXER_

Test Equipment

20. Two Signal Generators 3.5 MHz and 43.5 MHz Electronic Voltmeter.

Initial Control Setting

21. MHz Tuning - 03.

Alignment Procedure

22.

- Disconnect the two free coaxial leads which feed the lst Mixer from their connections to the R. F. Module and lst V. F. O. (lst V. F. O. connection is at a bracket on the underside of the receiver, see Fig. 18)
 - (2) Connect signal generator No. 1 to PL1 on the 1st Mixer lead normally fed from the R. F. Unit. Set this generator to 3.5 MHz at a p.d. of 10 mV, <u>measured at pin 2</u> on the 1st Mixer board.
 - (3) Connect signal generator No. 2 to PL3 on the 1st Mixer lead which is normally fed from the 1st V. F. O. (underside of the receiver). Set this generator to 43.5 MHz at an e. m. f. of 200 mV.
 - (4) Connect the electronic voltmeter to pin 6 on the 2nd Mixer board.
 - (5) Set the System switch to MAN and check that the electronic voltmeter reads $10 \text{ mV} \pm 3 \text{ dB}$.

FILTERS

23. No information is given on the alignment of the 40 MHz or the 37.5 MHz Bandpass Filters because it is considered that the equipment and specialized skill required for satisfactory alignment of these filters is outside the scope of the average service department. A factory aligned unit should be fitted in the unlikely event of a defect in either of these units.

<u>2ND V.F.O.</u>

Test Equipment

24. Digital Frequency Meter (Counter) Electronic Voltmeter

Procedure

NOTE: The 2nd VFO cover need not be removed.

- 25. (1) Refer to Chapter 3 page 3-6 and perform the KHz Tuning Calibration Check in paragraph 8. If the frequencies are not correct make fine adjustment of C4, accessible through a hole in the module cover.
 - (2) Connect the electronic voltmeter to the '2nd VFO OUT' socket on the rear panel across a 75 ohm termination. A level of 50 mV should be obtained.
 - (3) Transfer the electronic voltmeter to the test point TP2 in the 3rd Mixer module (Fig. 11). A level of not less than 50 mV should be obtained.

<u>Test Equipment</u>

26. Digital Frequency Meter (Counter) Electronic Voltmeter

Initial Control Settings

27. MHz Tuning - not set to '00'

Procedure

- 28.
- (1) Refer to Chapter 3 and perform the MHz Tuning Calibration Check in paragraph 9.
 - (2) The 1st V.F.O. output check is described in Chapter 5 page 5-4 operation (5).

R.F. MODULE

Test Equipment

29. Electronic Voltmeter Signal Generator (75 ohms source) DC Voltmeter Coil Trimmer Tool (supplied with the receiver)

Initial Control Settings

30. System Switch - MAN
 MHz Tuning - not on 00 MHz
 RF RANGE - Wideband (WB)
 AE ATT - Minimum (MIN)
 RF Gain Control - Maximum (fully clockwise)

Antenna Filter Alignment

31. The filter alignment is unlikely to need attention. The procedure is described in para. 36 for use if needed as part of a major overhaul.

Aerial Attenuator (AE ATT) Check Fig. 4 Fig. L-4 32. (1) Set the receiver controls according to para. 30 above. (2)Check that approximately -4 volts is present on the a.g.c. line. (measure at microswitch ISA). Disconnect the 1st Mixer lead from the outlet (SKT2) on (3) the R.F. Module. (4) Connect the electronic voltmeter across the outlet SKT2. Set the signal generator to 3.5 MHz and connect the out (5) put to the R.F. INPUT (Antenna) socket. (6) With the receiver AE ATT control set to MIN, adjust the signal generator output for a suitable dB reference on the electronic voltmeter. Note the signal generator and voltmeter levels. (7) Set the AE ATT control one step towards MAX. (8) Increase the signal generator output to restore the reference level established in (6). (9) Note the increase in signal generator output, which should be 10 dB ± 2 dB. (10)Repeat operations (6), (7) and (8). The increase in attenuation obtained at each setting of the AE ATT control should be as follows:

R.F. Module (Continued)

AE ATT Switch Setting		Change in Attenuation	
5	MIN	0 dB	(Reference)
4		-10 dB	±2 dB
3		-20 dB	±2 d B
2		-30 dB	±2 d B
1	MAX	-40 dB	±2 dB

Coil and Capacitor Assembly Alignment

- NOTE 1: The cores of coils 2L1 to 2L5 and capacitors 2C1 to 2C5 can be adjusted via holes in the receiver rear panel. Remove the plate marked "R.F. INPUT" to obtain access.
- NOTE 2: A special double-ended plastic trimming tool is supplied with the receiver for the adjustment of 2L1 to 2L5. Note that the longer and thinner end of the tool must be used for this adjustment.
- 33. (1) Remove the cover from the R.F. Module.
 - (2) Using the plastic trimming tool set all coil cores, 2L1 to 2L5, to the extreme ends of the coil formers.
 - (3) Set the trimmer capacitors 2C1 to 2C5 to the fully withdrawn position (minimum capacitance).
 - (4) Connect the signal generator (75 Ω source) to the R.F. INPUT (Antenna) socket.
 - (5) Connect an electronic voltmeter to the R.F. Module output. (SKT² on the side of the unit). Set the voltmeter initially to the 10 mV range.
 - (6) Set the R.F. TUNE control fully anti-clockwise and then turn clockwise about 45 degrees so that the knifing slot of the first segment of the vanes is aligned with the edge of the static vanes. (see illustration opposite) The dot on the R.F. TUNE control should coincide with the 1 MHz engraving on the scale.

1st Segment of Vanes in view



- (7) Set the signal generator to a frequency of 1.0 MHz at an e.m.f. of 3 mV.
- (8) Set the receiver controls as follows:

System switch - MAN R. F. RANGE - '1-2' R. F. GAIN - maximum clockwise AE ATT - MIN MHz tuning - not on '00'.

- (9) Check that approximately -4 volts is present on the a.g.c. line.
- (10) Re-fit the cover to the R.F. Module.
- (11) Insert the trimming tool into the core aperture of 2L1. Engage the secondary core which is at the end nearer to the rear of the receiver. Check that the core is fully anti-clockwise, then screw the core slowly clockwise for a maximum indication on the electronic voltmeter. The core must be set to the first tuning point obtained. Adjust the voltmeter range as necessary.
- (12) Push the trimming tool right through to the further end to engage the primary core. Check that the core is fully clockwise, then slowly screw the core inwards (anti-clockwise) to obtain a maximum voltmeter reading at the first tuning point. The primary tuning is fairly insensitive and close attention is required to observe the resonant point.
- (13) Set the signal generator to 2.0 MHz.
- Adjust the R.F. TUNE control for a maximum indication on the electronic voltmeter and adjust the trimmer capacitor
 2C1 for a maximum voltmeter reading. Note this maximum reading as a reference.

- (15) Tune the signal generator as follows and note the output readings obtained.
 - (a) 1.8 MHz
 - (b) 2.2 MHz

The voltage ratio at resonance (operation 14) relative to the off-tune readings in (a) or (b) should be not less than 6 to 1.

- (16) Set the signal generator to 1.5 MHz.
- (17) Adjust the R.F. TUNE control for a maximum indication on the electronic voltmeter and note this reading as a reference. Check that the R.F. TUNE control cursor lies within the scale calibration marks.
- (18) Set the signal generator as follows and note the output readings obtained.
 - (a) 1.65 MHz
 - (b) 1.35 MHz

The voltage ratio at resonance (operation 17) relative to the off-tune readings in (a) and (b) should be not less than 6 to 1.

- (19) Set the signal generator to 1.0 MHz and repeat the procedure of (17).
- (20) Repeat the procedure of (18) at signal generator frequencies of 900 kHz and 1.1 MHz.
- (21) Align the remaining circuits, that is the 2-4 MHz, 4-8 MHz, 8-16 MHz and 16-30 MHz ranges, as described in operations (6) to (20) but using the frequencies specified in the following alignmenttable. Set the R. F. RANGE control as stated in the left hand column. The right hand column is available for the tester to insert the readings obtained. The voltage ratio of the resonance reading relative to the 'off-tune' readings must in all cases be not less than 6:1

Refer to the alignment table on the next page

ALIGNMENT TABLE

Coil and Capacitor Assembly

R.F. Range Setting	Signal Generator Setting	Adjust	Output Readings Obtained
2 - 4 MHz	2.0 MHz	2L2	
11	4.0 MHz	2 C2	
11	3.6 MHz and 4.4 MHz		
п	3.0 MHz		
11	2.7 MHz and 3.3 MHz		
11	2.0 MHz		
19	1.8 MHz and 2.2 MHz		
4 - 8 MHz	4.0 MHz	2 L3	
н	8.0 MHz	2C3	
11	7.2 MHz and 8.8 MHz		
11	6.0 MHz		
11	5.4 MHz and 6.6 MHz		
11	4.0 MHz		
11	3.6 MHz and 4.4 MHz		
8 - 16 MHz	8.0 MHz	2L4	
11	16.0 MHz	2C4	
11	14.4 MHz and 17.6 MHz		
11	12.0 MHz		
11	10.8 MHz and 13.2 MHz		
11	8.0 MHz		
11	7.2 MHz and 8.8 MHz		
16 - 30 MHz	16.0 MHz	2L5	
11	30.0 MHz	2C5	
11	27.0 MHz and 33.0 MHz		
11	24.0 MHz		
11	21. 6 MHz and 26. 4 MHz		
11	16.0 MHz		
11	14.4 MHz and 17.6 MHz		

R.F. Amplifier Alignment

- Refer to paragraph 32 and set up the receiver and test equipment according to operations (1) to (4). The signal generator level should be set to 3 mV.
 - (2) Sweep the signal generator across the 1 to 30 MHz passband, note the electronic voltmeter readings and check that the response does not vary by more than 6 dB. If necessary adjust coils 3L1, 3L2, 3L3 and 3L4 on the amplifier board for minimum undulation in the passband.
35.

36.

- The signal generator and electronic voltmeter should be connected as in paragraph 32. Set the signal generator to 5.5 MHz, c.w., 10 mV e.m.f.
 - (2) Check that the System switch is at A.G.C. Med.
 - (3) Connect a d.c. voltmeter -ve lead to the collector of 3VT1 (Fig. L-4).
 - (4) Adjust potentiometer 4RV1 (which is on the Filter component board mounted on the underside of the receiver) until the collector voltage of 3VT1 just 'bottoms'. A reference to the electronic voltmeter reading should show that the bottoming point of 3VT1 coincides with maximum r.f. gain. Refer to Chapter I paragraph 34 for a description of the procedure.
 - (5) Disconnect the signal generator and electronic voltmeter. Reconnect the 1st Mixer lead to the R.F. Module.

Antenna (0-30) MHz Filter Alignment

- NOTE: This alignment should not normally be required and should be considered only as part of a major overhaul.
 - (1) Connect the electronic voltmeter to the output lead of the 0-30 MHz filter.
 - (2) Connect the signal generator $(75\Omega \text{ source})$ to the R. F. INPUT (Antenna) socket. Set the generator to 40 MHz and an e.m.f. of 1 volt.
 - (3) Adjust the coil ILl for a minimum reading on the electronic voltmeter.
 - (4) Set the signal generator to 56 MHz and adjust coil IL2 for a minimum reading on the electronic voltmeter.
 - (5) Set the signal generator to 43 MHz and adjust coil IL3 for a minimum reading on the electronic voltmeter.
 - (6) Sweep the signal generator frequency from 1 to 30 MHz. Observe the electronic voltmeter readings and check that the response does not vary by more than $l\frac{1}{2}$ dB.
 - (7) Remove the test equipment.

CHAPTER 5

FAULT LOCATION

INTRODUCTION

1. This chapter provides fault location advice at two levels. Paragraphs 2 to 11 assume that the only test equipment available is a universal test meter (Multimeter). The object being to locate an elementary fault. Paragraphs 12 to 16 assume the use of additional test equipment, and will direct the user to the appropriate paragraphs in Chapter 4 (Alignment).

PRELIMINARY CHECKS

- 2. If the receiver is newly installed check the following items:
 - (a) AC/DC selector switch correctly set. (rear panel).
 - (b) 2nd V. F. O. switch set to INT (front panel).
 - (c) MHz tuning control not set to '00'.
 - (d) Power connection: Note that the polarity of a d.c. supply must be correct. (Refer to Chapter 2).
 - (e) Antenna connected.

INITIAL FAULT LOCATION PROCEDURE

<u>Controls</u>

- 3. Set the receiver as follows and check for signal or noise.
 - (1) System switch to MAN.
 - (2) DET B.F.O. switch to A. M.
 - (3) Meter switch to R. F.
 - (4) A. F. GAIN to maximum (clockwise)
 - (5) R. F. GAIN to maximum (clockwise)

- (6) I.F. BW Switch to 3 kHz
- (7) AE ATT control to the MINIMUM position (MIN)
- (8) R. F. RANGE switch to "WB"
- (9) Ensure that the MHz tuning control is not set to '00'.

Basic Diagnosis

4. The most useful indication in elementary fault diagnosis is receiver noise, or 'mush'. The controls should be set as listed in paragraph 3 and the receiver tuned over a suitable portion of the h. f. band. At each step of the MHz tuning control make a fine adjustment and listen for a rise in receiver noise level. If no noise can be heard, check that the phones are serviceable and, if possible, listen at an alternative audio outlet as well as at the phones jack socket. Note the receiver meter reading. If a reading is obtained, reduce the R. F. GAIN and the meter should deflect to the right. This indicates that the h.t. supply and a.g. c. line are normal.

POWER CHECK

5. If the receiver appears dead (no noise or meter reading) and the dial lights are not illuminated, check the H.T. fuse and then measure the -16 volts h.t. at one of the following points:

- (a) On the terminal block TB3 at the pin adjacent to the Fine Tune potentiometer (Fig. 18).
- (b) At the centre tag of the 2nd V. F. O. EXT/INT switch.
- (c) At the microswitch 1 SB. The centre tag is the h.t. feed to the receiver front end. The reading should be present in all settings of the MHz control except '00'. When '00' is selected the reading appears on the top tag. It should be present on the bottom tag in all modes of operation.
- (d) The rear panel terminal H.T. R.F. should give a -16 volt reading except when the MHz control is set to '00'.

GENERAL CHECK

(2) Set the System switch to CHECK B. F. O. and rotate the DET - B. F. O. MODE switch through all settings. Select +6 to -6 and rotate the B. F. O. variable control. Listen for the b. f. o. heterodyne whistle and observe the receiver meter indications. These indications can be used for diagnosis as indicated in the following Table.

TABLE 1

"CHECK B. F. O. " INDICATIONS

Meter Indication Observed		B.F.O. Whistle Audible	Diagnosis			
(a)	Yes	Yes	Receiver is serviceable from 3rd Mixer input through to audio output. Also 1 MHz is functioning.			
(b)	Yes	No	Fault probably in detector board of I.F. Module (Fig. 12) or audio amplifier board or connections. (Fig. 14).			
(c)	No	No	Possi	ble areas of fault are:-		
			(a)	1 MHz reference (Fig. 6)		
			(b)	lst i.f. amplifier unit (Fig. 11)		
			(c)	Main i.f. amplifier board of I.F. Module (Fig. 12).		
			(d)	Transistor stage VT3 in 3rd Mixer (Fig. 11).		

7. If both whistle and meter reading are obtained in CHECK B. F. O. set the System switch to CAL. If no calibration whistles are obtained as the kHz tuning is rotated, it suggests a fault in the 2nd V. F. O. or the mixer stage of the 3rd Mixer. 8. If the CHECK B.F.O. and CAL tests are satisfactory, set the controls as listed in paragraph 3 and make a front end check as follower

follows:-

FRONT END CHECK

9. Set the DET - B. F. O. switch to A. M. Listen intently and slowly rotate the MHz tuning control. If a very slight rise in noise level can be heard as the MHz tuning passes through each resonant point it suggests that the 37.5 MHz loop is functioning and therefore the fault is more likely to be in the antenna circuit, R. F. Module or 1st Mixer. Make the check in the WB setting of the R. F. RANGE switch as well as in the tuned antenna condition (adjust R. F. TUNE control). Thoroughly check all front end connections as follows:

- (1) Check Antenna
- (2) Check continuity through the 500 mA fuselink in the R. F. Module.
- (3) Ensure that the muting relay is not energized. Touching an earth to pin G of the 12-way socket on the rear panel should cause the relay to operate. At the same time listen for any change of noise level.
- (4) Check the connections between the R.F. Module and 1st Mixer (on the side of the R.F. Module).

If, when tuning the MHz control as described in the previous paragraph, no noise can be heard, the 1st V.F.O. or its connections may be faulty. Check connections from 1st V.F.O. to 1st Mixer and 37.5 MHz Generator respectively at the bracket on the underside of the main chassis. Ensure that they are not reversed.

R.F. H.T. Check

11. On the terminal block TB1 at the rear of the receiver connect the terminal H. T. R. F. to the terminal H. T. L. F. If the receiver then functions correctly the microswitch ISB (Fig. 16, Fig. 18) should be checked. Make a voltage check at the microswitch (centre tag).

1 MHz Check

Set the System switch to CAL and tune the KHz control to the 100 kHz calibration check frequencies. If the calibration whistle is heard at each point it indicates that the 1 MHz reference is functioning. If no calibration whistles are heard, turn the System switch to CHECK B. F. O. and set the DET - B. F. O. switch to +6, +3, -3 and -6 kHz in turn. If, again.

no heterodyne whistles are heard, it indicates a faulty 1 MHz crystal oscillator. Check that the crystal is correctly fitted in the upper deck of the 37.5 MHz Generator Module. If test equipment is available check with an electronic voltmeter for an e.m.f. of 100 mV at the rear panel socket "1 MHz OUT".

A.G.C. FAULT

- If the receiver operates satisfactorily with manual r.f. gain control (System switch to MAN) but overloads on strong signals in the a.g. c. settings of the System Switch check as follows:
 - (1) Tune the receiver to a strong signal. Set the System switch to A.G.C. Med and the Meter switch to R.F. If the meter indicates a reading appreciable greater than l microvolt the a.g.c. board in the I.F. Module is serviceable. If no reading is obtained the fault is probably in the I.F. Module.
 - (2) If the meter reading is satisfactory, connect the test meter negative lead to the terminal A.G.C. R.F. on the rear panel (positive lead to chassis). As the receiver is tuned through a powerful signal the a.g.c. level should change from -4V (no signal) to approximately 0 volts (strong signal). If no reading is obtained check the microswitch ISA adjacent to the MHz tuning shaft. (Fig. 16 Fig. 18). The bottom tag of the microswitch should show the a.g.c. level in all operating modes; the centre tag in all settings of the MHz control except '00' and the top tag only in the '00' setting of the MHz tuning control.
 - NOTE: The levels quoted in para. 13 cannot be given exactly because the level of a strong signal is not defined.

DETAILED CHECK

- 14. (1) Connect a signal generator, 75 ohm source to the antenna socket and set to 3.6 MHz at 10 mV p. d. (20 mV e. m. f.)
 (2) Disconnect the coaxial lead from the front end of the I.F. Module. Terminate the free lead in 100 ohms.
 (3) Connect the electronic voltmeter across the 100 ohm termination.
 (4) Set R.F. RANGE to WB
 - AE ATT to MIN System Switch to MAN R.F. GAIN fully clockwise.

(5) Carefully tune the MHz and kHz tuning controls to 03.500 At the resonant point the electronic voltmeter reading should rise to 100 mV approx. If no reading, or very low reading is obtained proceed as described in next paragraph.

Fault Prior to I.F. Unit

- 15. (1) Refer to para. 12 and check that 1 MHz oscillator is functioning. If no whistles are heard check the 1 MHz module as described in Chapter 4 para. 18.
 - (2) Move the 2nd V.F.O. switch between the INT and EXT positions and listen for a change of noise level. If noise level does not change, the 3rd Mixer may be faulty. Make sure that the 2nd V.F.O. switch is returned to the INT position. Refer to Chapter 4 for further tests.
 - (3) If the 3rd Mixer check is satisfactory, but a fine adjustment of the MHz tuning control fails to produce any rise in noise level, proceed as follows:
 - (4) On the 2nd Mixer connect an electronic voltmeter to test point TP2. Rotate the MHz tuning control slowly, and note the voltmeter reading as a frequency is selected. The reading should rise to approximately 80 mV at each resonant point.
 - (5) If approximately 80 mV is not obtained in (4) check the lst V.F.O. outputs. Disconnect the free coaxial leads from their respective bulkhead sockets on the (underside of the main chassis. (Fig. 18). Terminate the lead which feeds the lst Mixer in 47Ω and check for approx. 100 mV with the electronic voltmeter. Terminate the lead which feeds the 37.5 MHz Generator in 27Ω , and check for approx. 100 mV with the electronic voltmeter.
 - (6) If the 1st V. F. O. outputs are satisfactory the 37.5 MHz Generator module must be checked in accordance with Chapter 4 paragraph 19.
 - If the 37.5 MHz checks are satisfactory check the R.F.
 Module as follows:

- (8) Disconnect the r.f. output lead from the side of the module (SKT2) and connect an electronic Voltmeter to the socket.
- (9) Connect a signal generator (75Ω source) to the R.F. INPUT (antenna) socket. Set the signal generator to the levels indicated in Table 2, according to whether the receiver is in the R.F. TUNE or the WB mode, and check the electronic voltmeter readings. The System switch should be at MAN, AE ATT to MIN and RF GAIN to maximum.

TABLE 2

	Output from R.F. Module					
Signal Generator Level (e.m.f.)	In Tuned Mode	In WB Mode				
3 mV 10 mV	18 mV	- 10 mV				

- (10) A low output from the R.F. Module may indicate an a.g.c. fault. With the System switch to MAN, and the R.F. GAIN to maximum, the d.c. reading on the a.g.c. line should be -4 volts. A suitable measuring point is at the rear panel terminal 'a.g.c., r.f.', or if the cover is removed at the centre contact of the microswitch ISA (Fig.18). If this a.g.c. level is not obtained a fault in the i.f. unit is probable. If the -4V level is satisfactory the r.f. amplifier gain may be tested with the a.g.c. removed, as follows:-
- (11) On the r.f. amplifier board connect two 0.1 μ F capacitors between the -16V line and the junction of diodes 3D1 and 3D2, and 3D3 and 3D4, respectively, thereby removing the gain control from 3VT2 and 3VT3. Note the resulting increase in output. If the increase in gain is significantly greater than 4 dB the a.g.c. circuit should be investigated. The forward resistance of the diodes 3D1 to 3D4 and 3D7 to 3D10 should be checked. When measured on the 'ohms x 100' range of the AVO8 testmeter the forward resistance of any one of the diodes should not exceed 45 Ω .

Fault in the I.F. Unit

16. A systematic check on the I.F. Unit is described in Chapter 4.

CHAPTER 6

ROUTINE MAINTENANCE

GENERAL

 The RA. 1217 receiver should require no mechanical maintenance until a considerable period of service has elapsed, provided that the receiver has been treated with reasonable care. It is important that modules should be carefully but firmly replaced following removal and that all covers should be clean and secure, to maintain the high standard of screening which is necessary. Make sure that no cover-screws have suffered stripped threads due to overtightening. Whenever a module has been removed ensure that on replacement it is clean and free from grease or corrosion to ensure good electrical bonding. Ensure that retaining screws are made tight.

LUBRICATION

2. No lubrication is needed for at least the first year of service. Fast moving shafts are carried in sealed races which require no lubrication. Certain slow moving surfaces are carried in 'oilite' bearings which after appreciable service may each be given a single drop of thin molybdenised oil at intervals of approximately six months. A drop or two of oil may be given to the stop-collars of the MHz and KHz tuning system. Remove any surplus lubricant to prevent the accumulation of dirt.

<u>CHAPTER ?</u>

DISMANTLING AND RE-ASSEMBLY

WARNING: BEFORE DISMANTLING ANY PART OF THE RECEIVER DISCONNECT THE MAIN POWER SUPPLY AT ITS SOURCE.

INTRODUCTION

 This chapter describes how to remove the modular units of the receiver. Detailed re-assembly instructions are not given in those cases where it is feasible to interpret the dismantling procedures in the reverse order. Detailed instructions are given however, for re-assembly of the R. F. Unit, 1st V. F. O. and 2nd V. F. O. due to the requirement for accurate setting of variable controls in these modules. Reference to figure 18 will assist identification of the required module.

- CAUTION: (1) When refitting a module which requires the mating of a fixed plug and socket. Take care that the plug and socket are correctly aligned before applying pressure to the module. Otherwise pin damage may result.
 - (2) It is important that all module retaining screws should be firmly tightened to ensure good electrical bonding throughout the receiver. Loose screws can cause spurious frequencies to develop. When tightening screws, do not apply such force that screw threads might become stripped.

RECEIVER TOP COVER

2. The top cover of the rack mounted receiver is retained by 15 screws. Note the securing screw near to the centre which locates in the 1 MHz unit. All the cover screws must be tight during normal operation to ensure satisfactory screening. This cover must be removed as a first step in any dismantling operation except for removal of the power unit.

UNDERSIDE SCREWS

3. Those screws on the underside of the receiver which secure the various modules are marked with a circle of paint.

RECEIVER BOTTOM COVER

4. A small cover on the underside of the receiver is held by 8 screws. This cover may be removed for access to the audio board, the H.T. Filter board and the connectors of the 2nd and 3rd Mixer modules.

RA. 1217

FRONT PANEL

- 5. The removal of the panel is normally of no benefit in servicing, but it is necessary if the B.F.O. assembly is to be removed.
- 6. Tools: Hexagonal keys Screwdriver Spanner.

Front Panel Removal

7. (1) Using a suitable hexagonal key slacken the grub-screws and remove all control knobs.

- (2) The panel is held by four chromium plated screws. Remove these nuts and screws.
- (3) Remove the two chromium screws in the centre of the panel adjacent to the meter.
- (4) The panel can now be withdrawn.

POWER UNIT

- 8. <u>Removal</u>
 - (1) Release the four captive retaining screws from the unit on the rear panel and ease the unit gently towards the rear.

9. <u>Re-assembly</u>

 Insert the unit squarely and apply gentle pressure, evenly, to ensure correct mating of the plug and socket, and replace the four retaining screws.

I.F. MODULE

10. Tools: Medium and large screw drivers: Spanner . 25 inch A. F.

11. <u>Removal</u>

- (1) Remove the chassis side plate (6 screws).
- (2) Using the spanner; undo the external coaxial connector from the socket at the forward end of the module.
- (3) Remove two screws from the underside.
- (4) Remove two screws from the right-hand side (viewed from the rear) of the rear connector panel.
- (5) Withdraw the module towards the rear, releasing the 37-way connector at the front end while doing so.

R.F. MODULE

- 12. Tools: Hexagonal Key Screwdrivers Spanner . 25 inch A. F.
- 13. <u>Removal</u>
 - (1) Remove the chassis side plate (6 screws)
 - (2) At the rear of the front panel release the grub screws in the boss of the R. F. Range calibrated scale.
 - (3) Slacken the grub screws in the control knob and remove.
 - (4) Within the receiver behind the front panel, on the side, remove the two screws which hold the shaft bearing bracket.
 - (5) Slacken the grub screws in the control shaft flexible coupler. Pull the control shaft forward a short distance out of the coupler.
 - (6) Release the five retaining screws on the underside of the chassis.
 - (7) Remove the coaxial connector from the side of the module adjacent to the 1st Mixer.
 - (8) Release the connector plug from the forward end. Earlier modules have soldered connections on the side; these must be unsoldered after noting the pins and colour coding of the wiring.
 - (9) Draw the module forward a short distance and then lift upwards.

R. F. Module Refitting

14.

- (1) Remove the cover from the R. F. Module.
 - (2) Place the module approximately into its position in the chassis and solder the connections to the pins on the side. (In latter versions of the module these soldered connections are not fitted, connections are made by a plug and socket at the front end).
 - (3) Connect the coaxial lead from the 1st Mixer to the socket on the side of the R. F. Module.
 - (4) Place the module in position and screw-up the five retaining screws on the receiver underside.

- (5) Insert the connector plug at the forward end (only on later versions of the module).
- (6) Turn the Range switch shaft by hand to its fully anti-clockwise setting (viewed from the front).
- Insert the control shaft into the coupler. (Ensure that the R. F. Range calibrated scale is mounted on the shaft).
- (8) Tighten the grub screws in the shaft coupler.
- (9) Set the R. F. Range scale so that 'WB' is displayed in the window and then tighten the grub screws in the boss of the Range scale.
- (10) Put the R. F. Range control knob on to the shaft and tighten the grub screws. Rotate the control and check that the Range figures are correctly positioned in the window.
- (11) In the R.F. Module set the variable capacitor so that one segment of vanes is clear (see illustration in Chapter 4).
- (12) Put the knob on the control shaft so that the white spot is in line with the panel markings 1 to 16 and tighten the grub screws.
- (13) Replace the module cover and chassis side plate.

37.5 MHz GENERATOR

1 MHz OSCILLATOR AND CALIBRATOR

15. These two units are contained in separate sections mounted in a single module. Following removal of the module the two units can be separated, if necessary, but this involves delicate work with a lowwattage soldering iron.

Tools: Screwdrivers.

16. <u>Removal</u>

- On the underside remove the cover plate and disconnect the 37.5 MHz Generator coaxial lead from the connector bracket (Fig. 18).
- (2) Release the four screws on the underside of the receiver.
- (3) Unplug the multi-way connector from the module.
- (4) Lift out the module.

17. <u>Sub-Unit Separation</u>

 The upper deck is the 1 MHz and calibrator unit, the lower unit contains the harmonic generator, harmonic mixer and 37.5 MHz circuit.

- (2) Remove the cover from the upper unit (three screws).
- (3) Release the four captive screws in the base of the upper unit. The upper unit can now be hinged upwards.
- (4) Remove the cover from the lower unit to obtain access for adjustment and servicing.
- (5) To completely remove a unit it is first necessary to unsolder two wires from the lower unit, using a lowwattage soldering iron, as follows:
- (6) Unsolder the violet-coloured wire from the base of the lower unit.
- (7) The blue coaxial lead can be unsoldered from the <u>under-side</u> of the small transverse board on the lower unit. Undo the screws holding the board and lift upwards to give access to the unsoldering point. Use the minimum heat necessary to free the joint.

1st MIXER and 40 MHz FILTER

18. Tools: Screwdriver Spanner .25 inch AF

19. <u>Removal</u>

- (1) Disconnect the coaxial lead from the side of the R. F. Module.
- (2) On the receiver underside remove the cover plate and disconnect the 1st Mixer lead from the connector bracket.
- (3) On the underside, adjacent to the H.T. Filter Board release the screw which is located in a tubular shroud. (The one further from the rear panel).
- (4) On the upperside release one screw at each end of the 1st Mixer and 40 MHz unit.
- (5) Pull the unit upwards taking care not to trap the free coaxial leads which are attached.

2nd MIXER

20. Tools: Screwdriver

21. <u>Removal</u>

- (1) Remove the cover plate from the underside of the receiver.
- On the underside remove two of the four retaining screws.
 (from diagonally opposite corners).

- (3) Slacken off the two remaining screws and push them upwards, thereby partly easing the module out of its connector. Then completely remove these screws.
- (4) On the upperside grip the module and pull upwards. If sufficient grip cannot be obtained, remove the module cover and carefully pull on the pillars.

3rd MIXER

22. Follow the same procedure as for the 2nd Mixer in the previous paragraphs.

37.5 MHz BANDPASS FILTER

- 23. Tools: Screwdriver.
- 24. <u>Removal</u>
 - On the receiver underside remove the cover plate and release the retaining screw which is located in a tubular shroud adjacent to the H.T. Filter board. (The screw nearer the rear of the receiver).
 - (2) On the upper chassis release the two end screws from the 37.5 MHz Filter Unit.
 - (3) Ease the unit upwards, keeping it level to avoid distortion of the fixed coaxial connectors.

<u>lst V. F. O.</u>

25. Tools: Screwdriver Hexagon key Spanner . 25 AF Soldering iron (earlier receivers only)

NOTE: Instructions for refitting are given in paragraph 27.

26. <u>Removal</u>

- (1) Disconnect the violet h.t. lead from the small terminal block adjacent to the 1st V.F.O. If this terminal block is not fitted refer to operation (2).
- (2) On earlier receivers it is necessary to unsolder the violet h.t. lead from the H.T. Filter board on the underside of the receiver. (Remove the cover plate).
- (3) Slacken the grub-screws in the coupler on the 1st
 v. f. o. tuning shaft.
- (4) On the receiver underside remove the cover plate and disconnect the two lst v.f.o. coaxial leads from the connector bracket.

- (5) Hold the 1st v.f. o. firmly in position with one hand. With the other release the four retaining screws on the underside of the receiver.
- (6) When the retaining screws are free the 1st v.f.o. can be drawn free of the shaft coupler and lifted out.

lst V.F.O. Refitting

27.

- NOTE: It is assumed that all other parts of the receiver are installed and are serviceable. A digital counter to read up to 70 MHz is required.
 - Place the 1st v.f.o. module into its correct position on the chassis and slide the tuning shaft into the shaft coupler. Do not tighten the grub screws. Check that the coaxial leads are fed through to the underside of the receiver.
 - (2) Hold the 1st v.f.o. module with one hand and make it secure with the four retaining screws on the underside.
 - (3) Connect up the h.t. lead at terminal block TB2 (or resolder the violet lead if this was disconnected in operation (2).
 - (4) Attach a digital counter to one of the 1st v.f.o. coaxial output leads.
 - (5) Set the MHz control to 01'
 - (6) Connect power to the receiver and set the System switch to MAN.
 - Move the lst v.f.o. section of the tuning shaft by hand (not the MHz control) until the digital counter reads 41.5 MHz ±10 kHz.
 - (8) Check that the MHz control is at '01' and tighten the grub-screws in the shaft coupler. On completion the digital counter must read 41.5 MHz ±20 kHz.
 - (9) Refer to Chapter 3 paragraph 9 and perform the MHz calibration check.
 - (10) On completion connect both the 1st v.f.o. output coaxial leads to their correct sockets at the bracket on the underside of the receiver.

2nd V.F.O.

- 28. Tools: Spanner 10/32 U.N.F. Hex Hexagonal key
 - NOTE: Instructions for refitting are given in paragraph 30.

29. <u>Removal</u>

- (1) Using the hexagonal key slacken the grub screws in the shaft coupling.
- (2) Detach the connector which carries the chassis wiring.
- (3) Completely remove the three bolts which hold the module to the cast lugs of the chassis.
- (4) Draw the module away so that the tuning shaft slips out of the shaft coupler.

2nd V.F.O. Refitting

NOTE: A digital frequency meter (counter) is required.

30.

(1) Rotate the kHz tuning control fully <u>anti-clockwise</u> against its stop, then turn clockwise and set the readout to indicate '000'. This ensures that the correct end of the scale is selected.

- (2) Lock the kHz control at .000.
- (3) Set the CAL-FINE TUNE control to the mid-point of its rotation.
- (4) Remove the cover from the 2nd V. F. O. module and set the capacitor vanes by hand to the minimum capacitance position (moving vanes fully out of mesh and straight edges parallel).
- (5) Place the 2nd V. F. O. module into position on the chassis and enter the shaft into the flexible coupler on the control mechanism.
- (6) Make the module secure by inserting the three retaining bolts and tightening up.
- (7) Insert the connector plug into the module socket.
- (8) Connect the digital counter to the rear socket '2nd V.F.O.
 OUT'. Check that the 2nd V.F.O. switch is at INT.
- (9) Connect the power supply to the receiver and set the System switch to MAN.
- (10) By hand make a fine adjustment of the capacitor shaft to obtain a reading on the digital counter of 4.6 MHz plus or minus 50 Hz.
- (11) Tighten one grub screw in the shaft coupler and ensure that the counter reading has not changed while tightening.
- (12) Put the cover on the module and make tight the screws.

- (13) Set the System switch to CAL.
- (14) Refer to Chapter 3 paragraph 8 and perform the 2nd
 V. F. O. calibration check.
- (15) If results are satisfactory tighten both grub screws in the shaft coupler and check that the module cover is firmly screwed down. If the calibration check is outside the limits of ±500 Hz refer to (16) below.
- (16) The trimmer capacitor C2, which is accessible via a hole in the cover of the 2nd V.F.O. module (the upper of the two holes) may be adjusted to correct a calibration error of up to one kilohertz. Before making any adjustment of C2 ensure that the hand-setting of the capacitor vanes in operation (10) has been done as accurately as possible.

lst I.F. UNIT

- 31. This small unit is part of a large assembly which includes the i.f. bandwidth filters together with the Bandwidth switch and B.F.O. controls.
- 32. Tools: Screwdriver Spanner . 25 inch AF Soldering iron

33. <u>Removal of 1st I.F. Unit</u>

- (1) Remove the main I.F. Module (Refer to paragraph 11) and unplug the 2nd Mixer module.
- (2) On the underside of the receiver remove the five screws which retain the bandpass filter assembly.
- (3) On the upper side remove the two screws which secure the lst I. F. Unit to the Bandpass filter assembly.
- (4) Move the bandpass filter assembly sideways. This gives access for the unsoldering of the wires from the 1st I. F. Unit. Ensure that the coaxial leads to the 1000Ω and 100Ω output pins are identified for correct re-connection.
- (5) Push the 1st I.F. Unit out of its mounting and lift out.
- NOTE: The coaxial connection to the 3rd mixer (wire number 8) may have to be unsoldered if it restricts the movement of the filter assembly.

B.F.O. Assembly

34. The B. F. O. oscillator and amplifier boards are part of a larger assembly including the B. F. O. controls, Bandwidth switch, i.f. bandpass filters and 1st I. F. Unit. The removal of this entire assembly involves the unsoldering of a complex cable harness and should not normally be necessary. To remove the B.F.O. boards proceed as follows.

- 35. Tools: Screwdriver Soldering iron Spanner . 25 inch AF
- 36. <u>Removal of B. F. O. Boards</u>
 - (1) Remove the main I. F. Module (Refer to paragraph 11).
 - (2) On the underside of the receiver remove the five screws which retain the bandpass filter assembly.
 - (3) Move the filter assembly sideways and unsolder all connections from the 1st I.F. Unit and the crystal filters.
 - (4) Lift out the bandpass filter assembly.
 - (5) The B. F. O. boards are now accessible. Unsolder the connections from the board to be removed. Remove the four retaining screws and lift out the board.

METER CONTROL BOARD

37. Tools: Screwdriver Spanner 4BA Soldering iron.

Removal

- (1) Remove the side plate from the receiver (Six screws)
- (2) Remove the four screws which secure the pillars of the control board to the side member of the receiver.
- (3) Unsolder the connecting wires and lift out the board.
- (4) When re-connecting identify the leads as follows. Also refer to Fig. 18.

R1	Wire 34) Twin	D1/D2	Red/Blue
D2/D4	Wire 34)screened	D3/D4	Blue/Orange
R 2	Violet	R3	Blue/White

H.T. FILTER BOARD

38. Tools: Soldering iron 4BA Spanner Screwdriver

39. <u>Removal</u>

- (1) Remove the small cover from the underside of the receiver
- (2) Unsolder the connecting wires from the H.T. Filter board.

- (3) Remove the two nuts and lift up the board sufficiently to unsolder the remaining connections.
- (4) Lift out the board.

AUDIO AMPLIFIER BOARD

40. Tools: Soldering iron Screwdriver

Removal of One-Watt Audio Board

- (1) Remove the small cover from the underside of the receiver.
- Unsolder the connecting leads from pins 2, 3, 4, 5, 6, 7,
 8, 9, of the audio board.
- (3) Remove the four retaining screws.
- (4) Carefully ease the transistors VT4 and VT5 out of their holders at the same time as the board is lifted out.
- (5) When re-assembling identify the connecting wires as follows:

Pin 2:	To audio transformer (Red/Orange)	Pin 6:	Violet (Red sleeve) Violet (White sleeve)
Pin 3:	To capacitor lCl5	Pin 7:	Sleeve 32
		Pin 8:	Earth, (Black)
Pin 4:	Sleeve 31	Pin 9:	To capacitor IC17.
Pin 5:	To capacitor 1C12 (Red/White)		

41. <u>10 mW Audio Board Connections</u>

Pin 1. Sleeved lead number 31 to slider of A. F. Gain potentiometer.

- Pin 2 (Orange/grey wire to ISKT 11 pin A. (Red wire of sleeve number 34 to meter diodes 3D2/3D4 (Fig.16).
- Pin 3 { Orange/brown lead to 1SKT11 pin B. Blue lead of sleeve number 34 to resistor 3R1 (Fig. 16).
- Pin 4 To screen of lead number 33.
- Pin 5 {Wire of lead number 33 to Phones jack JKl. (Wire of screened lead (or orange/white on some receivers) to ISKT11 pin D.
- Pin 6 Violet (two leads) 16V.
- Pin 7 Wire of lead number 32 to slider of A.F. Level potentiometer.
- Pin 8 Earthing wire to chassis tag.

42. B. F. O. and Bandwidth Assembly

The following items are mounted on an assembly which can be removed as a single unit, but such removal should not normally be attempted due to the numerous connections which have to be unsoldered.

- (a) B. F. O. oscillator and amplifier boards.
- (b) lst I.F. Amplifier unit.
- (c) DET B. F. O. Mode Switch.
- (d) Bandwidth (I. F. BW) switch.
- (e) R.F. Gain potentiometer.
- (f) A.F. Gain potentiometer.
- (g) Crystal bandwidth filters.

Tools: Screwdrivers Soldering Iron

Assembly Removal

- (1) Remove all the control knobs from the front panel.
- (2) Remove the four chromium screws from the front panel and remove the panel.
- (3) Remove the main I.F. Module.
- (4) On the underside of the receiver remove the five screws which secure the B.F.O. and Bandwildth Assembly.
- (5) At the front, above the B.F.O. controls, remove two screws. The assembly is now free to move, subject to the wiring connections which must be unsoldered.

CHAPTER 8

LIST OF COMPONENTS

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NOTE: Component values are quoted as follows:

Resistors

Capacitors

No suffix	= ohms	No suffix	= microfarads
Suffix 'k'	= kilohms	Suffix 'p'	= picofarads
Suffix 'M'	= megohms		

REPLACEMENT MODULES AND ASSEMBLIES

- NOTES. 1. Table 1, below, lists the "plug-in" modules which can be quickly fitted in the receiver and are therefore essential to a rapid service procedure. Table 2 on the next page lists the principal component board assemblies likely to be required in servicing the various modules.
 - 2. When ordering a replacement module please state the module title and the part number. When ordering a circuit board or subassembly please state the module in which the part is fitted, as well as the name of the item and the part number, as listed in Table 2.
 - 3. Always quote the receiver title (RA.1217A, RA.1217B etc) and the serial number.

TABLE 1

Module Assemblies

Name

Part Number

R.F. Unit (see NOTE 2 below) CA.38538/A or B 1st Mixer and 40 MHz Filter BA.28211 2nd Mixer CA.30959 3rd Mixer CA.35970 lst VFO CA.28120 2nd VFO CA.28101/B 37.5 MHz Generator with 1 MHz Oscillator Amp and Calibrator CA.28276/B 37.5 MHz Band-Pass Filter Unit BA .28192 I.F. Module (455 kHz output) DA.38580/A I.F. Module (100 kHz output) DA.38580/B Power Unit PU.1153 CA.38640

NOTE 1: Replacement printed circuit boards are listed in Table 2 on the following page. Note that two types of i.f. converter boards are available (100 kHz and 455 kHz) and two types of audio amplifier board (10mW and 1-watt). Ensure that the correct type is ordered as fitted to the receiver being serviced.

NOTE 2: R.F. Units

R.F. Unit type CA.38538/A has soldered connections via pins on the side of the unit. R.F Unit type CA.38538/B is fitted with a Cannon multi-way connector. Type CA.38538/A is fitted in earlier versions of the RA.1217.

TABLE 2

Sub-Circuit Assemblies

R.F. Unit	Part No.	lst. Mixer	Part No.
0-30 MHz Filter R.F. Range Coil &	AA .28188/A	Mixer Board	BA.28215
Switch Assy. R.F. Amp Board Attenuator Switch Assy.	CA.38539 BA.28155/B AA.34081	40 MHz Filter	BA.28197
2nd. Mixer		3rd. Mixer	
Mixer Board	BA 28177	Mixer Board	BA.35966
lst. VFO		2nd. VFO	
Oscillator Board Amplifier Board Amplifier Board	BA .35195 BA .32535 BA .28128	Oscillator Board Amplifier Board Component Board	BA .35808 BA .35807 AA .39599
<u>1 MHz Amp and Calibrator</u>		Power Unit Type PU.1	153 (AC/DC)
Complete Assembly l MHz Amplifiers OSC & Calibrator	BA.28285/B BA.32858	Component Board	BA •37549
OSC & CALIDIACOF	BA.32860		
37.5 MHz Generator	BA • 52000	Audio Amp. Board 10 mW. or	BA.31462
37.5 MHz Generator Complete Sub-Assembly	ва.28284/в	lO mW.	BA.31462 BA.39442
37.5 MHz Generator	BA .28284/B BA .32854 BA .35836	lo mW. or	-
<u>37.5 MHz Generator</u> Complete Sub-Assembly Harmonic Gen. Board Harmonic Filter Assembly	BA 28284/B BA 32854	lO mW. or l watt.	EA.39442
<u>37.5 MHz Generator</u> Complete Sub-Assembly Harmonic Gen. Board Harmonic Filter Assembly Harmonic Mixer Board	BA .28284/B BA .32854 BA .35836 BA .37894	lO mW. or l watt. <u>Coñtrol Board</u>	EA .39442 BA .38883
37.5 MHz Generator Complete Sub-Assembly Harmonic Gen. Board Harmonic Filter Assembly Harmonic Mixer Board 37.5 MHz Amp. Board	BA .28284/B BA .32854 BA .35836 BA .37894	lO mW. or l watt. <u>Coñtrol Board</u>	EA .39442 BA .38883
<u>37.5 MHz Generator</u> Complete Sub-Assembly Harmonic Gen. Board Harmonic Filter Assembly Harmonic Mixer Board 37.5 MHz Amp. Board <u>BFO (600 kHz)</u> Oscillator Board	BA.28284/B BA.32854 BA.35836 BA.37894 BA.32850 BA.32850	lO mW. or l watt. <u>Control Board</u> <u>H.T. Filter Board</u> <u>lst. I.F. Amp. Unit</u>	BA.39442 BA.38883 AA28179/B BA.28203
<u>37.5 MHz Generator</u> Complete Sub-Assembly Harmonic Gen. Board Harmonic Filter Assembly Harmonic Mixer Board 37.5 MHz Amp. Board <u>BFO (600 kHz)</u> Oscillator Board Amplifier Board	BA.28284/B BA.32854 BA.35836 BA.37894 BA.32850 BA.30540 BA.30542 BA.30533 AA.30535 BA.28236	lo mW. or l watt. <u>Control Board</u> <u>H.T. Filter Board</u> <u>lst. I.F. Amp. Unit</u> Circuit Board Converter: 455 kHz Osc. Board (455 kHz) Amp. Board (455 kHz)	EA .39442 BA .38883 AA28179/B BA .28203 BA .31474 BA .38568A
<u>37.5 MHz Generator</u> Complete Sub-Assembly Harmonic Gen. Board Harmonic Filter Assembly Harmonic Mixer Board 37.5 MHz Amp. Board <u>BFO (600 kHz)</u> Oscillator Board Amplifier Board <u>I.F. Module</u> I.F. Amp. Board H.T. Supply Filter Detector Board	BA .28284/B BA .32854 BA .35836 BA .37894 BA .32850 BA .30540 BA .30542 BA .30533 AA .30535	lo mW. or l watt. <u>Control Board</u> <u>H.T. Filter Board</u> <u>lst. I.F. Amp. Unit</u> Circuit Board Converter: 455 kHz Osc. Board (455 kHz)	EA .39442 BA .38883 AA28179/B BA .28203 BA .31474 BA .38568A BA .34783A BA .38568B

Cct. Ref.	Va	lue	Desci	ription	Rat	t.	Tol. %		Racal Part		Manufacturer
				M	A IN	CHAS	SSIS ITEI	MS			
						(F1	<u>3. 16)</u>				
NOTES	1:			prefixed ut are not				in	vari	.ous	parts of the
	2:			prefixed embly.	'2'	are	located	in	the	Band	lwidth and BFO
	3:	Compo	nents	prefixed	'3'	are	located	on	the	Mete	er Control Board.
	4:			prefixed nderside)	'4'	are	located	on	the	н.т.	, Filter board.
	oh	ms	Resis	stors							
lrl	011		not i								
1R2 1R3			not u not u								
1R4	1.	2k		l Oxide			5		90634	6	Electrosil TR5
1R5	68			l Oxide					91011		Electrosil TR4
1R6 1R7	56 10			l Oxide l Oxide			5 5 5		90749 90827	16 16	Electrosil TR5 Electrosil TR4
1R8	10			l Oxide			5		90749		Electrosil TR5
1R9	33		Meta:	l Oxide			5		90815		Electrosil TR5
2R1 2R2 2R3	39	0	Compond not in not i		0.	1	10		90250)3	Erie 15
2R4	10	0		l Oxide			5		90827	6	Electrosil TR4
2R5			not i								
2R6 2R7	3.	3k	not i Metal	lsed 1 Oxide			5		90099)1	Electrosil TR4
3R1		- 9k	Met a	l Oxide			5		90602		Electrosil TR5
3R2	22			l Oxide			5 5		90655		Electrosil TR5
3R3	5۰	6k	Meta.	l Oxide			5		90603	2	Electrosil TR5
				ntiometers							
lRV1 lRV2	lk		not i	used er: preset					ASW35	0 11	
IRVZ IRV3	50			Fine Tune:	va	riabi	le	-	A5W))	OTT	Reliance HEL 05-B0-5 Style 2650/138
2RV1	10			Level: pre					90061		Plessey MP
2RV2 2RV3	lk 10		R.F. A.F	Gain) Gain) ^{dual}	coi	ntro;	1		bsw38 bsw38		
3RV4	47			Meter Sett:					2007-	120	
			~ .		0	pres			90863	5	Ples sey MP
4RV1	2.	2k	AGC S	Set: Linea:	r				90983	6	Plessey MP
			Сата	citors							
101	Δ	01	Cera		vo: 35(lts n	20		90212	2	Leman 310K
4V 1	•0	U 4	(Cr di	11-12 C	וכב	0	20		70212	. 🦳	Lemeo 310K

Cct. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer			
	MAIN CHASSIS ITEMS (continued)								
		Capacitors (co		. 16)					
			ontd) volts		_				
1C2 1C3 1C4 1C5	20.5p 0.1 0.22 0.1	Variable B.F.(Polyester Ceramic Polyester	0. Tune 250 50 250	20 -20+50 10	ad 3 8584 909428 908338 909847	Mullard C280AE/P100K T.C.C. CML10 Mullard C280AE/A100K			
1C6 1C7 1C8 1C9 1C10	0.1 0.1 0.1	not used not used Polyester Polyester Polyester	250 250 250	20 20 20	909428 909428 909428	Mullard C280AE/P100K Mullard C280AE/P100K Mullard C280AE/P100K			
1C11 1C12 1C13 1C14 1C15 1C16 1C17	0.1 500 10 .0022 .033 0.1 500	Polyester Electrolytic Electrolytic Ceramic Polyester Polyester Electrolytic	250 25 20 350 150 250 25	20 -20+50 20 20 5 20 -20+50	909428 900748 905399 902126 908114 909428 900748	Mullard C280AE/PlooK Hunts MEF37T S.T.C. 472/LWA/401GA Lemco 310K Wima Tropyfol M. Mullard C280AE/PlooK Hunts MEF37T			
201 202 203	0.1 Refer to l	Polyester 1C2 in B.F.O. Electrolytic	250 Assembly 40	10 • Page 8 -10+100	909847 - 29. 910952	Mullard C280AE/AlOOK Mullard C426 AS/G1			
4C1	0.1	Polyester	250	٥٢	909847	Mullard C280AE/A100K			
4C2 4C3	50	not used Electrolytic	25	-10+50	908798	Mullard C426AR/F50			
፲፹፲		Transformers Audio output	transform	ner	CT39257				
ISA ISB ISC/1 ISC/2 ISD		Microswitch a Microswitch h)Power switch)of System sw	Switches Microswitch a.g.c. Microswitch h.t.)Power switching: part)of System switch 1SE 2nd V.F.O. INT/EXT: break		911132 911132 BSW38693	Unimax DF5T Type USMF			
1SE 1SF 2SA 2SB			before ma break be teb	ıke	ASW38678 BSW38691 912063 BSW38719 BSW38585	E.M.I. S5			

Cet. Ref.	Value	Description Rat. Tol %	L. Racal Part No.	Manufacturer						
	MAIN CHASSIS ITEMS (continued)									
		<u>(</u> Fig. 16	<u>.)</u>							
זייניב		Connectors								
	(Free)									
XLPI 1PI2		Antenna plug: 75 ohms coax. Chassis wiring to Power mod		Transradio BN 1/7						
		15 w	vay 908598	Cannon DAM15P						
1LP3		I.F. module input: coaxial elbow	L: 90 83 70	Belling Lee L1403/RFP/Ag						
	free: 12-w sory Set f		906612	Plessey Mk4 2CZ83302/5 Plessey 508/1/03013/205						
Socket	ts									
lskT1 lskT2 lskT3 lskT4 lskT5		I.F. connector: free: 37-wa Double ended bulkhead adap Double ended bulkhead adap lst Mixer: insert HT to lst Mixer	otor 908405	Cannon DCM37S Belling Lee L1403/BS/Ag Belling Lee L1403/BS/Ag Cannon DM53743-5001 Sealectro 5BC						
lskt6 lskt7 lskt8 lskt9 lskt10		37.5 MHz Filter: insert 2nd Mixer connector - mult 3rd Mixer connector - mult 37.5 MHz Gen. connector-mult 2nd VFO connector-multiway The coaxial inserts in the SKT9 and SKT10 can be rener	tiway 911160 11tiway 908602 7 908599 2 Cannon connec 2wed separately							
lsktij	٦	insert DM53743-5001: Part 1 12-way outlet: fixed.	No. 908600. 906607	Plessey Mk4 CZ63979/5						
	NOTE:	The free coaxial plugs for	connection to	o the rear panel						
י (ודאיה י	~	sockets 1SKT12 to 1SKT17 as	are BNC 75Ω Par	t number 905034						
1SKT12 1SKT13		2nd VFO IN: fixed, 75Ω Coa. 2nd VFO OUT: fixed, 75Ω Coa								
1SKT 14	<u>ī</u> 4	L.F. fixed, 750 Coa	axial 90745 7	Transradio BN 12/7						
lSKT15		PAN fixed, 75Ω Coa	axial 907457	Transradio BN 12/7						
1SKT16		l MHz IN fixed, 75Ω Coa:								
1SKT17 1SKT18	-	l MHz OUT: fixed, 75Ω Coa Connector 5-way, free (not		Transradio BN 12/7						
		fitted to earlier receiver	rs) 911785	Cannon DEM5W1S						
Coaxis	al insert	to SKT18	907076	Cannon 53742-5001						
JKl Phones	s jack plu	Phone jack socket ug (free) to fit JKl	9015 0 9 9015 57	Igranic P71 Igranic P50						
		Terminal Blocks								
TBL		6-way. Refer to I.F. Modu								
TB2		lst V.F.O. h.t. connection	n AD 3983 51	l (not fitted on early receivers)						
ТВЗ		(in fine tune circuit)		Wingrove & Rogers TS8-04						
		Meter								
Ml		'S' Meter	BD35519							
1LP1-1	1LP3	Dial Lamps	908605	Vitality 14v690						
				8 - 5						
RA.121	17			Vol. 2						

Cct. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer
		MAII	CHASSI	S ITEMS (co	ontinued)	
3D1 to 31	D4:	<u>Diodes</u> Meter rectifie: Crystal Bandpa:		7 6	900071	Mullard OA91
	13 kHz 8 kHz 6.5 kHz 6.0 kHz	BD38733/A 99 BD38733/B 99 BD38733/H 99	5467 5028 6029 5943	3 kHz	BD38733/C BD38733/D BD38733/F BD38733/F	995 029 9955 03
		<u>10 i</u>	nW AUDIO	AMPLIFIER	(BC.31462)
R1 R2 R3 R4 R5	ohms 5.6k 2.2k 5.6k 330 2.2k	<u>Resistors</u> Metal Oxide Metal Oxide Metal Oxide Metal Oxide Metal Oxide		5 5 5 5 5	908273 908270 908273 908268 908270	Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4
R6 R7 R8 R9 R10	15 k 5.6k 22k 150k 150k	Metal Oxide Metal Oxide Metal Oxide Metal Oxide Metal Oxide		5 5 5 5 5	908280 908273 908269 908277 908277	Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4
R11 R12 R13 R14 R15	470 470 1k 150k 10k	Metal Oxide Metal Oxide Metal Oxide Metal Oxide Metal Oxide		5 5 5 5 5	900992 900992 908267 908277 900986	Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4
R16 R17 R18	68k 2.2k 68	Metal Oxide Metal Oxide Metal Oxide		5 5 5	908279 908270 908278	Electrosil TR4 Electrosil TR4 Electrosil TR4
C1 C2 C3 C4 C5	80 .022 2.5 80 2.2	<u>Capacitors</u> Electrolytic Polyester Electrolytic Electrolytic Tantalum	16 250 16 16 20	-10+50 20 -10+50 -10+50 20	908810 900082 908808 908810 908316	Mullard C426 AR/E80 Mullard C280 AE/P22K Mullard C426 AS/E2.5 Mullard C426 AR/E80 U.Carbide K2R2J20S
C6 C7 C8 C9 C10 C11 C12	80 80 .001 80 10 0.1 50	Electrolytic Electrolytic Silver Mica Electrolytic Electrolytic Polyester Electrolytic	80 80 500 16 16 250 25	-10+50 -10+50 20 -10+50 -10+50 10 -10+50	908810 908810 908315 908310 900068 909847 908798	Mullard C426 AR/E80 Mullard C426 AR/E80 Erie Microcap Mullard C426 AR/E80 Mullard C426 AR/E10 Mullard C426 AR/E10 Mullard C426 AR/F50
T1 T2 T3		Transformers Transformer as Transformer as	sembly		CT 31476 CT 31478 CT 31477	
VTl		<u>Transistors</u> p-n-p			911929	Mullard BC107
Vr2,Vr3,	Vrr4	n-p-n			909017	Texas 2N929
						8 - 6
RA 1217						Vol. 2

Cct. Ref.	Value	Description	Rat.	Tol. \$	Racal Part No.	Manufacturer				
	AUDIO AMPLIFIER BOARD									
(BC.39442)										
R1 R2 R3 R4 R5 R6 R7 R8 R9	ohms 4.7k 1.5k 12k 1k 390 1k 1 1 15	Resistors Composition Composition Composition Composition Composition Wirewound Wirewound Composition	watts 0.1 0.1 0.1 0.1 0.1 1.5 1.5 0.1	10 10 10 10 10 10 10 10 10	902516 902521 902521 902508 902503 902508 911767 911767 902486	Erie 15 Erie 15 Erie 15 Erie 15 Erie 15 Erie 15 Welwyn W21 Welwyn W21 Erie 15				
R10 R11 R12 R13 R14 R15	33k 3.9k 120 3.3k 22k 100	Composition Composition Composition Composition Composition Composition	0.1 0.1 0.1 0.1 0.1 0.1	10 10 10 10 10 10	902526 902515 902497 902514 902524 902524 902496	Erie 15 Erie 15 Erie 15 Erie 15 Erie 15 Erie 15				
R16 R17 R1 8 R19	39 0 1k 470 10	Composition Composition Composition Composition	0.1 0.1 1/4	10 10 10 10	902503 902508 902504 902411	Erie 15 Erie 15 Erie 15 Erie 16				
RV1 RV2	200k 2k	Potentiometers Capacitors	<u></u>		911766 911765	Morganite 81E Morganite 81E				
C1 C2 C3 C4 C5 C6* C7 C8	32 1 10p 80 0.1 1 470p 10	Electrolytic Electrolytic Polystyrene Electrolytic Polyester Tantalum Polystyrene Tantalum	volts 10 40 125 16 250 25 125 15	-10+50 -10+100 2 -10+50 20 20 5 20	911764 910952 900814 908810 909428 905362 908735	Mullard C426AR/D32 Mullard C426AS/G1 Salford PF Mullard C426AR/E80 Mullard C280AE/P100K S.T.C. TAG 1.0/25 Salford PF S.T.C. TAG10/15				
VT1 VT2 VT3 VT4 VT5 VT6, V	77, VT8	<u>Transistors</u> n-p-n p-n-p n-p-n n-p-n n-p-n Diodes			911929 911929 911928 908753 908753 909927	Mullard BCl07 Mullard BCl07 Mullard BCY71 Mullard BFY51 Mullard BFY51 Texas 2N3711				
Dl and	L D2	<u>Diodes</u> Semi-conductor	~		906001	Texas 1S130				
		ullard C426 AS/		rly produc	-					

*NOTE: C6 is Mullard C426 AS/G1 in early production versions

$ \begin{array}{c c c c c c c c c c c c c c c c c c c $	Cct. Ref.	Value	Description R	at.	Tol. ¢	Racal Part No.	Manufacture	2 r
Coll Board and Capacitor Assembly watts $Ohms$ $Vatts$ 92R11 KDMetal 0xide52R2100Metal 0xide52R3100Metal 0xide52R4100Metal 0xide52R5100Metal 0xide52R5100Metal 0xide52R652arbon H1. Stab.52R356Carbon H1. Stab.52R435Carbon H1. Stab.52R547Carbon H1. Stab.52R635Erie M6R435Carbon H1. Stab.52R620630Erie M6R547Carbon H1. Stab.52R620630Erie M6R653Carbon H1. Stab.52R02arbon H1. Stab.52R02arbon H1. Stab.52R182Metal 0xide52R182Metal 0xide52R220Metal 0xide52R356Metal 0xide52R4	,		#**** *	<u>R.</u> F.	MODULE			
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$ \begin{array}{cccccccccccccccccccccccccccccccccccc$			Carbon Hi. Stab.		5			
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$		33	Carbon H1. Stab.	•	5			
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	R5	47	Carbon H1. Stab.	•	5	906435	Erie N6	
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	R6	33	Carbon Hi. Stab.		5	908301	Erie NG	
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Thermistor					5		Electrosil	TR4
Thermistor	3r18				5	9 08 690	Welwyn F25	
	3 R19	22	Metal Oxide		5	9114 9 5	Electrosil	TR ¹ 4
			Thermistor					
3TH1 909839 Mullard VA1038	3TH1					909839	Mullard VA	1038

Cct. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer
<u> </u>		<u>R.</u>	F. MODUI	E (continu	ued)	
		Potentiometer				
4RV1	2.2k	(On H.T. Filt	er below	chassis)	909836	Plessey MP Linear
Madula		Capacitors				
Module			volts			
C1 C2	0.1	not used Polyester	250	20	909428	Mullard C280AE/P100K
<u>0 - 30</u>	MHz Fil	ter (AC.28188/A)			
1C1 1C2 1C3 1C4 1C5	68p 82p 18p 68p 68p	Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene	30 30 30 30 30	220 20 10 20 20 20 20 20 20	908321 908322 908323 908321 908321 908321	Suflex HS7/A Suflex HS7/A Suflex HS7/A Suflex HS7/A Suflex HS7/A
	-	itor Assembly	-	C		
201 202 203 204 205	бр бр бр бр бр	trimmer: tub trimmer: tub trimmer: tub	ular ular ular ular ular		901987 901987 901987 901987 901987 901987	Mullard COO4EA/6E Mullard COO4AE/6E Mullard COO4AE/6E Mullard COO4AE/6E Mullard COO4AE/6E
2 С ба 2СбЪ 2С7		le: R.F. TUNE. 157 pF each sec		8.5 pF))		Wingrove & Rogers C78-22/
207 208 209 2010 2011	4.7p 680p 680p 6.8p	Ceramic Silver Mica Silver Mica Ceramic	750 300 300 750	12p 2 2 12p	902007 902254 902254 902009	Lemco 310P100 J & M C12S j & M C12S Lemco 310P 100
R.F. A	mplifier	Board (BC.2815	<u>5)</u>			
301 302 303 304 305	0.1 0.1 0.1 0.1 0.1	Ceramic Ceramic Ceramic Ceramic Ceramic	30 30 30 30 30 30	-25+50 -25+50 -25+50 -25+50 -25+50	906675 906675 906675 906675 906675	Erie 811T/30 Erie 811T/30 Erie 811T/30 Erie 811T/30 Erie 811T/30
306 307 308 309 3010	0.1 68p 100p 27p 220p	Ceramic Polystyrene Polystyrene Polystyrene Polystyrene	30 30 30 30 30	-25+50 21 21 21 21 21 21 21 21 21	906675 908321 908241 908325 908320	Erie 811T/30 Suflex HS7/A Suflex HS7/A Suflex HS7/A Suflex HS7/A
3011 3012 3013 3014 3015	33p 150p 0.1 0.1 100p	Polystyrene Polystyrene Ceramic Ceramic Polystyrene	30 30 30 30 30	±lp 2. -25+50 -25+50 2 ¹ 2	906497 908 33 1 906675 906675 908241	Suflex HS7/A Suflex HS7/A Erie 811T/30 Erie 811T/30 Suflex HS7/A

Cct. Ref.	Value	Description R	lat.	Tol. %	Racal Part No.	Manufacturer			
R.F. MODULE (continued)									
3016	18p	Polystyrene 3	50	±lp	907171	Suflex HS7/A			
3017	220p		50	2 ¹ /2	908320	Suflex HS7/A			
3018	33p	· · ·	50	+]n -2	906497	Suflex HS7/A			
3C19	82p	· • •	50	±1p 2]	908322	Suflex HS7/A			
3020	0.1		50	-25+50	906675	Erie 811T/30			
3021	68p		30	2호	908321	Suflex HS7/A			
		Traductore and Ma		-		,			
0 - 30	MHz Fil	<u>Inductors and Tr</u> ter	ansior	ners					
<u></u> 1L1		Aerial Filter Co	.47		00 20062				
112					CT.32963				
		Aerial Filter Co			СТ.32964				
113		Aerial filter Co	11		CT.32963				
<u>Coil a</u>	nd Swite	h Assembly							
2L1		Coil Assembly			CT-32934				
212		Coil Assembly			CT.32933	/в			
2L3		Coil Assembly			CT.32932				
214		Coil Assembly			CT.32931/B				
2L5		Coil Assembly			CT 32930				
<u>R.F.</u> A	mplifier	Board							
3L1		Coil Assembly:	L.P. F	ilter	ст.28154				
312		Coil Assembly:			CT.28152				
3L3		Coil Assembly:			CT.28154				
314		Coil Assembly:			CT.28152				
3T1		Transformer Asse	emblv		ст.28149				
372		Transformer Asse			CT 28149				
		Switches							
2SA		R.F. RANGE Swite	2h		BSW38550				
SB		AE ATT (R.F. Att	enuato	r Switch)	BSW28141				
Switch	BB Asse	mbly, complete		,	AA 34081				
		Transistors							
<u>R.F. A</u>	mplifier	Board							
3vf1					900618	Texas 28303			
3VT2					909111	RCA 2N3478			
3VT3					909111	RCA 2N3478			
		Diodes							
<u>R.F. A</u>	mplifier								
3Dl to	> 3D4	See NOTE below			911796	Hughes HPS1672/B			
	nd 3D6		liodes		908343	Texas 18920			

RA.1218/RA.1217

Cct. Ref.	Value	Description	Ret.	Tol. %	Racal Part No.	Manufacturer
3D7 to 3D11 an	-	See NOTE at foot of page Semi-conductor of Plugs and Socket	8-10 diodes	IODULE	(continued) 911796 909837	Hughes HPS1672/B Hughes HS9003
R.F. IN PL1 Coaxial	PUT insert t	Antenna: socket Connector fixed o PL1				Transradio BN12/7 Cannon DEM5W1P Cannon DM53740-5001
lFS1 Fusehol	der for l	<u>Fuse</u> 500 mA Aerial pr FS1 <u>Miscellaneous I</u> t		.on	906850 908352	Beswick TDC134 Belling Lee L1383
X1 3X1		Ferrite Bead Ferrite Bead on	3V T3			Mullard FX1242 Mullard FX 1242
		rester (Spark Gaj OΩ 17.6V)			Siemens America Inc. Type Bl-A230 C.P. Clare Type F F.L. 7631

Cct. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer
		FIRST	MIXER A	ND 40 MHz	FILTER	
			(BC	.28211)		
		Resistors				
lst.	Mixer Boa	rd (BC.28215)				
	ohms					
Rl	56	Metal Oxide		5	908289	Electrosil TR4
R2	15k	Metal Oxide		5 5 5 5 5 5	908280	Electrosil TR4
R3	12k	Metal Oxide		5	908274	Electrosil TR4
$\mathbf{R}^{\mathbf{\bar{4}}}$	2.7k	Metal Oxide		5	908294	Electrosil TR4
R5	2 .7 k	Metal Oxide		5	908294	
40 MH	z Filter	(AC.28197)				
Rl	lk	Metal Oxide		5	908267	Electrosil TR4
		Potentiometer:	3			
lst.	<u>Mixer Boa</u>	rd				
RVl	2.2k				909838	Ancillary Dev. Type T.O.
		Capacitors				
lst.	Mixer Boa	rd (BC.28215)				
Cl	0.1	Ceramic	30	-25+50	906675	Erie 8111/30
C2	0.1	Ceramic	30	-25+50		
C3	0.1	Ceramic	30	-25+50		
C4	0.1	Ceramic	30	-25+50		
C5	39p	Polystyrene	30	lp	905374	Suflex HS7/A
c 6	39p	Polystyrene	30	lp	905374	Suflex HS7/A
			-	-		
		(AC.28197)				
Cl	47p	Polystyrene	30	2호	908318	Suflex HS7/A
C2	47p	Polystyrene	30	2 2	908318	Suflex HS7/A
C3	47p	Polystyrene	30	2출	908318	Suflex HS7/A
C4	47p	Polystyrene	30	2출	908318	Suflex HS7/A
C5	4 7 p	Polystyrene	30	ୟ ଧ ସ ସ ସ ସ ସ ସ ସ ସ ସ ସ ସ ସ ସ ସ	908318	Suflex HS7/A
C6	47p	Polystyrene	30	2 = 2	908318	Suflex HS7/A
CZ	56p	Polystyrene	30	2출	908 3 19	Suflex HS7/A
c 8	470p	Polystyrene	30	2 2	908317	Suflex HS7/A
		Inductors				
lst.	Mixer Boa	rd				
Tl		Transformer			CT 28212	

Cct. Ref.	Value	Description	Rat.	Тоl. %	Racal Part No.	Manufacturer
		FIRST MIXER	AND 40	MHz (conti	nued)	
40 MHz	Filter					
L1 12 13 14 15		Coil Assembly Coil Assembly Coil Assembly Coil Assembly Coil Assembly			CT 31031 CT 31030 CT 31030 CT 31030 CT 31030 CT 31030	
16 17 18		Coil Assembly Coil Assembly Coil Assembly			CT 31030 CT 31030 CT 31030 CT 31030	
		Transistors				
VT1 VT2					906517 906517	Texas 2N918 Texas 2N918
		Plugs and Sock	cets			
PL1 PL2 DL7		From R.F. Unit -16V supply			906 3 91 908 3 40	Belling Lee L1403/FP/Ag. Sealectro FT-M-4
PL3 PL4		Connects to bu To 2nd. mixer		adaptor KT2	908 3 70 908341	Belling Lee L1403 RFP/Ag. Cannon insert DM 53740-5001

Cct. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer			
			SECC	ND MIXER					
			(BC	.30959)					
		<u>Resistors</u>							
Module Chassis ohms									
1R1	270	Metal Oxide			908284	Electrosil TR4			
1R2 +	1R3 180	not used		F	000105	Pleatment D mpl			
1R4	100	Metal Oxide		5	909125	Electrosil TR4			
<u>Circui</u>	t Board	(BC.28177)							
Rl	47k	Metal Oxide		5	908391	Electrosil TR4			
R2	4.7k	Metal Oxide		5 5 5 5 5	900989	Electrosil TR4			
R3	12k	Metal Oxide		5	908274	Electrosil TR4			
R4	680	Metal Oxide		5	908390	Electrosil TR4			
R5	8.2k	Metal Oxide		5	908275	Electrosil TR4			
r6	2 .2k	Metal Oxide		5	908270	Electrosil TR4			
R7	3.3k	Metal Oxide		5	900991	Electrosil TR4			
r8	lOk	Metal Oxide		5 5 5 5 5	900986	Electrosil TR4			
R9	3•9k	Metal Oxide		5	900990	Electrosil TR4			
RlO	470	Metal Oxide		5	900992	Electrosil TR4			
Rll		not used							
R12	56	Metal Oxide		5	908289	Electrosil TR4			
		Capacitors	volts						
Cl	.047	Polyester	250	20	909227	Mullard C280 AE/P47K			
Č2	.047	Polyester	250	20	909227	Mullard C280 AE/P47K			
C3	68p	Polystyrene	30	$2\frac{1}{2}$	908321	Suflex HS7/A			
C4	0.1	Polyester	250	lŌ	909847	Mullard C280 AE/AlOOK			
C5	0.1	Polyester	250	10	909847	Mullard C280 AE/A100K			
с 6	.01	Polyester	250	20	910485	Mullard C280 AE/PlOK			
C7	.01	Polyester	250		910485	Mullard C280 AE/PloK			
C8	150p	Polystyrene	30	20 2 1	908331	Suflex HS7/A			
C9	.01	Polyester	250	20	910485	Mullard C280 AE/PLOK			
ClO	•047	Polyester	250	20	909227	Mullard C280 AE/P47K			
CII	47p	Polystyrene	30	$2\frac{1}{2}$	908318	Suflex HS7/A			
C12	0.1	Polyester	250	10	909847	Mullard C280 AE/AlOOK			
C13	100p	Polystyrene	30	21/2	908241	Suflex HS7/A			
C14	300p	Polystyrene	30	10 2½ 2½	908335	Suflex HS7/A			
C15	0.1	Polyester	250	lŌ	909847	Mullard C280 AE/A100K			
C16	15p	Polystyrene	30	lp	908336	Suflex HS7/A			
C17	-25 56p	Polystyrene	30	21/2	908319	Suflex HS7/A			
c18	150p	Polystyrene	30	1p 2 <u>1</u> 22	908331	Suflex HS7/A			
C19	15p	Ceramic	750	5	902047	Lemco 310NPO			
Cct. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer			
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		SE	COND MIX	ER (contin	nued)				
		Transformers	and Indu	ctors					
T1 T2		37.5 MHz coup 2-3 MHz outpu			CT 28317 CT 28316				
L1 L2 L3 L4 L5		Mixer load Filter coil a Filter coil a Filter coil a Filter coil a	ssembly ssembly		CT 28310 CT 28311 CT 28312 CT 28313 CT 28314				
ы		Filter coil a	ssembly		CT 28315				
		Transistors							
VT1 VT2 VT3					909414 909414 910866	Mullard 2N3323			
		Plugs and Soc	kets						
PLl		9-way Coaxial inser	ts A l t	оАЧ	90 830 0 908341	Cannon DM9W4P Cannon DM5374 0- 5001			
		Ferrite Boad							
1 X1 1X2		Ferrite Bead Ferrite Bead	Assembly	7	AA40269	Mullard FX1242			

Cet. Ref.	Value	Description	Rat.	Тоl. %	Racal Part No.	Manufacturer
			THIR) MIXER		
			(BC.	35970)		
		<u>Resistors</u>				
Module	e Chassis					
lRl	ohms 39	Composition	watts	% 10	902491	Erie 15
<u>Circu</u>	it Board	(BC.35966)				
R1 R2 R3 R4 R5 R6 R7 R8 R9 R10 R11 R12 R13 R14	18k 82k 3.9k 2.2k 470 33 82 680 33 5.6k 18k 1k 22 1k	Metal Oxide Metal Oxide Composition Metal Oxide	0.1	5 5 5 5 5 5 5 5 5 5 5 5 5 5 5 5 5 5 5	908272 908691 900990 908270 900992 908690 908290 908390 908690 908273 908273 908272 908267 902488 908267	Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4 Welwyn F25 Electrosil TR4 Electrosil TR4 Welwyn F25 Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4
R15 R16	330 68	Metal Oxide Metal Oxide		5 5 5	908268 910487	Electrosil TR4 Welwyn F25
	e Chassis	Capacitors		,	<u>7-0-01</u>	ncanju roj
·			volts			
101 102 103 104 105 106	820p 820p .0033 50 0.1 470p	Polystyrene Polystyrene Silvered Mica Electrolytic Polyester Polystyrene	30 30 200 25 250 30	2 1 2 1 2 -10+50 20 2 1 2	908389 908389 902204 908798 909428 909428 908317	Suflex HS7/A Suflex HS7/A JMC CX22S/200 Mullard C426/AR/F50 Mullard C280 AE/Pl00K Suflex HS7/A
<u>Circu</u>	it Board	(BC.35966)	volts			
C1 C2 C3 C4 C5	0.1 0.1 680p 150p .015	Polyester Polyester Silver Mica Polystyrene Silver Mica	250 250 300 30 125	10 10 2 2 1	909428 909428 902254 9083 3 1 910928	Mullard C280 AE/PlOOK Mullard C280 AE/PlOOK J.M.C. C12S Suflex HS7/A S.T.C. 454-1WA-75

Cet. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer
		TH	ERD MIXE	R (conti	nued)	
c6 C7 C8 C9 C10	680p 39p 0.1 0.1 68p	Silver Mica Polystyrene Polyester Polyester Polystyrene	300 30 250 2 5 0 30	2 1p 20 20 2 1 2	902254 905374 909428 909428 908 3 21	J.M.C. C12S Suflex HS7/A Mullard C280 AE/P100K Mullard C280 AE/P100K Suflex HS7/A
C11 C12 C13	150p 0.1 0.1	Polystyrene Polyester Polyester	30 250 250	2 1 20 20	908 33 1 909428 909428	Suflex HS7/A Mullard C280 AE/P100K Mullard C280 AE/P100K
		Transformers a	nd Indu	ctors		
Module 1L1 1L2	e Chassis	Coil Assembly Coil Assembly			CT 31026 AA 38847	
<u>Circu</u> Tl T2 T3	it Board	Transformer Transformer Transformer			CT 35968 CT 35968 CT 35968	
11 12 13 14 15	·	Coil Assembly Coil Assembly Coil Assembly Coil Assembly Coil Assembly			CT 35969 CT 31023 CT 35971 CT 31022 CT 34745	
16 17		Coil Assembly Coil Assembly			CT 31021 CT 31020	
VT1 VT2 VT3		<u>Transistors</u> n.p.n. n.p.n. n.p.n. Diodes			90 08 93 900893 900911	S.T.C. BSY27 S.T.C. BSY27 Texas 2N3478
1D1 D1-D4		(mounted on th Diode Quad on	circuit		908349 909846	Hughes HD1871 Cosem A505GE
PLl		Plugs and Soch (Plug Shell (Plug coaxial		(4)	908388 908341	Cannon DBM9W4P Cannon DM53740-5001

Cct. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer
			FIRST	V.F.O.		· · · · · · · · · · · · · · · · · · ·
			(BC.	28120)		
		<u>Resistors</u>				
Oscil.	lator Board	d_(1st. V.F.O.) (AC.3	51 <u>95)</u>		
	ohms		watts			
Rl	100	Metal Oxide		5	908276	Electrosil TR4
R2	10	Composition	0.1	10	902484	Morganite XL
R3	4.7k	Metal Oxide		5	900989	Electrosil TR4
R 4	6.8k	Metal Oxide		5 5	900987	Electrosil TR4
R5	68	Metal Oxide		5	908278	Electrosil TR4
r6	lOk	Metal Oxide		5	900986	Electrosil TR4
Amplit	fier Board	(1st V.F.O.)	(AC.325	35)		
Rl	68	Metal Oxide		5	908278	Electrosil TR4
R2	1.5k	Metal Oxide		5 5 5 5	908296	Electrosil TR4
R3	8.2k	Metal Oxide		5	908275	Electrosil TR4
R4	5.6k	Metal Oxide		5	908273	Electrosil TR4
R5	lOk	Metal Oxide		5	900986	Electrosil TR4
Rб	4.7k	Metal Oxide		5	900989	Electrosil TR4
R7	56	Metal Oxide		5 5 5 5	908289	Electrosil TR4
R8	330	Metal Oxide		5	908268	Electrosil TR4
R 9	220	Metal Oxide		5	900988	Electrosil TR4
RIO	680	Metal Oxide		5	908390	Electrosil TR4
Rll	100	Composition	0.1	10	902496	Erie 15
Ampli	fier Board	(1st V.F.O.)	(AC.281	.28)		
Rl	68	Metal Oxide		5	908278	Electrosil TR4
R2	1.5k	Metal Oxide		5 5 5	908296	Electrosil TR4
R3	8.2k	Metal Oxide		5	908275	Electrosil TR4
\mathbf{R}^{4}	5.6k	Metal Oxide		5	908273	Electrosil TR4
R5	lOk	Metal Oxide		5	900986	Electrosil TR4
r6	4.7k	Metal Oxide		5	900989	Electrosil TR4
R7	56	Metal Oxide		5	908289	Electrosil TR4
RŚ	330	Metal Oxide		5	908268	Electrosil TR4
R 9	68	Metal Oxide		5	908278	Electrosil TR4
RlO	680	Metal Oxide		5 5 5 5 5	90 83 90	Electrosil TR4
		Capacitors				
Modul	e Assembly		•••• ³ +	đ		
	¥_	Monohart a dama	volts	ø		
101	.047	Megahert z tuni Ceramic	-		CA .27752 908722	Frie Filteress 1001 of 1
102	• V " † (Ceramic	200		900122	Erie Filtercon 1201-051

Cct. Ref.	Value	Description	Rat	Tol %	Racal Part No.	Manufacturer
				O. (cont	······································	
Oscille	tor Board	AC.35195)				
0201110		(<u>((())))</u>)				
Cl	6.8p	Ceramic	750	$\frac{1}{2}p$	902075	Erie N750A
C 2	8.5p	Trimmer			908732	Erie 562-013
C3	.047	Polyester	250	20	909227	Mullard C280 AE/P47K
C4 C5	390p •047	Polystyrene Polyester	30 250	5 20	906710 909227	Suflex HS7/A Mullard C280 AE/P47K
6)	•047	roryester	270	20	909221	Multaru 0200 Abyr41k
C6	.0015	Ceramic	350	20	902124	Lemco 310K
<u>Amplifi</u>	er Board	(AC.32535)				
Cl	.047	Polyester	250	20	909227	Mullard C280 AE/P47K
C2	470p	Polystyrene	30	5	908396	Suflex HS7/A
C3	.00 <u>3</u> 3	Ceramic Hi k	350	20	902128	Lemco 310K
C4	.047	Polyester	250	20	909227	Mullard C280 AE/P47K
C5	6.8p	Ceramic	750	$\frac{1}{2}p$	902075	Lemco 310N750
C6	47p	Polystyrene	30	2 1	908318	Suflex HS7/A
C7	.001	Disc Ceramicon	500	-20+80	908832	Erie K7004/861
C8	.002	Disc Ceramicon	500	-20+80	_	Erie K800011/861
C9	82p	Polystyrene	30	2 1	908322	Suflex HS7/A
Amplifi	ler Board	(AC.28128)				
	-	Cl-C9 are identio Disc Ceramicon	cal to 500	the Ampl -20+80	ifier board 908832	AC.32535 above. Erie K7004/861
C10	.001	Transformers and 1			900092	Mie Wood/oor
Ll		Oscillator coil as			CT 28200	
			Ū			
		Transistors				
<u>Os</u> cilla	ator Board	<u>1</u>				
VTl		p.n.p.			910866	Texas GM 290A
Amplif:	ier Board	BA 32535				
VTl		p.n.p.				Texas GM 290A-R2044
VT 2		p.n.p.			910 8 66	Texas GM 290A
Amplif	ier Board	BA 28128				
VTl		p.n.p.			910866	Texas GM290A
VT2		p.n.p.				Texas GM290A-R2044
		Plugs and Sockets				
PLL		Cable and connect		-	AA 33091/9	
PL2		Cable and connect	or asse	mbly	AA 33091/6)

Cct. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer
			SECON	D V.F.O.		
			(cc.2	28101/в)		
		Resistors				
<u>Fine</u> I	une Board	L				
	ohms		watts	% 5	_	
2R1	47k	Metal Oxide		5	908391	Electrosil TR4
<u>Oscil</u>	lator Boan	rd (BA.35808)				
Rl	100	Metal Oxide		5	908276	Electros11 TR4
R2	390	Met a l Oxide		5	908472	Electrosil TR4
R3	2.2k	Metal Oxide		5	908270	Electrosil TR4
$\mathbf{R}^{\mathbf{k}}$	27k	Metal Oxide		5 5 5 5 5	908295	Electrosil TR4
R5	390	Metal Oxide		5	908472	Electrosil TR4
R6	270	Metal Oxide		5	908284	Electrosil TR4
R7	390	Metal Oxide		5	908472	Electrosil TR4
R8	150	Metal Oxide		5 5 5	909121	Electros11 TR4
Amolii	fier Board	1_(BA.35807)				
				-	908277	
Rl R2	150k 1.8k	Metal Oxide Metal Oxide		5 5	900277 908283	Electrosil TR4 Electrosil TR4
R3	T.OK	not used)	900205	ETECCIOSIT IK+
R4	1.8k	Metal Oxide		5	908283	Electrosil TR4
R5	150k	Metal Oxide		5 5	908277	Electrosil TR4
	-				•	
Rб	75	Metal Oxide		5 5 5 5 5	908288	Electrosil TR4
R7	6.8k	Metal Ocide		5	900987	Electros11 TR4
r8	lk	Metal Oxide		5	908267	Electrosil TR4
R9	lk	Metal Oxide		5	908267	Electrosil TR4
RIO	470	Metal Oxide		5	900992	Electrosil TR4
Rll	82	Metal Oxide		5	908290	Electrosil TR4
R12	18	Composition	0.1	10	902487	Erie 15
R13	68	Metal Oxide		5	908278	Electrosil TR4
R1 4	lOk	Metal Oxide			900 98 6	Electrosil TR4
R15	27k	Metal Oxide			908295	Electrosil TR4
R1 6	220	Metal Oxide			900988	Electrosil TR4
R17	22	Composition	0.1	10	902488	Erie 15
R18	lk	Metal Oxide			908267	Electrosil TR4
R19	4.7k	Metal Oxide		5	900989	Electrosil TR4
R20	27k	Metal Oxide		5 5 5	908295	Electros11 TR4
נסת	100	Noto] 0 1-		F	000006	
R21 R22	100 75	Metal Oxide		5 5	908276 908288	Electrosil TR ⁴
R 22	75	Metal Oxide		2	200200	Electros11 TR4

Cct. Ref.	Valuc	Description	Rat.	тоl. %	Racal Part No.	Manufacturer		
SECOND V.F.O. (continued)								
		Capacitors						
Module	Assemb.	ly	volts					
Cl	0 -	kHz tuning				Racal CA 30948		
C2	8.5p	kHz tuning tri	mmer		908732	Erie 562-013		
	'une Boai	<u>rd</u>			_			
2 C 1	0.1	Polyester	250	20	909428	Mullard C280AE/PlooK		
<u>Oscill</u>	lator Boa	ard (BA.35808)						
Cl	.01	Ceramicon	100	-20+80	900067	Erie K800011/CD801		
C2 C3	.01 .047	Ceramicon Polyester	100 250	-20+80 20	900067 909227	Erie K800011/CD801 Mullard C280 AE/P47K		
C4	47p	Ceramicon	750	2 ¹ / ₂	<i>JUJEE</i> 1	Erie N150 Style B		
C5	22p	Ceramicon	750	2 <u>1</u> 2 <u>1</u>	See Note	Erie NPO		
	NOTE:	The standard o	component ed may be	(NPO) ha in the E	s zero co	erature coefficient. efficient. The type. PlOO (+ve t.c.) or		
Amplif	fier Boar	rd						
Cl	1	Tantalum	35	20	909123	U. Carbide KIJ35S		
C2	.01	Polyester	250	20	910485	Mullard C280 AE/PlOK		
C3 C4	1 •047	Tantalum Polyester	35 250	20 2 0	909123 909227	U. Carbide KIJ35S Mullard C280 AE/P47K		
C5	.01	Polyester	250	20	9104 8 5	Mullard C280 AE/P47K		
сб	.01	Polyester	250	20	910485	Mullard C280 AE/Plok		
C7	•047	Polyester	250	20	909227	Mullard C280 AE/P47K		
c8	.01	Polyester	250	20	910485	Mullard C280 AE/PloK		
C9 C10	.01 .047	Polyester Polyester	250 250	20 20	910485 909227	Mullard C280 AE/PlOK Mullard C280 AE/PlOK		
0-0		Inductors		20	>->	······································		
NOTE:	To fe	not mounted on a	boord					
			a Doaru.					
L1 L2		not fitted Oscillator co:	il			Racal CT 35809		
		Transistors						
Oscili	letor							
VTl		n.p.n.		$\circ \mathbf{r}$	909113 906517 ⁰¹	Semi-conductor ST70 Texas 2N918		
Ampli	fier <u>B</u> oa	rd						
VTl		n.p.n.		(ST70	- 909113	Semi-conductors ST70		
VT2		n.p.n.			or - 906517			
VT3		n.p.n.		(511918	- 900717	Texas 2N918		

Cct. Ref.	Value	Description Re	at. Tol. %	Racal Part No.	Manufacturer					
	SECOND V.F.O. (continued)									
		Diodes								
Fine !	l'une Board	<u>l</u>								
2Dl ai	nd 202	Variable capacits	ance diodes	(15pF)	Motorola MV830					
0s cil)	lator Boar	<u>rd</u>								
D1 D2 D3 D4		Zener 7.5V ±5% Zener 7.5V ±5% Zener 6.2V ±5% Zener 6.2V ±5%		900897 900897 905 3 95 905395						
Ampli Dl D2 and	fier Board d D3	not used		908343	Texas 18920					
		Plugs and Sockets	5							
ЪΓΊ		Multi-way connect Coaxial inserts :		90 87 16 ff) 908341	Cannon DBM13W3P Cannon DM53740-5001					
Fine Tune Component Board Assembly AC39599 containing 2R1, 2C1, 2D1 and 2D2										

Cet. R ef .	Value	Description	Rat.	Тоl. %	Racal Part No.	Manufacturer				
	37.5 MHz GENERATOR									
	(BC.28284)									
		Resistors								
Harmon	nic Genera	ator (AC.32854)	2							
	ohms									
Rl	1.8k	Metal Oxide		5	908283	Electrosil TR4				
R2	1.2k	Metal Oxide		5	908285	Electrosil TR4				
R3	820	Metal Oxide		5 5 5 5 5	908282	Electrosil TR4				
R4	390	Metal Oxide		5	908472	Electrosil TR4				
R 5	390	Metal Oxide		5	908472	Electrosil TR4				
Harmo	nic Mixer	(AC.37894)								
Rl	270	Metal Oxide		5	908284	Electrosil TR4				
R2	56	Metal Oxide			908289	Electrosil TR4				
R3	15k	Metal Oxide		5 5 5 5	908280	Electrosil TR4				
R4	5.6k	Metal Oxide		5	908273	Electrosil TR4				
R5	120	Metal Oxide		5	908286	Electrosil TR4				
r6	6.8k	Metal Oxide		5	900987	Electrosil TR4				
R7	5.6k	Metal Oxide		5 5 5 5 5	908273	Electrosil TR4				
RŚ	15k	Metal Oxide		5	908280	Electrosil TR4				
R9	120	Metal Oxide		5	908286	Electrosil TR4				
R10	8 20	Metal Oxide		5	908282	Electrosil TR4				
37.5	MHz Amplif	fier (AC.32850)							
Rl	4.7k	Metal Oxide	_	5	900989	Electrosil TR4				
R2	4.7k	Metal Oxide		2	900909 900989	Electrosil TR4				
R3	47k	Metal Oxide		5	908391	Electrosil TR4				
R4	680	Metal Oxide		5	908390	Electrosil TR4				
R5	12k	Metal Oxide		5 5 5 5 5	908274	Electrosil TR4				
Rб	8.2k	Metal Oxide		5	908275	Electrosil TR4				
		Potentiometer								
RVl	2 .2 k	Mixer Balance	adjustn	lent	909838	Ancillary Developments Type TO5				
		Capacitors								
101	0.1	Polyester	250	10	909847	Mullard C280 AE/Alook				
102	0.1	Polyester	250	10	909847	Mullard C280 AE/AlooK				
103	150p	Polystyrene	30	21/2	908331	Suflex HS7/A				
						-				

<u>Harmo</u>						
<u>Harmo</u>		37.5	MHz GENE	ERATOR (CON	ntinued)	
	nic Gener	ator (AC.32854	2			
Cl		Trimmer 4.5/1	volts 5p		908796	Steatite N750 Type 7S Triko O2
C 2	-047	Polyester	250	20	909227	Mullard C280 AE/P47K
C3	.047	Polyester	250	20	909227	Mullard C280 AE/P47K
C4	.01	Polyester	250	20	910485	Mullard C280 AE/PlOK
C5	100p	Polystyrene	30	2 <u>1</u>	908797	Suflex HS7/A
Harmo	nic Filte	r (AC.32862 an	d AC.325	548)		
Cl	22p	Polystyrene		±1p	906703	Suflex HS7/A
C2	2.5-6p	Pre-set Ceram	te sub-m		907886	Steatite Triko 7SON033
C3	27p	Polystyrene		±lp	908325	Suflex HS7/A
C4	7 -3 5p	Pre-set Ceram	ic sub-m	-	908806	Steatite Triko 7802N150
C5	27p	Polystyrene		±lp	908325	Suflex HS7/A
	-12	20-90091000		P	<i><i>yoeyeyey</i></i>	
C6	2.5-6p	Pre-set Ceram	ic sub-n	nicro	907886	Steatite Triko 7SO2NO33
C7	27p	Polystyrene		±1p	908325	Suflex HS7/A
c8	7 -3 5p	Pre-set Ceram	ic sub-m	-	908806	Steatite Triko 7S02N150
C9	18p	Polystyrene		±1p	907171	Suflex HS7/A
	nic Mixer				_	
Cl	0.1	Polyester	250	10	909847	Mullard C280 AE/AlOOK
C 2	0.1	Polyester	250	10	909847	Mullard C280 AE/AlOOK
C3	.01	Ceramic	100	-20+8 0	909102	Erie K800011 CD801
C4	.01	Ceramic	100	-20+80	909102	Erie K800011 CD801
C5	.01	Ceramic	100	-20+80	909102	Erie K800011 CD801
c 6	.01	Ceramic	100	-20+80	909102	Erie K800011 CD801
<u>37.5</u>	MHz Ampli	fier (AC.32850)			
Cl	0.1	Polyester	250	10	909847	Mullard C280 AE/AlOOK
C 2	0.1	Polyester	250	10	909847	Mullard C280 AE/A100K
с3	.047	Polyester	250	20	909227	Mullard C280 AE/P47K
C4	68p	Polystyrene	30	2 1	908321	Suflex HS7/A
C5	0.1	Polyester	250	10	909847	Mullard C280 AE/Plook
c 6	.047	Polyester	250	20	909227	Mullard C280 AE/P47K
C7	15p	Ceramic	750	5	902047	Lemco 310NPO
		<u>T</u> ransformers	and Indi	actors		
Harmo	nic Filte	r (AC.32862 an	d AC.325	<u>548)</u>		
Ll		Coil Assembly	. <u> </u>		CT 32956	}
12		Coil Assembly			CT 32954	
		Coil Assembly			CT 32956	
13 14		Coil Assembly			CT 32954	
<u>.</u> .		JOIT AGGEMDIN				

Cct. Value Ref.	Description Rat. Tol. %	Racal Part No.	Manufacturer
	37.5 MHz GENERATOR (con	tinued)	
<u>Harmonic Mixer</u>	(AC.37894)		
ГJ	Coil Assembly	ст 32957	
37.5 MHz Amplii	fier (AC.32850)		
Tl	Coil assembly inc. capacitor C4		Racal CT 28317
	Transistors		
Harmonic Genera	ator		
VTI	p.n.p.	910866	Texas GM290A
Harmonic Mixer			
VT1 and VT2	p.n.p.	910 8 66	Texas GM290A
37.5 MHz Amplif	fier		
VTL	p.n.p.	910866	Texas GM290A
VT2	p.n.p.	910866	Texas GM290A
	Diodes		
Harmonic Genera	ator		
D1 D2		90 8347 908347	Hughes HG 5085 Hughes HG 5085
37.5 MHz Amplif	lier		
Dl and D2		908347	Hughes HG 5085
	Plugs and Sockets		
PL1	Connects to bulkhead adaptor	908370	Belling Lee L1403/RFP/Ag
SKTI	Connects to PL1 on 37.5 MHz Filter Unit	907076	Cannon insert DM53742-5001
	Ferrite Beads		
FBl	On H.T. Filter	900461	Mullard FX 1115
	<u>37.5 MHz Filter</u>		
	unit is faulty a replacement sho Department quoting Part number B		tained from the Racal
	<u>Plugs (fixed)</u>		
PL1 PL2	Coaxial insert Coaxial insert	908341 908341	Cannon DM 53740-5001 Cannon DM 53740-5001

Ll	Coaxial insert	908341	Cannon DM 53740-5001
12	Coaxial insert	908 3 41	Cannon DM 53740-5001

Cct.	Value	Description	Rat.	Tol.	Racal	Manufacturer
Ref.				%	Part No.	

1 MHz AMP. OSC. & CALIBRATOR

(CC 28285)

Note: This assembly is contained in the 37.5 MHz Generator Module.

Resistors

		<u> </u>				
	ohms		watts			
lRl	56	Metal Oxide		5	910545	Welwyn F25
1R2	22k	Composition	0.1	10	902524	Erie 15
	Amplifier				-	
				_		
Rl	22k	Metal Oxide		5 5 5 5	908269	Electrosil TR4
R2	6.8k	Meta] Oxide		5	900987	Electrosil TR4
R3	lk	Metal Oxide		5	908267	Electrosil TR4
R4	8 20	Metal Oxide		5	9082 8 2	Electrosil TR4
R5	10	Composition	0.1	10	902484	Erie 15
		-			-	
R6	lk	Metal Oxide		5	90 8 267	Electrosil TR4
R7	8.2k	Metal Oxide		5 5	908275	Electrosil TR4
r8	6.8k	Metal Oxide		5	900987	Electrosil TR4
R9		Composition	0.1	10	902490	Erie 15
RlO	33 560	Metal Oxide	0.1	5	909841	Electrosil TR4
NTO.	,00	HE DAT UNICE		/	909041	FICCOLORIT 114
R11	33	Metal Oxide		5	908690	Welwyn F25
)) lk) F		
R12		Metal Oxide		2	908267	Electrosil TR4
R13	18k	Metal Oxide		2	900994	Electrosil TR4
R14	5.6k	Metal Oxide		5 5 5 5 5	908273	Electrosil TR4
R15	1.5k	Metal Oxide		5	908296	Electrosil TR4
_						
R16	1.2k	Metal Oxide		5	908285	Electrosil TR4
R17	39k 82	Metal Oxide		5	908292	Electrosil TR4
R18	82	Metal Oxide		5	908290	Electrosil TR4
R19	68	Metal Oxide		5 5 5 5	908278	Electrosil TR4
				-	2 .	
<u>l M</u> Hz	Oscillato	r and Calibraton	r Board	(BC.3286	0)	
Rl	39k	Metal Oxide		5	908292	Electrosil TR4
R2	lOk	Metal Oxide		5 5 5 5 5	900986	Electros il TR4
R3	4.7k	Metal Oxide		5	900989	Electrosil TR4
R4	100	Metal Oxide		5	908276	Electrosil TR4
R5	6.8k	Metal Oxide		5	900270	Electrosil TR4
лJ	0.04	MCORT OXIDE)	900901	FIGGEOSIT 1V4
Rб	2.2k	Metal Oxide		5	908270	Electrosil TR4
	68			5 5 5 5 5	908270 908278	Electrosil TR4
R7		Metal Oxide		2		
r8 FO	4.7k	Metal Oxide		2	900989	Electrosil TR4
R9	2.2k	Metal Oxide		2	906020	Electrosil TR4
RIO	lOk	Metal Oxide		5	906023	Electrosil TR4

Cct. <u>Ref.</u>	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer
		1 MHz AMP	. OSC.	& CAL.	(continued)	
R11 R12	ohms 4.7k 47	Metal Oxide Composition	watts	5 10	900989 902492	Electrosil TR4 Morganite XL
R13 R14	4.7k 39k	Metal Oxide Metal Oxide		5 5 5 5	900989 910107	Electrosil TR4 Electrosil TR4
R15 R16	150k 4.7k	Metal Oxide Metal Oxide		2 5	908470 900989	Electrosil TR4 Electrosil TR4
101	0.22	<u>Capacitors</u>			908338	TCC CML10
<u>lMHz Ar</u>	mplifier 1	Board (BC.32858)		_		
C1 C2 C3 C4 C5	.047 .047 .047 .047 .047	Polyester Polyester Polyester Polyester Polyester	<u>volt</u> 250 250 250 250 250	20 20 20 20 20 20	9.09227 909227 909227 909227 909227 909227	Mullard C280 AE/P47K Mullard C280 AE/P47K Mullard C280 AE/P47K Mullard C280 AE/P47K Mullard C280 AE/P47K
C6 C7 C8 C9 C10 * C10 *	0.1 .047 .001 .047 680p 0.1	Polyester Polyester Ceramic Hi-K Polyester Polystyrene Polyester	250 250 350 250 30 250	10 20 20 2 <u>1</u> 10	909847 909227 902122 909227 908455 909847	Mullard C280 AE/AlOOK Mullard C280 AE/P47K Lemco 310K Mullard C280 AE/P47K Suflex HS7/A (See NOTE) Mullard C280 AE/AlOOK
* NOTE	:- In late	er production if		type 2	N1396 then Cl	•
C11 C12 C13 C14	150p .0015 .047 0.1	Polystyrene Ceramic Hi-K Polyester Polyester	30 350 250 250	21 20 20 10	908331 902124 909227 909847	Suflex HS7/A Lemco 310K Mullard C280 AE/P47K Mullard C280 AE/A100K
		and Calibrator				
C1 C2 C3	.047 .047 270p	Polyester Polyester Polystyrene	250 250 30	20 21 21 21 21 21 21 21 21 21 21 21 21 21	909227 909227	Mullard C280 AE/P47K Mullard C280 AE/P47K Suflex HS7/A
C4 C5	470p .001	Polystyrene Polystyrene	30 30	25	908317 908583	Suflex HS7/A Suflex HS7/A
c6 C7	22p 7-35p	Polystyrene Pre-set: Steat	30 ite Ce:	lp ramic	906703	Suflex HS7/A
C8 C9 C10	0.1 0.1 560	Sub-M Polyester Polyester Polystyrene	Licro 7 250 250 30	10 10 2 ¹ /2	908806 909847 909847	Triko O2 N1500 Mullard C280 AE/AlOOK Mullard C280 AE/AlOOK Suflex HS7/B
C11 C12 C13 C14 C15 C16	0.1 .0015 0.1 560 0.1 0.1	Polyester Polystyrene Polyester Polystyrene Polyester Polyester	250 30 250 30 250 250	10 2 1 10 20 10 10	909847 910932 909847 908452 909847 909847	Mullard C280 AE/AlOOK Suflex HS7/B Mullard C280 AE/AlOOK Suflex HS7/A Mullard C280 AE/AlOOK Mullard C280 AE/AlOOK

Cct <u>Ref.</u>	Value	Description	Rat.	Tol %	Racal Part No.	Manufacturer
		<u>l Milz</u>	AMP. OSC	. & CA	L. (continue	ed)
		Transformers	and Indu	ctors		
1MHz Am	plifier H	Board				
L1		Coil Assembly			CT 32955	
1MH7 OB	c. and Ca	librator Board				
Ll		Coil Assembly			CT 31443	
L2		Coil Assembly			ст 3144 2	
L3		Coil Assembly			CT 31444	
		Transistors				
l MHz Am	plifier <u>I</u>	Board				
VTl		p.n.p.			909414	Motorola 2N3323
VT2		p.n.p.			909414	Motorola 2N3323
VT3		p.n.p. (earlie	r version	ns onl	n) 909414	Motorola 2N3323
	un and (p.n.p. (fitted Calibrator Board		r vers	lons)	Amperex 2N1396
	sc. and (······································	-			
VT1 VT2		p.n.p.			909414	Motorola 2N3323
VTZ VT3		p.n.p. n.n.n			909414 909414	Motorola 2N3323 Motorola 2N3323
V1)		p.n.p.			909414	Motorola 2N3323
		Diodes				
<u>1 MHz A</u>	mplifier	Board				
Dl		Semi-conductor			900620	Mullard OA200
<u>l MHz O</u>	sc. and (Calibrator Board	<u>l</u>			
Dl to D	4	Semi-conductor	•		900620	Mullard QA200
D5		See NOTE below	r		900620	Mullard QA200
D6		Semi-conductor	1		900652	Mullard AA213
NOTE:-	In later	production diod	le D5 is	replac	ed by resis	tor R16
		Plugs and Socl	<u>kets</u>			
PLl		(Plug shell	4 • •		908388	Cannon DBM9W4P
		(Plug inserts	(4)		907080	Cannon D M53741-5001
		Crystals				
XLI		l MHz crystal	•		CD 38871/A	
		Crystal Holden	•		900397	X2/UG
		-	al XLl is t for eme		-	the RA.1218

Cet. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer			
	<u>B.F.O.</u>								
Resistors									
600 kHz Oscillator Board (BC.30540)									
R1 R2 R3 R4 R5	ohms 220 5.6k 5.6k 5.6k 10k	Metal Oxide Metal Oxide Metal Oxide Metal Oxide Metal Oxide		5 5 5 5 5 5	900988 908273 908273 908273 908273 900986	Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4			
Ampli	fier Board	d (BC.30542)							
R1 R2 R3 R4 R5	5.6k 15k 1k 180 4.7k	Metal Oxide Metal Oxide Metal Oxide Metal Oxide (part of CT 3	5217)	5 5 5 5 10	908273 908280 908267 909125 908246	Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4 Nutec RKL10			
	Capacitors								
<u>B.F.O</u>	. Assembly	<u>У</u>	volts						
201 102	0.1 4-20p	Polyester B.F.O. Variab	250	l0 g	909847 ad 38584	Mullard C280 AE/AlOOK			
600 ki	Hz Oscill	ator Board (BC	30540)						
C1 C2 C3 C4	0.1 7-35p 7-35p	Polyester pre-set pre-set not used	250	10	909428 908806 908806	Mullard C280 AE/PlOOK Steatite 7S Triko O2/N1500 Steatite 7S Triko O2/N1500			
C5	7-35p	pre-set			90 8 806	Steatite 75 Triko 02/N1500			
C6 C7 C8 C9 C10	7-35p 390p 82p 0.1 .01	pre-set Polystyrene Ceramic Polyester Polyester	30 750 250 250	212 2 20 20	908806 908243 902099 909428 910485	Steatite 7S Triko O2/N15OO Suflex HS7/A Erie N75OB Mullard C28O AE/P1OOK Mullard C28O AE/P1OK			
CII	- 047	Polyester	250	20	909227	Mullard C280 AE/P47K			
Ampli:	fier Board	d (BC.30542)							
C1 C2* C3	.0 ¹ 47 180p 0.1	Polyester Polystyrene Polyester f transformer a	volts 250 30 250	20 2 ¹ / ₂ 10	909227 90 788 4 909847	Mullard C280 AE/P47K Cuflex HS7/A Mullard C280 AE/AlOOK			
	TO PELO C.	a viensiormen d	ouromos,y	94 J/CAU					

Cct. Value Ref.	<u> </u>	01. \$	Racal Manufacturer Part No.							
	B.F.O. (continued)									
	Transformers and Inducto	rB								
600 kHz Oscillator Board										
Ll	Oscillator coil		CT 35217							
Amplifier Boa	Amplifier Board									
Tl	600 kHz output transform	er	CT 35216							
SB	<u>Switches</u> DET-B.F.O. Mode switch		BSW 38585							
	Transistors									
600 kHz Oscil	lator Board									
₩ I	n.p.n.		906433 S.T.C. BSY95A							
Amplifier Boa	rd									
VTI	n.p.n.		906433 S.T.C. BSY95A							

Cct. Ref.	Value	Description	Rat.	То]. %	Racal Part No.	Manufacturer				
	1st I.F. AMPLIFIER UNIT									
	(BC 31474)									
	NOTE:	This unit is a small module located adjacent to the crystal bandwidth filters (Fig. 18) The circuit is shown in Fig. 11.								
		Resistors								
R1 R2 R 3 R4 R5	ohms 18k 100 4.7k 3.9k 1k	Metal Oxide Metal Oxide Metal Oxide Metal Oxide Metal Oxide		% 5 5 5 5 5 5	908272 908276 900989 900990 908267	Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4				
R6 R7	470 10	Metal Oxide Composition	0.1	5 10	900992 902484	Electrosil TR4 Erie 15				
		Capacitors								
C1 C2 C3* C4 C5*	0.1 .047 180p .047 .01	Polyester Polyester Polystyrene Polyester Silver Mica	volts 250 250 30 250 500	10 20 2 2 20 20	909847 909227 907884 909227 908245	Mullard C280 AE/AlOOK Mullaed C280 AE/P47K Suflex HS7/A Mullard C280 AE/P47K Erie Microcap				
C6*	.01	Silver Mica	50 0	20	908245	Erie Microcap				
* C	ontained	in L l ass embly								
LI		Inductors Coil Assembly			CT 31472	2				
VTl		<u>Transistor</u> p.n.p.			90941 ⁾ 4	Motorola 2N3323				

				Tol. %	Racal Part No.	Manufacturer
			MAIN I	.F. MODUI	<u>LE</u>	
		Resistors				
_	ohms			% 5	<u> </u>	
2R1 2R2	lk	Metal Oxide not used		5	908267	Electrosil TR4
2R3	120	Metal Oxide		5	9082 8 6	Electrosil TR4
NOTE:	2R1 and	2R3 are not mo	unted on	a board.	•	
H.T. St	upply Fi	lter (AC 30535)			
Rl	100	Metal Oxide	4	5	908276	Electrosil TR4
τ ም Δτ	mplifier	BORNA (BC 305	33)			
<u>1.r. A</u>	18k	Board (BC.305 Metal Oxide		5	908272	Flootnootl mph
R1 R2				5		Electrosil TR4
	3.9k	Metal Oxide		5	900990	Electrosil TR4
R3	330	Metal Oxide		5	908268	Electrosil TR4
R4	4.7k	Carbon Film		10	908246	Nutec RKL10
R5	33	Metal Oxide		5	908690	Welwyn F25
r6	lk	Metal Oxide		5	908267	Electrosil TR4
R7	22k	Metal Oxide		5	908269	Electrosil TR4
r8	4.7k	Metal Oxide		5 5 5	900989	Electrosil TR4
R9	2.2k	Metal Oxide		5	908270	Electrosil TR4
RlO	5.6k	Metal Oxide		5	908273	Electrosil TR4
Rll	4.7k	Metal Oxide		5	900989	Electrosil TR4
R12	3.9k	Metal Oxide		5	900990	Electrosil TR4
R13	10k	Metal Oxide		Ś	900986	Electrosil TR4
R14	220	Metal Oxide		5	900988	Electrosil TR4
R15	8.2k	Metal Oxide		5 5 5 5 5 5 5	908275	Electrosil TR4
R16	4.7k	Carbon film		10	908246	Nutec RKL10
R17	3.9k	Metal Oxide			900990	Electrosil TR4
R18	12k	Metal Oxide		5 5 5	900990 908274	Electrosil TR4
				2		
R19 R20	330 4.7k	Metal Oxide Carbon Film		5	908268 908246	Electrosil TR4 Nutec RKLlO
1120	4•1K			-	-	Marec Marto
R21	47	Metal Oxide		5 5	908298	Welwyn F25
R22	lk	Metal Oxide		5	908267	Electrosil TR4
Detecto	or Board	(cc.28236)				
Rl	3.9k	Metal Oxide		5	900990	Electrosil TR4
R2	33k	Metal Oxide		5	908291	Electrosil TR4
R3	39k	Metal Oxide		5	908292	Electrosil TR4
R4	5.6k	Metal Oxide			908273	Electrosil TR4
_ • •	56	Metal Oxide		5 5	908289	Electrosil TR4

Cct. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer
D-+	D	<u>I.</u>	F. MODUI	E (contin	nued)	
Ro	tor Board 100	Metal Oxide		5	908276	Electrosil TR4
R7	2.2k	Metal Oxide		5 5 5 5	908270	Electrosil TR4
R8	330	Metal Oxide		5	908268	Electrosil TR4
R 9	lOk	Metal Oxide		Ś	900986	Electrosil TR4
R 10		Not used		-		
Rll	4.7k	Metal Oxide		5	900989	Electrosil TR4
R12	820	Metal Oxide		5 5 5 5 5	908282	Electrosil TR4
R13	look	Metal Oxide		5	908293	Electrosil TR4
R14	4.7k	Metal Oxide		5	900989	Electrosil TR4
R15	lk	Metal Oxide		5	908267	Electrosil TR4
R1 6	33 6.8k	Composition	0.1	10	902490	Erie 15
R17		Carbon Film		10	908247	Nutec RKL10
R18	2.7k	Metal Oxide		5 5 5	908294	Electrosil TR4
R1 9	15k	Metal Oxide		5	908280	Electrosil TR4
R20	100	Metal Oxide		5	908276	Electrosil TR4
R21	2.2k	Metal Oxide		5	908270	Electrosil TR4
R22	5.6k	Metal Oxide		5	908273	Electrosil TR4
R23	33k	Metal Oxide		5 5 5 5	908291	Electrosil TR4
R24	3.9k	Metal Oxide		5	900990	Electrosil TR4
R25	56k	Metal Oxide	ă.	5	908287	Electrosil TR4
R26	1.2k	Metal Oxide		5	908285	Electrosil TR4
R27	6.8k	Metal Oxide		5 5	900987	Electrosil TR4
R28	22k	Metal Oxide		5	908269	Electrosil TR4
R29 R30		Not used Not used				
NJU	_				_	
R 31	82	Metal Oxide		5 5	908290	Electrosil TR4
R32	5.6k	Metal Oxide			908273	Electrosil TR4
R33	5.6k	Metal Oxide		5	908273	Electrosil TR4
AGC B	oard (BC	31466/в)				
Rl	15k	Metal Oxide		5	908280	Electrosil TR4
R2	3.9k	Metal Oxide		5 5 5	900990	Electrosil TR4
R3	330	Metal Oxide		5	908268	Electrosil TR4
R4	5.6k	Carbon Film		10	-	Nutec RKL10
R5	22	Composition	0.1	10	902488	Erie 15
R6	lk	Metal Oxide		5	908267	Electrosil TR4
R 7	18k	Metal Oxide		5 5 5	908272	Electrosil TR4
r8	12k	Metal Oxide		5	908274	Electrosil TR4
R9	lk	Metal Oxide		5	908267	Electrosil TR4
RIO	18	Composition	0.1	10	902487	Erie 15

Cct. Ref.	Value	Description	Rat.	то]. %	Racal Part No.	Manufacturer
		<u>I.</u>	F. MODULE	(cont	inued)	
	Board					
RII	lOk	Carbon Film		10	908249	Nutec RKL10
R12	150k	Metal Oxide		5	908277	Electrosil TR4
R13	120k	Metal Oxide		5	908281	Electrosil TR4
R14	lOk	Metal Oxide		5 5 5 5	900986	Electrosil TR4
R15	lOk	Metal Oxide		5	900986	Electrosil TR4
R16	120k	Metal Oxide		5 5	908281	Electrosil TR4
R17	68k	Metal Oxide		5	908279	Electrosil TR4
81R	820k	Composition	0.1	10	90254 3	Erie 15
R19	lk	Metal Oxide		5	908267	Electrosil TR4
R20	2 .2 k	Metal Oxide		5 5	908270	Electrosil TR4
R21	6.8k	Metal Oxide		5	900987	Electrosil TR4
R22	15k	Metal Oxide		5 5 5 5 5 5	908280	Electrosil TR4
R23	820	Metal Oxide		5	908282	Electrosil TR4
R24	3.3k	Metal Oxide		5	900991	Electrosil TR4
R25	470	Metal Oxide		5	900992	Electrosil TR4
	-		o - 21.002 /11	•		
			<u>sc 34783/v</u>		op Board)	
Rl	6.8k	Metal Oxide		5 5 5 5 5 5	900987	Electrosil TR4
R2	1.8k	Metal Oxide		2	908283	Electrosil TR4
R3	100	Metal Oxide		5	908276	Electrosil TR4
R4	270	Metal Oxide		5	908284	Electrosil TR4
R5	lk	Metal Oxide		5	908267	Elect r osil TR4
r6	100	Metal Oxide		5 5 5 5 5 5	908276	Electrosil TR4
R7	15k	Metal Oxide		5	908280	Electrosil TR4
r8	100	Metal Oxide		5	908276	Electrosil TR4
R9	1.2k	Metal Oxide		5	908285	Electrosil TR4
R10*	56k	Metal Oxide		5	908287	Electrosil TR4
R11	120	Metal Oxide		5	908286	Electrosil TR4
* R	10 is fit	tted only to the	e 100 kHz	board	3 4783/B	
Conve	rter Osci	illator Board (BC 38568/	VAR) ((Bottom Board	<u>1)</u>
NOTE:		for R12, resist 5 kHz (A) versio		s are	identical in	the lOO kHz (B)
	ohms					
Rl	100	Metal Oxide		5	908276	Electrosil TR4
R2		Metal Oxide		5 5 5 8	908279	Electrosil TR4
R3		Metal Oxide		5	908472	
R4		Metal Oxide		8		Electrosil TR4
R5		Metal Oxide		5	908269	Electrosil TR4
***		neval United		1	<i>,</i>	III -

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Cct. Ref.	Value	Description	Rat.	Tol. ¢	Racal Part No.	Manufacturer
		Ι.]	F. MODULE	Continu	led)	
R6 R7 R8 R9 R10	1k 3.3k 18k 100 1k	Metal Oxide Metal Oxide Metal Oxide Metal Oxide Metal Oxide		5 5 5 5 5	908267 900991 908272 908276 908267	Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4 Electrosil TR4
R11 R12 R13	33 39k 82k	Metal Oxide (100 kHz only (455 kHz only		5 5 5	908690 908292 908691	Welwyn F25 Electro si l TR4 Electrosil TR4
AGC B	oard	Potentiometer				
נעא	2.2M				908365	Plessey Type MP
<u>Modul</u>	e	Capacitors	volts			
C1 C2	0.22	Ceramic Ceramic	50 50	-20+50 -20+50	908338 908338	T.C.C. CML10 T.C.C. CML10
<u>H.T.</u>	Supply Fi	lter (AC.30535)				
C1 C2	•047 •047	Polyester Polyester	2 50 250	20 20	909227 909227	Mullard C280 AE/P47K Mullard C280 AE/P47K
<u>I.F.</u>	Amplifier	Board (BC.3053)	<u>3)</u>			
C1 C2 C3 C4 C5	0.1 .047 10 180p 0.1	Polyester Polyester Electrolytic Polystyrene Polyester	250 250 16 30 250	10 20 -10+50 2 1 10	909847 909227 900068 907884 907884	Mullard C280 AE/A100K Mullard C280 AE/P47K Mullard C426 AR/E10 Suflex HS7/A Mullard C280 AE/A100K
C6 C7 C8	.047 0.1 180p	Polyester Polyester Polystyrene	250 250 30	20 10	909227 909847	Mullard C280 AE/P47K Mullard C280 AE/A100K
C9 C10	0.1 0.1	Polyester Polyester	250 250	2 <u>늘</u> 10 10	907884 909847 909847	Suflex HS7/A Mullard C280 AE/A100K Mullard C280 AE/A100K
C9	0.1	Polyester	250	10	909847	Suflex HS7/A Mullard C280 AE/A100K
C9 C10 C11 C12 C13 C14 C15 C16 C17	0.1 0.1 .047 0.1 180p 50 .047 .047	Polyester Polyester Polyester Polyester Polystyrene Electrolytic Polyester Polyester Polyester	250 250 250 250 30 16 250 250	10 10 20 10 2 1 -10+50 20 20	909847 909847 909847 909847 909884 908798 909227 909227	Suflex HS7/A Mullard C280 AE/AlOOK Mullard C280 AE/AlOOK Mullard C280 AE/AlOOK Suflex HS7/A Mullard C426 ARF50 Mullard C280 AE/P47K Mullard C280 AE/P47K

Cct. Ref.	Value	Description	Rat.	Тоl. <i>\$</i>	Rac a l Part No.	Manufacturer
Detect	tor Board	<u>I.</u>]	F. MODULI	E (continu	ued)	
C6 C7 C8 C9 C10	120p 10 120p .047 0.1	Polystyr en e Electrolytic Polystyrene Polyester Ceramic	30 16 30 250 30	2] -10+50 2] 20 -25+50	9083 3 2 900068 908332 909227 906675	Suflex HS7/A Mullard C426 AR/E10 Suflex HS7/A Mullard C280 AE/P47K Erie 811T/30
C11 C12 C13 C14 C15	0.1 100p .047 180p 0.1	Ceramic Polystyrene Polyester Polystyrene Ceramic	30 30 250 30 30	-25+50 2 월 20 2 월 -25+50	906675 908241 909227 907884 906675	Erie 811T/30 Suflex HS7/A Mullard C280 AE/P47K Suflex HS7/A Erie 811T/30
C16 C17 C18 C19 C20	0.1 .047 330p .01 .0022	Ceramic Polyester Silver Mica Ceramic Ceramic	30 250 350 100 350	-25+50 20 2 20	906675 909227 902173 900067 902126	Erie 811T/30 Mullard C280 AE/P47K J.M.C. CX22S/350 Erie CD801 Lemco 310K
C21 C22 C23 C24 C25	7-35p 7-35p 0.1 0.1 10p	Variable Variable Ceramic Ceramic Polystyrene	30 30 30	-25+50 -25+50 2 2	908806 908806 906675 906675 908324	Steatite Triko 02/N1500 Steatite Triko 02/N1500 Erie 811T/30 Erie 811T/30 Suflex HS7/A
C 26	lOp	Polystyrene	30	2 1 2	908324	Suflex HS7/A
AGC B	oard (BC	31466/в)				
C1 C2 C3 C4 C5	.047 0.1 .0086 330p .047	Polyester Polyester Silver Mica Polystyrene Polyester	250 2 5 0 125 3 0 250	20 10 2 2 1 20	909227 909 847 908 33 7 908242 909227	Mullard C280 AE/P47K Mullard C280 AE/AlOOK S.T.C. 454LWA-74 Suflex HS7/A Mullard AE/P47K
C6 C7 C8 C9 C10	.047 33p .047 180p 390p	Polyester Polystyrene Polyester Polystyrene Polystyrene	250 30 250 30 30	20 1p 20 2] 2] 2]	909227 906497 909227 907884 908243	Mullard C280 AE/P47K Suflex HS7/A Mullard C280 AE/P47K Suflex HS7/A Suflex HS7/A
C11 C12 C13 C14 C15	0.1 0.22 0.1 6.4 0.64	Polyester Ceramic Polyester Electrolytic Electrolytic	250 50 250 25 64	10 -10+50 10 -10+50 -10+50	909847 908338 909847 905371 909311	Mullard C280 AE/AlOOK T.C.C. CML10 Mullard C280 AE/AlOOK Mullard C426 AR/F6.4 Mullard C426 AS/H0.64
C16 C17	0.1 50	Polyester Electrolytic	250 25	10 -10+50	909847 908798	Mullard C280 AE/A100K Mullard C426 AR/F50

Cct. Ref.	Value	Description	Rat.	То]. %	Racal Part No.	Manufacturer		
I.F. MODULE (continued)								
Conve	rter Ampl	ifier Board (B	<u>c 34783/</u>	VAR) (Top	Board)			
C1 C2 C3	.0033 330p 82p	Ceramic Polystyrene Ceramic	50 30 50	10 2 ¹ /2 10	908242	Erie 8003 W5R Suflex HS7/A Erie 831 N2200		
C4 C5	0.1 470p	Polyester Ceramic	250 50	10 10 10	909847	Mullard C280 AE/A100K Erie 8003 W5R		
C6 C7 C8 C9	0.1 0.1 470p 2.2	Polyester Polyester Polystyrene Tantalum	250 250 30 20	10 10 2 1 20	909847 909847 908317 908316	Mullard C280 AE/A100K Mullard C280 AE/A100K Suflex HS7/A U. Carbide K2R2J20S		
Conve:	rter Osci	llator Board (1	BC 38568	B/VAR) (BO	ttom Board	1)		
NOTE:	Except					for the 100 kHz and		
C1 C2 C3 C4 C5	.047 6.8p 150p 7-35p 470p	Polyester Ceramic Polystyrene Variable Polystyrene	250 750 30 30	20 12 21 22 22	909227 902041 908331 908806	Mullard C280 AE/P47K Lemco 310NPO Suflex HS7/A Steatite 7S Triko O2 Suflex HS7/A		
C6 C7 C8 C9 C9 C10	0.64 0.1 470p 18p 100p 470p	Electrolytic Polyester Polystyrene (455 kHz board (100 kHz board Polystyrene	64 250 30 1)30	-10+50 10 2½ 1p 2½ 2½	909311 909847 908317 908323 908241 908317	Mullard C426 AS/HO.64 Mullard C280 AE/A100K Suflex HS7/A Suflex HS7/A Suflex HS7/A Suflex HS7/A		
I.F. /	Amplifier	Board (BC 305)	33)					
L1 L2 L3		Coil Assembly Coil Assembly Coil Assembly				Racal CT 33004 Racal CT 33005 Racal CT 33006		
Detect	tor Board	-						
T1 T2		Transformer A: Transformer A:	•		CT 32961 CT 33002			
L1 12 L3		Not used Coil Assembly Coil Assembly			CT 31473 CT 32962			
<u>A.G.C</u> Ll	. Board	Coil Assembly			CT 33008			
12		Coil Assembly			CT 33007			

Cct. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer				
	I.F. MODULE (continued)									
Conver	rter Ampli	lfier Board								
L1 L2		Coil Assembly Coil Assembly			CT 32958 CT 34763					
Conver	rter Oscil	llator Board								
Ll		Coil Assembly			CT 32959					
Ll		Coil Assembly	(100) kHz)	CT 34764					
12		Coil Assembly			CT 32960					
12		Coil Assembly	(100 kHz)	CT 34765					
		Transistors								
<u>I.F.</u>	mplifier	Board								
VTl		p.n.p.			909414	Motorola 2N3323				
VT2		p.n.p.			906370	Texas 2N2412				
VT3		p.n.p.			909414	Motorola 2N3323				
VT4		p.n.p			909414	Motorola 2N3323				
Detect	or Board									
VTl	-	n.p.n.			906433	S.T.C. BSY95A				
VT2		n.p.n.			900656	Texas 2S733				
VT3		n.p.n.			906433	S.T.C. BSY95A				
VT4		n.p.n.				S.T.C. BSY95A				
VT5		n.p.n.			906433	S.T.C. BSY95A				
VT6		n.p.n.			906433	S.T.C. BSY95A				
AGC BC	ard									
VT1		p.n.p.			909414	Motorola 2N3323				
VT2		n.p.n.			906433	S.T.C. BSY95A				
VT3		n.p.n.			910839	Texas 2N3707				
vr4		n.p.n.			909413	S.T.C. 2N930				
VT5		p.n.p.			909414	Motorola 2N3323				
VT6		n.p n.			908361	Texas 2N1304				
Conver	ter Ampli	fier Board								
VT1					906433	S.T.C. BSY95A				
VT2		n.p.n.			9004JJ 906006	Mullard BC107				
VT3		n.p.n.			906006	Mullard BC107 Mullard BC107				
		n.p.n.				Martala DOTO				
	ter Oscil	lator Board								
VII		p.n.p.			909414	Motorola 2N3323				
VT2		n.p.n.			906433	S.T.C. BSY95A				

Cct. Ref.	Value	Description	Rat.	Tol. %		Racal Part No.	Manufacturer
	<u> </u>	<u>I.</u>	F. MODUI	E (contin	nue	ed)	
		Diodes					
<u>Main I</u>	F. Ampl	ifier					
Dl						908343	Texas 1S920
Detecto	or Board						
Dl						908343	Texas 15920
D2						908343	Texas 18920
D3		Not used			(900652	Mullard AAZ13 or
D4					(908349	Hughes HD1871
D5		Zener				908344	International MZ13T5
D6						908343	Texas 1S920
D7						908343	Texas 18920
AGC BOE	rd						
Dl						906720	Texas 1844
D2						908343	Texas 15920
D3 D4		7				90 8343 908344	Texas 15920
D4		Zener				900544	International MZ13T5
		Plugs and Soc	kets				
	it Modu	le Connectors				(0-0	
SKT4 PL3		Coax. Fixed I Plug, free, t				906878 908 3 70	Belling Lee L1403/CS/A
PL1		Main 37-way c				908970 908674	Belling Lee L1403/RFP/ Cannon DCF37P
1SKT1		Free 37-way c				908603	Cannon DCM37S
I.F. Co	nverter	Panel					
SKT1		Coaxial: fix	ed			906878	Belling Lee L1403CS/Ag
SKT2		Coaxial: fix	ed			906878	Belling Lee L1403CS/Ag
SKT3		Coaxial: fix	ed			906878	Belling Lee L1403CS/Ag
		Crystals					
Detecto	or Board						
XLI		1601.50 kHz				CD38871/	
XI5		1598.50 kHz				CD38871/	0
Convert	er Osci	llator Board (8)				
XLI		1145 kHz in 4				CD38871/1	
XL2		1500 kHz in 1	OO kHz C	onverter		съз8871/:	F
Termina	1 Strip						
TBl		5-way - rear	panel.			909928	Carr. Fastener
							44-79-593-5м

Cct. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer				
	POWER UNIT TYPE PU.1153 (AC/DC SUPPLY)									
			(BC	38640)						
Compon	Component Board Assembly BA37549									
	Components									
NOTE:	NOTE: Components pre-fixed 'l' are located on the chassis assembly. Those without prefix are on the component board.									
	Resistors									
1R1 R1 R2 R3 R4 R5 R6	100k 1k 1.8k 2.2k 2.7k 2.7k 1.2k	Metal Oxide Metal Oxide Metal Oxide Metal Oxide Metal Oxide Metal Oxide Metal Oxide		5 5 5 5 5 5 5 5 5 5	907866 906031 906026 906020 906347 906347 906346	Electrosil TR5 Electrosil TR5 Electrosil TR5 Electrosil TR5 Electrosil TR5 Electrosil TR5 Electrosil TR5				
	Potentiometers									
1RV1	1.5k				908689	Colvern 1106/98				
1C1 1C2 1C3 1C4 1C5 C1 C2	μF .02 .02 500 500 100 100 0.1	Capacitors Paper Paper Electrolytic Electrolytic Electrolytic Electrolytic Polyester	volts 350 350 64 64 50 6.4 250	20 20 -10+50 -10+50 -10+50 20	902279 902279 906759 906759 900506 911691 909428	Dubilier Minicap G15 Dubilier Minicap G15 Mullard C431 BR/H500 Mullard C431 BR/H500 Hunts MEFC43AT Mullard C426 AR/C100 Mullard C280 AE/P100K				
1T1	Transformer1T1Mains power transformerUSA. D - 01650									
IVTI VTI VT2	Tl p.n.p.				906 3 71 911565 911565	RCA 2N3055 Mullard BCY40 Mullard BCY40				
1D1 D1 D2 D3 D4		Diodes Zener: 18 volt Encapsulated n Encapsulated n Zener: 6.8V Zener: 4.7V	rectifier		911 123 909020 909020 911120 909717	Mullard OAZ234 Motorola MDA 920/3 Motorola MDA 920/3 Hughes HS2068A Mullard OAZ240				

Cct. Ref.	Value	Description	Rat.	Tol. %	Racal Part No.	Manufacturer
<u></u>			PU.1153	(continue	ed)	
		Fuselinks				
lFS1 lFS2 Fuseholders		MAINS 250 mA anti-surge HT. 1 amp. anti-surge			911700 912052 900412	Beswick TDC134 Beswick TDC134 Belling Lee L575
		Switches				
1SA 1SB		Slider: panel Slider: panel			912063 912063	E.M.I. Type S5 E.M.I. Type S5
		Connectors				
1SKT1		Socket: sub-m	inature	15 - way	908683	Cannon DAMF. 15.S
PLl		Plug: fixed 3	-pin, bi	.8 88	900011	Plessey Mk4 CZ63953/5
Socket	t, free,	for power conne	ection to	PII	905151	Plessey Mk4 20283283/5
Outlet	t Accesso	ry Set for free socket			905154	Plessey 508/1/03008/205









NOTE: THIS BOAND IS CONTAINED IN THE Rectangular box

NOTE: THIS BOARD IS ATTACKED TO THE UNDERSIDE OF THE 1xt.VFQ CHASSIS



OSCILLATOR (B.A. 35195)



BUFFER AMPLIFIER (B.C. 32535)



BUFFER AMPLIFIER (B.C. 28128)



201722



Component Layouts: 37.5 MHz Generatar and 37.5 MHz Filter Unit

2 Ni / 23



Ist MIXER (BC.2B215)



40 MHz FILTER (AC.28197)





NOTE: CAPACITORS C8, C11, C13, C14, C14, C17 AND C18 ARE WIRED ON REAR OF BDARD



OSCILLATOR BOARD (BC.35808)



AMPLIFIER BOARD (BC.35807)

Fig. L-10

Component Layouts: 2 nd V.F.O



Ist LE AMPLIFIER UNIT

Component Layout: 3rd Mixer Module and 1st I.E. Amplifier Unit



IST LE AMPLIFIER BOARD



MAIN IFAMPLIFIER BOARD



AUDIO AMPLIFIER



DETECTOR BOARD



A.G.C. BOARD

Component Layout: IF Unit (bc.2825)

Fig. L-12



600KHz Oscillator (BC30540)

Component Layout B.F.O.





R16-VT4 \bigcirc \bigcirc T 3 \odot \bigcirc T 2 VT 2 VT3 R6 \bigcirc \bigcirc T 1 R4 Г **C** 1 VTI 4

REAR

Fig. L-14a

Component Layout 10mW Audio Amplifier Board





Component Layout: Audio Amplifier Board (1 Watt)





Wire Number	Type or Colour	From	То	Wire Number	Type or Colour	From	То
1	Coaxial	SKT12	SKT10 (A1)	41	Violet (orange sleeve)	Audio Amp.(6)	SKT8 (1/2)
2	Coaxial	SKT13	SKT10 (A3)	42	Violet (orange sleeve)	SKT8 (1/2)	SKT 10 (8)
3	Coaxial	SKT15	SKT7 (A1)	43	Violet (yellow sleeve)	PL2 (4)	SKT10 (3)
4	Coaxial	SKT16	SKT9 (1)	44	Violet	Control Board (R2)	SE2R (5)
5	Coaxial	SKT17	SKT9 (A3)	45	Violet	ISD	SE2R (5)
6	Coaxial	SKT17	SKT1 (9)	46	Violet	1SD	R7
7	Coaxial	SKT7 (A4)	SKT8 (A4)	47	Violet	R7	SKT9 (3)
8	Coaxial	SKT8 (A1)	FL1	48	Violet (blue sleeve)	SKT9/5	H.T. Filter Board
9	Coaxial	SKT8 (A3)	SKT9 (A2)	49	Violet	SKT9/5	SKT1 (17)
10	Coaxial	SKT8 (A2)	SKT10 (A3)	50	Violet (red sleeve)	SKT 1/2	lst I.F. Amp.(2)
11	Coaxial	SKT7 (A3)	SKT4	51	Violet (red sleeve)	lst I.F. Amp.(2)	PL2 (2)
12	Coaxial	SKT7 (A2)	SKT6	52	Violet (black sleeve)	SE2R(10/11)	SKT9 (4)
13	Coaxial	SKT9 (A1)	SKT1	53	Violet (blue sleeve)	15 B	SKT1 (17)
14	Coaxial	SKT9 (A4)	PL3	54	Violet	SKT9 (3)	PL2 (1)
15	Screened	1SA	SKT1 (29)	55	Black/Orange	SE1R (5)	SKTI (16)
16	Screened	RF.AGC	1SA	56	Black/Green	SE1R (6)	2RV2
17	Screened	1SA	SKT1 (28)	57	Green/Grey	ISB	SE2R (6-9)
18	Screened	1SA	SKT11 (L)	58	Black/White	SKT1 (15)	SE1R (7-11)
19	Screened	SKT1 (26)	SKT11 (H)	59	Grey/White	SE3F	SB1F (9)
20	Screened	SKT1 (24)	SE1F (4)	60	Blue/Orange	Control Board D3-D4	Meter Switch SF
21	Screened	SKT1 (34)	ISA	61	Not Used		
22	Screened	SKT11 (21)	SA2F (1/2)	62	Red/Blue	Control Board D1-D2	Meter Switch SF
23	Screened	SA2F 1/2	SE1F (3)	63	Blue/White	Control Board (R3)	Meter Switch SF
24	Coaxial	SA2R 1-5	SEIF (2)	64	Blue/White	SKT11 (K)	ISB
25	Screened	SA2F 1-3	5KT1 (20)	65	Blue/White	ISB	SKTI (18)
26	Screened	SF (RF)	SKT1 (35)	66	Red/Green	SKT10 (5)	ISD
27	Screened	ILPl	ILP2	67	Green/White	SKT10 (4)	1SD
28	Screened	ILP1	1LP3	68	Green/Blue	SKT10 (1)	lRV3 (Slider)
29	Screened	IRV2	PL2	69	Red/White	SKT11 (G)	R.F. Module
30	Screened	ILP2	PL2 (10)	70	Violet (white sleeve)	SKTI1 (E)	Audio Amp. (6)
31	Screened	ZRV3	Audio Amp. (1)	71	Red/Orange	SKT1 (5)	SBIR (7)
32	Screened	2RV1	Audio Amp. (7)	72	Green/Blue	SKT1(6)	B. F. O. Amp. (1)
33	Screened	JK1	R8	73	Blue	SKT1 (11)	SB1F (11)
34	Screened	111	Control Board	74	Orange	SKT1 (12)	SB1F (10)
35	Screened	SKT1 (7)	2 R V 1	75	Red/Blue	R.F. Module -16V	H. T. Filter Board (C1)
36	Screened	SKT1 (13)	B.F.O. Amp.(5)	76	Green	R.F. Module	H.T. Filter Board 4 RV1
37	Screened	SKT1 (31)	SB2R (7)	77	Brown/Grey	PL2 (7)	SKT11 (F)
38	Violet (red sleeve)	lstI.F Amp.(2		78	Violet (Blue sleeve)	SKT7 (1/2)	H. T. Filter Board.
39	Violet (orange sleeve)	SKTl (37)	PL2 (3)	79	Screened	1RV2	1LP3
40	Not Used.			80	Brown/Grey	SA4F4	SKT1 (19)

	NOTE:	The screened and	coaxial cables c	carry numbered	marker sleeves.
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