RADIO FREQUENCY INTERFERENCE

This equipment conforms with the requirements of EEC Directive 76/889 as to limits of r.f. interference.

H 54881-030D:C1

Instruction Manual

No. EB 2015

for

IO MHz - 520 MHz AM/ FM Signal Generator

TF 2015

H 54811-030L Change No.C2

NEW AC SUPPLY INPUT CIRCUIT

WARN ING

I. As a result of revised international safety standards the a.c. supply input circuit of this equipment has been modified as follows.

(1) A fuse is fitted in both live and neutral supply lines.

(2) The fuses now precede the supply switch.



This means that in a situation where only the neutral fuse has ruptured, parts of the equipment could remain at supply potential.

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MARCONI INSTRUMENTS LIMITED ST. ALBANS HERTFORDSHIRE ENGLAND



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Chapter

General information

1.1 INTRODUCTION

TF 2015 is a general purpose a.m. 'f.m. signal generator covering the frequency range 10 MHz to 520 MHz in eleven switched bands. Fundamental frequency generation is by voltage tuned oscillators which, in conjunction with the panel mounted controls, enable easy tuning to narrow band communication receivers up to the highest carrier frequencies.

Outputs up to 200 mV e.m.f. can be obtained with up to 80% a.m. or up to 400 mV e.m.f. for c.w. or f.m. Output is maintained constant over the whole frequency range by an automatic level control loop and is adjustable by coarse and fine attenuators calibrated in dBm in 50 Ω . An auxiliary output is available for driving a counter or digital synchronizer.

Amplitude modulation and frequency modulation are both derived from an internal 1 kHz oscillator. AM depth is variable up to 80% by a directly calibrated control. FM is directly calibrated in two full-scale ranges of 10 kHz and 100 kHz peak deviation. External modulation may also be applied and a mixed a.m. and f.m. facility is available.

The instrument, which is compact and portable, can be operated from a.c. supplies or from an external battery.



Fig. 1.1 10 MHz-520 MHz AM FM Signal Generator TF 2015

1.2 DATA SUMMARY

FREQUENCY

Range ;

Discrimination :

Stability :

Calibration accuracy :

10 MH2 to 520 MHz in 11 ranges :

 10.0 MHz to 14.3 MHz
 86 MHz to 124 MHz

 14.3 MHz to 20.5 MHz
 124 MHz

 20.5 MHz to 29.4 MHz
 124 MHz to 177 MHz

 29.4 MHz to 42.0 MHz
 177 MHz to 254 MHz

 42.0 MHz to 60.0 MHz
 363 MHz to 520 MHz

 60.0 MHz to 86.0 MHz
 363 MHz to 520 MHz

Suitable for tuning into a receiver with a 6 kHz bandwidth (\pm 3 kHz at -3 dB points).

 $\pm 1.5\%$ with extra fine tune control centred.

At a constant ambient temperature in the range ± 10 °C to ± 35 °C and after 2 hours from switch on the drift does not exceed 2 parts in 10^5 in 5 minutes.

RF OUTPUT

Level:

Attenuators :

ALC:

Total level accuracy : (above $2 \mu V e.m.f.$)

Source impedance :

Counter output :

Leakage radiation :

-127 dBm to -7 dBm (0.2 μV to 200 mV e.m.f.) with up to 80% a.m.; up to -1 dBm (400 mV e.m.f.) with no a.m.

Coarse: 11 steps of 10 dB.

Fine: 0 to 10 dB continuously variable, directly calibrated in dBm.

With a terminated output, automatic level control maintains the level meter indication substantially constant.

With carrier switch in the on position :

±1 dB up to 100 MHz ±2 dB above 100 MHz

50 Ω . VSWR better than 1.3:1 with 10 dB or more in the coarse attenuator.

Greater than 80 mV into 50 \$2

Up to 520 MHz with a receiver sensitivity set at $1.0 \,\mu\text{V}$, a 1 inch 2-turn loop feeding the receiver cannot detect a signal of this level at a distance greater than 1 inch from generator. Allows measurements to be made on receivers with sensitivities down to $0.1 \,\mu\text{V}$.

General Information

AMPLITUDE MODULATION

Depth :

Monitor :

Calibration accuracy :

Envelope distortion :

Internal frequency :

AF output :

External frequency range :

External input requirement :

FREQUENCY MODULATION

Deviation :

Frequency modulation performance :

Monitor :

Internal frequency :

External frequency response characteristics :

External input requirement :

SPURIOUS SIGNALS

Carrier harmonics :

Non-harmonic components : (coherent)

FM on c.w. :

Continuously variable up to an indicated 80% by directly calibrated control.

Reference level indicator.

 $\pm 5\%$ modulation depth at an indicated 30% rising to $\pm 10\%$ modulation depth at an indicated 80%.

Less than 5% at modulation depths up to 30%.

1 kHz ±10%.

Fixed level greater than 1.5 V r.m.s. into 10 kΩ.

50 Hz to 20 kHz.

Less than 1.5 V r.m.s. into 1 kG to set reference level.

Continuously variable in two ranges with fullscale settings of 10 kHz and 100 kHz.

Accuracy within $\pm 15\%$ of range full-scale at 1 kHz modulating frequency.

Distortion less than 3% using internal modulating oscillator. However, between 11 MHz and 60 MHz the maximum deviation is 15 kHz for 3% distortion and $\pm 15\%$ accuracy. Within this frequency range deviation above 15 kHz may be subject to an additional 2% distortion and an additional -5% deviation error.

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Reference level indicator.

 $1 \text{ kHz } \pm 10\%$.

50 Hz to 20 kHz within ± 1 dB of the response at 1 kHz. Usable over modulation frequency range 30 Hz to 100 kHz.

Less than 1.5 V r.m.s. into 1 kG to set reference level.

At least 26 dB below the carrier at output levels up to -7 dBm.

None.

With telephone psophometric weighting, equivalent deviation is less than 60 Hz up to 254 MHz and less than 100 Hz up to 520 MHz.

General Information

AM on c.w. : With telephone psophometric weighting, less than 0.2% modulation depth. FM on a.m. : With telephone psophometric weighting, leas than 500 Hz deviation at 30% modulation depth with 1 kHz modulation frequency. AM on f.m. : With telephone psophometric weighting, at 30 kHz deviation, less than 3% a.m. above 21 MHz and 5% a.m. below.

POWER REQUIREMENTS

AC mains

External d.c. :

DIMENSIONS & WEIGHT

Any voltage within the extreme limits 190 to 264 V or 95 to 132 V, at any frequency between 45 Hz and 500 Hz. Consumption ; approx, 10 VA.

24 V to 32 V (negative earth).

Height	Width	Depth	Weight
140 mm	286 mm	311 mm	5,4 kg
5½ in	$11\frac{1}{4}$ in	121 in	12 lb

1.3 VERSIONS AND ACCESSORIES

Code No	Versions
52015-015	Standard version.
52015-301	(Joint Services designation : Generator Signal CT584/2). Alternative version with BNC counter output connector and additional accessories.
	Accessories supplied
43123-076	Mains cable.
23443-814	Adapter BNC to TNC.
43126-012	RF cable TM 4969/3, BNC plug to BNC plug.
41690-102	Protective front panel cover.
46883-209	(Supplied with 52015-301 only)
	Spare fuses kit comprising:-
	2 off 23411-054, 160 mA time lag. 2 off 23411-055, 250 mA time lag. 2 off 23411-056, 0.5A time lag.

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	Accessories available	
41690-044	Carrying case.	
44411-001	Matching unit TM 5569; a that converts the effective the generator from 50 Ω to to Belling-Lee type L734/1	source impedance of 0.75Ω . BNC socket
44411-019	Matching T pad TM 5579/9 loads. Input/output voltag to BNC socket.), for matching to 75 Ω ge ratio 2:1. BNC plug
44425-501	20 dB attenuator pad TM 5: socket.	573; BNC plug to BNC
43281-007	RF fuse unit TM 9881; for resistors when connecting plug to BNC socket.	protecting attenuator to active loads. BNC
54451-121	455 kHz i.f. probe	
54451-061	470 kHz i.f. probe (ea	ch provides a crystal
54451-071	10.7 MHz i. f. probe	ntrolled signal at a andard i, f, for use in ceiver alignment.

Chapter

Operation

2.1 PRELIMINARY REQUIREMENTS

Prior to connecting the instrument to the a.e. supply check that :

(1) The voltage selector switch is set to accept the local supply (voltage regulation eliminates transformer tap changing except between 95/132 v and 190/264 V ranges).

To change the range, remove the locking plate, set the switch correctly, reverse the plate and refit.

(2) Fuses are of the required rating - see Sect.
2.4 - and secure in the holders.

2.2 AC SUPPLY CABLE

The a.c. supply cable is fitted at one end with a female plug which connects with the instrument. When fitting the supply plug ensure that the conductors are connected as follows :

Earth	-	Green/yellow
Neutral	-	Blue
Live	-	Brown

2.3 BATTERY OPERATION

For battery operation it is only required to connect a 24 V battery to the appropriate terminals located at the rear of the instrument.

A mains/battery switch is not provided since this input functions automatically and the circuits are protected against reverse polarity. This feature could be used for emergency standby battery operation which would cut in automatically in the event of a mains supply failure.

2.4 FUSES

For a. c. operation the instrument is normally Supplied with a fuse (FS1) rated at 160 mA and with the mains selector switch set for supply voltages of 190 to 264 V. For mains voltages of 95 to 132 V change the fuse for one rated at 250 mA.

For battery operation, the input circuit incorporates a fuse (FS2) rated at 500 mA.

All fuses are slow blow (time lag) types.

Note. To avoid damage to the resistors in the coarse RF OUTPUT attenuator RF Fuse Unit TM 9884 (43281-007) should be used when TF 2015 is applied to the testing of transceivers.

2.5 CONTROLS AND CONNECTORS

Front panel -- Fig. 2.1

 AC SUPPLY switch. Positioned up to switch ON.

PILOT LAMP. Lit when a.c. supply is switched on or when external battery is connected.

- (3) CARRIER switch. Set at ON for normal c. w., f. m., or a. m. operation. The +6 dB position provides a high output for c. w. or f. m. only. The OFF position allows the carrier to be switched off without switching the instrument off. (See Sect. 2.20)
- (4) CARRIER RANGE switch. Selects the required frequency range and exposes the appropriate tuning scale.



Fig. 2.1 Front panel controls

- (5) TUNE control. This is a nineteen position switch to set the generator approximately to the required frequency.
- FINE TUNE control. Provides continuous tuning between each of the nineteen positions of (5) above.
- EXTRA FINE TUNE control. An uncalibrated fine tuning control which allows very precise setting of the carrier frequency.
- (8) FUNCTION switch. Selects c. w., internal or external a.m. or f.m. or simultaneous f.m. and a.m.
- 9 DEVLATION control. Full-scale value is 10 or 100 kHz depending on setting of x1, x10 multiplier.
- (10) DEVIATION x1, x10. Multiplier for use with DEVIATION control.

- (11) SET MOD control. Adjusted to set the pointer of the MOD & CARRIER LEVEL meter to the reference mark for f.m. or a.m.
- (12) AM DEPTH. Adjusted to provide the desired a.m. depth from 0 to 80°c.
- (13) MOD & CARRIER LEVEL meter. Indicates correct reference level for a.m. and f.m. and also correct a.l.c. operation in the c.w. mode.
- RF OUTPUT (coarse). Stepped attenuator providing increments of 10 dB from 0 to 110 dB.
- (15) RF OUTPUT (fine). Continuously variable to interpolate between 10 dB steps.
- (16) RF OUTPUT connector. BNC 50 Ω. Output in - dBm given by sum of coarse and fine RF OUT-PUT control settings. Add +6 dBwhen CARRIER switch is in +6 dB position. (See Sect. 2.20)

2015 (2a)



Fig. 2.2 Rear panel connectors

Rear panel - Fig. 2.2

- MAINS INPUT connector. Bulgin P580. The a.c. supply is connected through this plug which mates with the connector fitted to the supplied mains cable.
- 2 VOLTAGE SELECTOR switch. Selects either 95 to 132 V or 190 to 264 V to suit local a.c. supply.
- (3) AC FUSE (FS1). Mains input fuse rated at 160 mA (time lag) for 190 to 264 V or 250 mA (time lag) for 90 to 132 V.
- BATTERY FUSE (FS2). Battery input luse rated at 500 mA (time lag).
- 5 BATTERY TERMINALS. Battery input terminals (negative is connected to chassis).
- (6) ENT MOD IN connector. BNC input for external modulation.
- (7) INT MOD OUT connector. BNC. 1 kHz internal modulation oscillator output when FUNCTION switch is set at INT AM or FM.

 SYNC SWEEP IN connector. BNC, accepts a frequency sweep voltage or the control voltage from a digital synchronizer.

 IF PROBE SUPPLY connector. DIN loudspeaker socket provides - 22 V for external i. f. probe.

(10) COUNTER OUT. BNC (or TNC) 50 Ω. For connection to an external counter or digital synchronizer.

2.6 SETTING FREQUENCY

- (1) Set SUPPLY switch at ON and check that the pilot lamp is lit.
- (2) Using the output cable supplied connect TF 2015 to the equipment under test.
- (3) Set FUNCTION switch at CW and the CARRIER RANGE switch to select the desired frequency range.
- (4) Set CARRIER switch at ON and check that the meter pointer is within the white box.

- (5) Rotate the stepped TUNE control to position the pointer of the tuning scale as close as possible to the desired frequency. Then adjust the FINE TUNE control to position the pointer correctly. For final precise adjustment use the EXTRA FINE TUNE control.
 - <u>Note</u>: After thermal equilibrium has been reached the typical band switching restabilization time is 10 minutes.

2.7 CW OUTPUT

For an unmodulated c. w. output.

- (1) Set FUNCTION switch at CW.
- (2) For normal operation set CARRIER switch at ON and check that the meter pointer is within the white box. For extra output set CARRIER switch at +6 dB. (See Sect. 2.20)
- (3) Set RF OUTPUT controls as required.

2.8 INTERNAL AMPLITUDE MODULATION

- Set FUNCTION switch at INT AM and CARRIER switch at ON. Note: amplitude modulation is not possible when CARRIER switch is at +6 dB.
- (2) Adjust SET MOD control to position the meter pointer at the centre of the white box.
- (3) Adjust AM DEPTH control to provide desired modulation. This control is direct reading with an accuracy which varies such that the true modulation depth may lie between 25% and 35% when 30% is indicated or between 70% and 90% when 80% is indicated.

2.9 INTERNAL FREQUENCY MODULATION

- Set FUNCTION switch at INT FM and CARRIER switch at ON. For extra output set CARRIER switch at +6 dB.
- (2) Adjust SET MOD control to position the meter pointer at the centre of the white box.
- (3) Adjust DEVIATION control to provide a deviation from 0 to 10 kHz (with multiplier at x1) or 0 to 100 kHz (with multiplier at x10).

2.10 EXTERNAL AM OR FM

- (1) Set FUNCTION switch to appropriate position i.e.: EXT AM or EXT FM and CARRIER switch at ON.
 - $\frac{\text{Note: external a, m, is not possible when }}{\text{CARRIER switch is at +6 dB}}$
- (2) Connect the external modulation source to EXT MOD IN (on rear panel); then adjust its level to position the meter pointer at the centre of the white box.
- (3) The desired modulation depth or deviation is then obtained by adjusting the relevant controls i.e. : AM DEPTH or DEVIATION.

2.11 INTERNAL AM WITH EXTERNAL FM

- (1) Set FUNCTION switch at INT AM and CARRIER switch at ON.
 - Note: amplitude modulation is not possible when CARRIER switch is at +6 dB,
- (2) Adjust the SET MOD control to position the meter pointer at the centre of the white box.
- (3) Connect the external modulation source to EXT MOD IN (on rear panel).
- (4) Set FUNCTION switch at EXT FM. Then adjust the level of the external modulation to position the meter pointer at the centre of the white box.
- (5) Set FUNCTION switch at INT AM/EXT FM. Then adjust AM DEPTH and DEVIATION controls to provide the required levels of modulation.

2.12 INTERNAL FM WITH EXTERNAL AM

- (1) Set FUNCTION switch at INT FM and CARRIER switch at ON.
 - Note: amplitude modulation is not possible when CARRIER switch is at +6 dB.
- (2) Adjust the SET MOD control to position the meter pointer at the centre of the white box.

Operation



Fig. 2.3 Conversion scale: dBm to e.m.f. and p.d.

- (3) Connect the external modulation source to EXT MOD IN (rear panel).
- (4) Set FUNCTION switch at EXT AM. Then adjust the level of the external modulation to position the meter pointer at the centre of the white box.
- (5) Set FUNCTION switch at EXT AM/INT FM. Then adjust AM DEPTH and DEVIATION controls to provide the required levels of modulation,

2.13 INTERNAL MODULATION OUTPUT

When FUNCTION switch is at INT AM, INT FM, INT AM/EXT FM, or EXT AM/INT FM, the 1 kHz modulating signal at a nominal 1 V (into a high impedance load), is available at the INT MOD OUT connector on rear panel. This signal may be used, for example, to trigger an oscilloscope at he modulating frequency or for a. f. tests.

2.14 OPERATION WITH DIGITAL SYNCHRONIZER

TF 2015 can be used with a digital synchronizer for a high degree of frequency stability. The synchronizer is driven from the COUNTER OUTPUT socket (rear of instrument) and its control voltage is applied to the SYNC/SWEEP connector.

2.15 SWEEP FACILITY

A voltage swing of 0 to +18 V applied to the SYNC/SWEEP IN connector will vary the carrier frequency over the frequency coverage of any one range. If a linear sweep is required, a suitable non-linear waveform must be used and to maintain the r.f. levelling accuracy, slow sweep speeds are necessary.

2.16 ATTENUATOR CALIBRATION

The RF OUTPUT controls (coarse and fine attenuators) are calibrated in dBm. Where e.m. f. or p. d. levels are to be used reference should be made to the conversion scale on the top of the instrument. The scale is as shown in Fig. 2.3.

2.17 MATCHING TO HIGH IMPEDANCE LOADS

To match a load that is greater than 50 Ω to the output of TF 2015 a resistor R_S is required to be added in series with the generator output as in . Fig. 2.4. The value of R_S is given by the difference between the load and the generator impedances, i.e.

$$R_s = R_L - R_o$$

in which case the voltage across the load, $\boldsymbol{V}_{L},$ is given by

$$V_{L} = \frac{E}{2}$$



Fig. 2.4 High impedance matching

When a series resistor is employed to match a receiver input impedance of 75 Ω the output impedance of the signal generator will be mismatched. Therefore it is preferable to use Matching Pad type TM 5573/3 giving a convenient 2:1 attenuation. Using this pad both the output impedance of the generator and the input impedance of the receiver are correctly matched.

2.18 USE OF IF PROBES

The i.f. probes generate crystal controlled signals at customary receiver i.f's of 470 kHz and 10.7 MHz. They are powered from the 22 V IF PROBE plug on the generator. IF probes with other crystal frequencies can be supplied to special order.

The probes provide an auxiliary test signal for use in conjunction with the normal r.f. signal from the generator in receiver testing. In operation, the probe is positioned close to the receiver so that its signal is inductively coupled into the receiver i.f. circuit while the r.f. output from the generator is connected in the normal way to the receiver input.

This facilitates a number of receiver tests such as the following :

- (1) <u>Checking receiver i.f.</u> When the signal generator is tuned to the nominal frequency of a receiver channel, any difference between the receiver i.f. and the probe frequency will produce a beat note in the receiver output. Readjusting the signal generator for zero beat, using the EXTRA FINE TUNE control, ensures that the generator is correctly tuned to the r.f. circuits in the receiver.
- (2) Overriding receiver de-sensitization. It is often difficult to tune a signal generator to a receiver incorporating some forms of desensitization, such as squelch or a battery economizer, which respond to the presence of an i. f. signal. This is because the varying r. f. signal may traverse the receiver pass band too quickly, so that an i. f. signal is not

present for long enough to sensitize the squelch or economizer circuit. By using the i.f. probe to inject an i.f. signal the receiver can be held in the sensitive condition while the r.f. generator is tuned into the pass band.

2.19 QUIETING SENSITIVITY MEASUREMENTS

- (1) When making quieting sensitivity measurements on broadband equipments the carrier signal into the equipment under test can be removed by setting the CARRIER switch at OFF.
- (2) For narrowband equipments e.g. mobile transceivers, the following procedure is preferable :
 - (a) Tune TF 2015 to the centre of the receiver pass band.
 - (b) Using the TUNE switch detune TF 2015 and note the noise output level of the receiver.
 - (c) Return the TUNE switch to its original position and then adjust the attenuator controls for the required quieting ratio – note the quieting sensitivity.

2.20 CORRECT USE OF THE +6 dB OUTPUT FACILITY

It should be noted that use of the +6 dB output facility is necessary only when c.w. outputs greater than 100 mV into 50 Ω are required e.g. :

With the CARRIER switch at ON and the coarse and fine RF OUTPUT controls at 0 and 7 dB respectively the output developed across a 50 Ω load will be 100 mV or 50 mV with the fine control at 13 dB (-6 dB). With the CARRIER switch at +6 dB and with the fine control at 13 dB the signal output level across 50 Ω then becomes 100 mV. Therefore, by adjusting the fine control from 13 dB to 7 dB the output level can be progressively increased from 100 mV to 200 mV.

For symbols and abbreviations see introduction to this chubter

R	latio Down		Ra	tio Up
VOLTAGE	POWER	DECIBELS	VOLTAGE	POWER
1.0	1.0	0	1-0	1 0
9886	.9772	-1	1-012	1 023
9772	.9550	-2	1-023	1 047
9661	.9333	-3	1-035	1 072
-9550	.9120	-4	1-047	1 096
9441	.8913	-5	1-059	1 122
9333	-8710	6	1·072	1·148
9226	-8511	7	1·084	1·175
9120	-8318	8	1·096	1·202
9016	-8128	9	1·109	1·230
8913	-7943	1.0	1·122	1·259
·8710	·7586	1·2	1 148	1-318
·8511	·7244	1·4	1 175	1-380
·8318	·6918	1·6	1 202	1-445
·8128	·6607	1·8	1 230	1-514
·7943	·6310	2·0	1 259	1-585
-7762	-6026	2 · 2	1·288	1.660
-7586	-5754	2 · 4	1·318	1.738
-7413	-5495	2 · 6	1·349	1.820
-7244	-5248	2 · 8	1·380	1.905
-7079	-5012	3 · 0	1·413	1.995
-6683	-4467	3·5	1·496	2·239
-6310	-3981	4·0	1·585	2·512
-5957	-3548	4·5	1·679	2·818
-5623	-3162	5·0	1·778	3·162
-5309	-2818	5·5	1·884	3·548
-5012	·2512	6	1.995	3-981
-4467	·1995	7	2.239	5-012
-3981	·1585	8	2.512	6-310
-3548	·1259	9	2.818	7-943
-3162	·1000	10	3.162	10-000
-2818	-07943	11	3·548	12-59
-2512	-06310	12	3·981	15-85
-2239	-05012	13	4·467	19-95
-1995	-03981	14	5·012	25-12
-1778	-03162	15	5·623	31-62

Decibel conversion table

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Contraction of

Decibel conversion table (continued)

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Ratio	Down			Ratio Up
VOLTAGE	POWER	DECIBELS	VOLTAGE	POWER
-1585	-02512	16	6·310	39-81
-1413	-01995	17	7·079	50-12
-1259	-01585	18	7·943	63-10
-1122	-01259	19	8·913	79-43
-1000	-01000	20	10·000	100-00
-07943	6·310 x 10 ⁻³	22	12-59	158-5
-06310	3·981 x 10 ⁻³	24	15-85	251-2
-05012	2·512 x 10 ⁻³	26	19-95	398-1
-03981	1·585 x 10 ⁻³	28	25-12	631-0
-03162	1·000 x 10 ⁻³	30	31-62	1,000
-02512	6.310 x 10-1	32	39-81	$\begin{array}{r} 1.585 \times 10^{3} \\ 2.512 \times 10^{3} \\ 3.981 \times 10^{3} \\ 6.310 \times 10^{3} \\ 1.000 \times 10^{4} \end{array}$
-01995	3.981 x 10-1	34	50-12	
-01585	2.512 x 10-1	36	63-10	
-01259	1.585 x 10-1	38	79-43	
-01000	1.000 x 10-1	40	100-00	
7·943 x 10 ⁻³	6·310 x 10 ⁻⁵	42	125-9	1 585 × 10'
6·310 x 10 ⁻³	3·981 x 10 ⁻⁵	44	158-5	2 512 × 10'
5·012 x 10 ⁻³	2·512 x 10 ⁻⁵	46	199-5	3 981 × 10'
3·981 x 10 ⁻³	1·585 x 10 ⁻⁵	48	251-2	6 310 × 10'
3·162 x 10 ⁻³	1·000 x 10 ⁻⁵	50	316-2	1 000 × 10'
2·512 x 10 ⁻³	6·310 x 10 ⁻⁶	52	398-1	1.585 × 10 ⁵
1·995 x 10 ⁻³	3·981 x 10 ⁻⁶	54	501-2	2.512 × 10 ⁵
1·585 x 10 ⁻³	2·512 x 10 ⁻⁶	56	631-0	3.981 × 10 ⁵
1·259 x 10 ⁻³	1·585 x 10 ⁻⁶	58 • .	794-3	6.310 × 10 ⁵
1·000 x 10 ⁻³	1·000 x 10 ⁻⁶	60	1,000	1.000 × 10 ⁶
5.623 x 10 ⁻⁴	3.162 x 10 ⁻⁷	65	1.778 × 10 ³	3-162 × 10 ⁵
3.162 x 10 ⁻⁴	1.000 x 10 ⁻⁷	70	3.162 × 10 ³	1·000 × 10 ⁷
1.778 x 10 ⁻⁴	3.162 x 10 ⁻⁸	75	5.623 × 10 ³	3-162 × 10 ⁷
1.000 x 10 ⁻⁴	1.000 x 10 ⁻⁸	80	1.000 × 10 ⁴	1·000 × 10 ⁸
5.623 x 10 ⁻⁵	3.162 x 10 ⁻⁹	85	1.778 × 10 ⁴	3·162 × 10 ⁸
3.162 x 10 ⁻⁵	$\begin{array}{l} 1.000 \times 10^{-9} \\ 1.000 \times 10^{-10} \\ 1.000 \times 10^{-11} \\ 1.000 \times 10^{-11} \\ 1.000 \times 10^{-12} \\ 1.000 \times 10^{-13} \\ 1.000 \times 10^{-14} \end{array}$	90	3.162×10^{4}	1.000 × 10°
1.000 x 10 ⁻⁵		100	1.000×10^{5}	1.000 × 10 ¹⁰
3.162 x 10 ⁻⁶		110	3.162×10^{5}	1.000 × 10 ¹¹
1.000 x 10 ⁻⁶		120	1.000×10^{6}	1.000 × 10 ¹²
3.162 x 10 ⁻⁷		130	3.162×10^{6}	1.000 × 10 ¹³
1.000 x 10 ⁻⁷		140	1.000×10^{7}	1.000 × 10 ¹⁴

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Chapter 3

Technical description

This chapter, which outlines the overall and circuit functions of TF 2015, is intended to be read with reference to the circuit diagrams and illustrations contained within this manual.

3.1 MECHANICAL CHARACTERISTICS

All printed circuit boards and sub-assemblies in the instrument are allocated a unit identification number in the sequence A0 to A12 and A16 to A23 and where practical the assembly is marked with this number. The main sub-assembly layout is shown in Figs. 4.3 and 4.4.

Construction consists of front and rear panels supported by side frames which also carry an r.f. The r.f. box contains eleven separate oscilbox. lator circuits A1 to A11 mounted on a mother board A12 which is hinged to allow access to the associated components. The box also contains a screened compartment A16 containing the r.f. amplifier, modulator and counter amplifier circuits. The automatic level control (a.l.c.) circuits A17 are mounted on top of this compartment. The low r.f. leakage lid is retained by four fixing screws. The underside of the r.f. box carries the filter box and units A18, A19 and A20.

Units A21 and A23 are mounted on one of the side frames and the opposite frame carries the mains transformer. Inputs and outputs are conveyed through BNC connectors and terminals located on the front panel and at the rear of the instrument.

Since TF 2015 is a portable instrument, a removable cover is clasp fitted to the case to provide front panel protection and also stowage space for some accessories. A prop fixed to the lower section of the case enables the instrument to be tilted for easy operation of the controls.

3.2 OVERALL FUNCTION

A separate voltage tuned oscillator is used for each of the eleven ranges. Output from the selected oscillator, which is maintained at a constant level by the first automatic level control (a.l.c.) loop, is applied to a pre-amplifier. The frequency of each oscillator is controlled by means of a pair of varactor diodes across the tuned circuit. The capacitance of the diodes, and hence the resonant frequency of the tuned circuit, is determined by the control voltage applied to the diodes. This arrangement provides for both tuning and frequency modulation, a variable d.c. control voltage being applied for tuning purposes and an a.f. modulating signal for f.m.

Output from the pre-amplifier is applied to a modulator and at a suitable level to a counter amplifier. The r. f. output from the modulator which is approximately 30 mV is passed to a wide band amplifier incorporating an interstage current controlled attenuator.

Output from the amplifier is applied to the input of the coarse attenuator (coarse r. f. output control) and through a detector circuit to a second a. l. c. loop which, operating in conjunction with the fine attenuator (fine r. f. output control) sets and then maintains the level of the r. f. input to the coarse attenuator.

The r.f. carrier can be amplitude or frequency modulated using either the internal modulation oscillator or an external signal. For a.m. the a.f. signal is applied to the modulator through a push-pull driver circuit and for f.m. to the tuning circuits. To ensure that deviation is maintained acceptably constant over each frequency range and to minimize distortion, tracking and pre-distortion circuits are employed.

A Wien bridge oscillator operating at a nominal frequency of 1 kHz is used for internal modulation and to provide an a.f. output. For both a.m. and f.m. the internal a.f. signal is routed through the SET MOD control which is adjusted to obtain a modulation reference level. The desired a.m. depth or f.m. deviation is then obtained by adjustment of the appropriately indicated controls.

Inputs are provided for connection of an external modulating signal and for application of a suitable sweep waveform. A 22 V d.c. output is available as a supply for certain associated equipments, e.g. i.f. probes, and an output is provided for connection to a frequency counter. In c. w. operation the meter indicates correct operation of the a.l.c. In the modulated modes it monitors the a.f. reference level.

3.3 CIRCUIT FUNCTIONS

Block schematic diagram Fig. 3.1 interrelates the various circuits employed in TF 2015 whilst the sections which follow explain the function of each.

3.4 RF OSCILLATORS

Circuit diagram Fig. 7.1 Unit A12 (sub-assemblies A1 to A11)

All the oscillators use a modified Hartley circuit and differ only in the value of components. Oscillators A1 to A9 employ a field effect transistor, A10 and A11 use bipolar transistors. Feedback is established by inductive coupling between the two sections of the primary winding of transformer T1.

Frequency is determined by varactor diodes D1 and D2 operating as variable capacitors to tune the primary winding of T1. The capacitance values of D1 and D2 are varied by application of d. c. voltage levels set by the tuning circuit A21.

For ranges 1 to 9, trimmer capacitor C4 is adjusted to correctly set the high frequency end of the range with the core of T1 adjusted at the low frequency end. On ranges 10 and 11 only C4 is adjustable.

The oscillators are individually activated by switched application of the d. c. supply whilst output from the selected oscillator is obtained from the secondary winding of T1 through the appropriate steering diode, D1 to D11. Diode D12 is incorporated to reduce the effective capacitance of D1 to D7 which could influence the frequency coverage of ranges 8 to 11.

3.5 TUNING AND FREQUENCY MODULATION

Circuit diagram Fig. 7.3 Units A19 and A21

Because the varactors used for tuning have a non-linear capacitance/voltage characteristic, a 19-position switch with a chain of selected resistors is used to provide a stepped linear change of frequency. Potentiometer AOR 38 interpolates between switch positions. The additional potentio-

meter A0R37 enables TF 2015 to be precisely tuned to a receiver under test.

For a frequency modulated output the a.f. signal is routed to the varactors through the f.m. controls, tracking resistors A0, R5 to R15, the predistortion circuit A23, amplifier TR3, TR4, TR5 and the tuning control circuits. The tracking resistors are automatically selected by the range selector switch to maintain equal deviation over each frequency range. A21 TR1 is used as a low impedance source for the f.m. drive.

3.6 FIRST ALC LOOP

Circuit diagram Fig. 7.2 Units A16 and A17

Output from the selected oscillator is maintained at a nominal level of 50 mV by the first automatic level control (a. l. c.) loop. The oscillator output is applied to a detector circuit, D1 and D2, which is referenced by the set voltage at the junction of R1 and R2 on A17.

Output from the detector is applied to one input of a differential pair, TR1 and TR2, where it is compared with a reference level preset by network R4, R5, R6 and R7. Any difference in the level of the signals either towards positive or negative appropriately varies the bias to TR3 which then varies the current in the f. e. t. in the selected oscillator to re-establish the required output level.

Diodes A17D1 and D2 compensate for the temperature changes of the detector diodes, A16D1 and D2. Capacitors A17, C2 and C11 are incorporated to minimize spurious f.m. which could be produced by a. I. c. circuit noise.

3.7 PRE-AMPLIFIER

Circuit diagram Fig. 7.2 Unit A16

The controlled output level from the selected oscillator is applied to a wide band pre-amplifier providing a gain of approximately 12 dB, i.e. a nominal output of 200 mV.

The circuit consists of four stages, TR1, TR2, TR3 and TR4, and employs both common base and common emitter configurations. The common base circuits and screens mounted on the printed circuit board are incorporated to provide the reguired inter-stage isolation.

3.8 MODULATOR AND AMPLITUDE MODULATION

Circuit diagram Fig. 7.2 Unit A16

The circuit configuration is basically that of a single balanced bridge modulator employing p-i-n diodes operating as variable (current controlled) r.f. resistors to achieve minimum r.f. carrier distortion.

Initially, with no d. c. or a. f. inputs, the two inputs from T1 are 180° out of phase and, since R17 and R18 are equal, the bias currents and consequently the resistance of the diodes are equal. With these conditions the bridge is balanced so no signal is present at T2.

To provide a constant r. f. output for c. w., d. c. offsets from circuits A17 TR7 and TR8 are applied to partially unbalance the bridge. For amplitude modulation the a. f. signal from the pushpull driver circuit A17 TR4, TR5 and TR6 is applied to the bridge modulator. The a. f. signal produces offset levels proportional to its amplitude so the bridge becomes progressively unbalanced to an extent which is greater or less during each positive and negative half cycle of the a. f. waveform.

3.9 COUNTER AMPLIFIER

Circuit diagram Fig. 7.2 Unit A16

Part of the r. f. signal voltage at the output of the pre-amplifier is applied to a common emitter amplifier, TR6, which serves to drive an external frequency counter. Output at the COUNTER OUT connector is nominally 100 mV into 50 Ω_{c}

3.10 AM DRIVER AND DC OFFSET CIRCUITS

Circuit diagram Fig. 7.2 Unit A17

The two circuits TR7 and TR8 operate to set the level of the d.c. offsets applied to the modulator and consequently the r.f. level to A16 TR5.

With the CARRIER switch at ON (normal r. f. output) TR7 and TR8 are non-conducting so the d. c. offset levels to the modulator are determined by R23 and R24. When the CARRIER switch is at +6 dB TR7 and TR8 are switched on to connect R25 and R26 in parallel with R23 and R24 to further unbalance the modulator, reducing its insertion loss by approximately 8 dB. Diode D3 is incorporated to prevent current flowing in R26 when TR7 and TR8 are in the off condition.

The a.m. driver serving the modulator consists of a push-pull a.f. amplifier, TR4 and TR6, which applies offset levels proportional to the varying amplitude of the a.f. cycle to produce an amplitude modulated output as explained in Section 3.8. TR5 maintains the amplifier current constant during the a.f. cycle.

3.11 WIDE BAND AMPLIFIER

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Circuit diagram Fig. 7.2 Unit A16

This consists of three common emitter stages TR5, TR7 and TR8 with an interstage current controlled variable attenuator. This operates in conjunction with the second a. l. c. loop and fine r. f. output control to set and maintain constant the signal level at the input to the coarse r. f. output attenuator unit A22.

The attenuator which controls the gain of the amplifier employs p-i-n diodes D7 and D8 as current controlled variable resistors, their resistance being determined by bias levels controlled by A17 TR9 and TR10 (see section 3.12).

3.12 SECOND ALC CIRCUIT

Circuit diagram Fig. 7.2 Unit A17

The signal level at the output of the wide band amplifier is detected by diodes D9 and D10 and and then passed to A17 IC1 where \mathbf{I} is compared with a level dictated by the setting of the fine r. f. output control, A0R16. Any difference in level between the two applied signals causes the output of IC1 to change. This changes the currents through TR9 and TR10 which controls the resistance of the variable attenuator diodes, D7 and D8, and establishes the required input to the coarse r. f. output control (see Section 3. 13).

3.13 FINE RF OUTPUT CONTROL

Circuit diagram Fig. 7.2 Units A16, A17, A18

The calibration accuracy of the fine r. f. output control is primarily dependent upon correct performance of the second a. l. c. loop incorporating circuits A18 TR2, the a. l. c. amplifier A17, the interstage attenuator circuit A16 D7 and D8 and the detector circuit A16 D9 and D10.

Technical description

The circuit A18 TR2 provides a fixed reference voltage to the detector D9 and D10 to produce a voltage level at A17 IC1 pin 2 which is determined by the difference between the reference voltage and the detected r. f. voltage. It also provides a preset voltage to A17 IC1 pin 3 at a level which is dictated by the setting of the fine r. f. output control A0 R16. Output from IC1 pin 6. which depends on the voltage levels at pins 2 and 3, controls the current through the differential pair TR9 and TR10 such that the resistances of D7 and D8 in the interstage attenuator are set to establish the required amplifier gain and therefore the desired r. f. output level.

When the fine attenuator control A0 R16 is adjusted for a different r. f. output, the voltage leve at IC1 pins 2 and 3 will momentarily be unequal, which causes the output at IC1 pin 6 to change and consequently the current through TR9 and TR10. The increased or decreased current sets the attenuator D7 and D8 to appropriately vary the gain of the wide band amplifier to reestablish equal voltage levels at IC1 pins 2 and 3 and a higher or lower signal level to the input of A22.

3.14 MODULATION OSCILLATOR

Circuit diagram Fig. 7.3 Unit A19

A Wien bridge circuit operating at a nominal frequency of 1 kHz is employed for internal modulation of the r. f. signal and to provide an a. f. outbut the INT MOD OUT connector.

The input of TR1 is connected across the shunt arm R2, C1 of the bridge and its output is bassed to the amplifier circuit TR2. The oscilator output signal is across R8 and it is applied o the series arm of the bridge C2. R6 to obtain bositive feedback. During the non-oscillating condition prior to 'switch on' the amplifier gain s high. At 'switch on' oscillation commences readily, causing the resistance of thermistor R7 o fall and the amplifier gain to stabilize at a value ust sufficient to maintain oscillation.

For amplitude modulation the 1 kHz signal s routed through the SET MOD control and AM DEPTH control to the a.m. driver circuit A17.

For frequency modulation the 1 kHz signal s applied through the SET MOD control and DEVIATION controls to the switched i.m. trackng resistors A0 R5 to R15 (ganged to range selector switch) and then through the predistortion circuits A23 and f.m. driver TR3, TR4, TR5 to the tuning circuits A21.

3.15 PREDISTORTION CIRCUIT

Circuit diagram Fig. 7.3 Unit A23

This circuit, which is especially applicable to ranges 1 and 2, minimizes f. m. distortion by predistorting the a.f. waveform to compensate for the capacitance/voltage characteristic of the varactors. The circuit is designed to expand the positive half of the a.f. waveform and compress the negative half.

Circuit performance is largely dependent upon the attenuator R2 and R3 which is set by the shunt diodes D1 and D2.

During the positive half of the a.f. waveform D1 is caused to conduct by the positive signal level at A21 pin 13 and since D1 is effectively in parallel with R2 the insertion loss of the attenuator is reduced to increase the amplitude of the positive half of the waveform to the f.m. driver. During the negative half of the waveform D2 is caused to conduct, effectively reducing the value of R3 and limiting the waveform excursion.

The circuit consisting of R7 to R10 is designed to reduce frequency drift by appropriately varying the d.c. supply to the tuning network R17-R36 to compensate for temperature rise within the instrument.

3.16 -6 dB OUTPUT

Circuit diagram Figs. 7.2 and 7.3 Units A16, A17, A18

When the CARRIER switch is at ± 6 dB the levels of the d. c. offsets to the modulator A16 are increased to reduce its insertion loss (see Section 3.10). Resistor R25 in the circuit A18 TR2 is connected (through D6) in parallel with R24 to increase the current through the network R22, R23 and A0 R16 and consequently the set voltage output from A0 R16 to IC1 pin 3. The a.1. c. loop then operates to increase the r.f. output by 6 dB.

With the CARRIER switch at -6 dB amplitude modulation is not possible and is prevented by connecting the a.m. drive line to earth.

3.17 MONITOR CIRCUITS

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Circuit diagram Figs. 7.2 and 7.3 Unit A18

When the FUNCTION selector is set for a.m. or f.m. modulation the output from the internal or external modulation oscillator is routed to the monitor circuit A18 TR1.

The a.f. is rectified by the circuit D1 and D2 and the positive d.c. output is applied to the meter ME1 through the emitter follower TR1 and diode D3. To obtain the desired sensitivity the meter is backed off by a set d.c. level from the network R10, R11 and R12 and by preset adjustment of R5 which sets a reference level to the detector circuit D1 and D2. Thermistor R9 provides the necessary temperature compensation to the network R10, R11 and R12 while D3 prevents the application of reverse currents to TR1.

Since the calibration accuracy of the fine r. f. control is largely dependent upon the performance of the a.l.c. loop A17, the meter (when the FUNC-TION selector is set at CW) serves only to indicate correct operation of the a.l.c. loop. The circuit R12 to R16 with R18 and R19 forms a limits bridge with R17 preset to position the pointer of the meter at the centre of the white box on the meter scale. Provided the output from A17 IC1 pin 6 remains within certain limits, diodes D4 and D5 are nonconducting.

Incorrect operation of the a.l.c., which could be caused by malfunction of the preceding r.f. circuits, produces an output from IC1 which will be above or below the expected limits. The appropriate diode then conducts causing the bridge condition to change such that the meter pointer will be positioned at zero or f.s.d.

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3.18 COARSE RF OUTPUT CONTROL

Circuit diagram Fig. 7.4 Unit A22

This provides a loss of 110 dB in steps of 10 dB. The pad sections consist of resistive networks with a characteristic impedance of 50 Ω . The unit is divided into compartments to ensure maximum shielding between pads. Pads are connected into circuit by microswitches housed in screened compartments and operated in pairs by leaf springs actuated by cams on the control spindle.

3.19 POWER UNIT

Circuit diagram Fig. 7.4 Unit A20

The power unit, which operates by switch selection from a. c. supplies of 95 to 132 V or 190 to 264 V, 45 to 500 Hz, consists basically of mains transformer A0T1, a full wave rectifier circuit A20 D1, D2, D3 and D4 and a conventional voltage regulator circuit. The +22 V output from the regulator supplies all the circuits except the frequency control sections which are supplied from a second sub-stabilizer which provides +19 V. The 19 V regulator sampling lines maintain the voltage across the tuning network. TR6 and R9 provide a low noise voltage reference for the sub-stabilized supply.

3.20 SUPPLY FILTERS

Filters are incorporated in supply lines to various units to ensure that spurious r.f. signals are not introduced into the circuits.

All the filters are contained in a separately screened box which consists of two screen divided sections. Each filter employs two series connected chokes and lead-through type shunt capacitors. Chapter

Maintenance

4.1 INTRODUCTION

This chapter contains information for keeping the equipment in good working order and for checking its overall performance. Before attempting any maintenance on the equipment you are advised to read the preceding Technical Description chapter.

4.2 SCREW FASTENERS

The majority of screw threads used in the instrument are metric of various sizes but in some positions BA threads are used. All chromium plated screws and all screws tinted blue are metric. Ensure that screws removed are refitted in original positions.

4.3 ACCESS TO SUB-ASSEMBLIES AND COMPONENTS

Removal of case

The case is in two sections. Remove the six screws at the rear of the instrument and then slide off top and bottom sections.

Sub-assemblies A1 to A12

A1 to A11 are r.f. oscillator assemblies mounted on a mother board. A12, which is contained within the r.f. box. To obtain access remove the cover of the r.f. box by removing the four fixing screws. UnitA12 is hinged to provide access to the oscillator assemblies as shown in Fig. 4.1. The component layout for each assembly is as shown in Fig. 4.2 and Fig. 4.4.

Units A16 and A17

The position of these units is shown in Figs. 4.1 and 4.4. Access to A16 is obtained by re-

moving the four screws holding the screening plate and then by carefully folding this back with A17 attached. A17 is directly accessible - see Fig. 4.4.

Units A18, A19, A20, A21, A23

The position of these units is shown in Fig. 4.3 and all are directly accessible from the underside of the instrument.

Supply filters

These are contained in a separately screened compartment. Inputs and outputs can be accessed through the associated capacitors which are accessible by carefully lifting A18 upwards and folding A12 back as shown in Fig. 4.1.

Unit A22

To remove the attenuator, carefully disconnect the two semi-rigid coaxial cables, remove the control knob and nut holding the unit to the panel. remove the bracket screw then lift the unit from the chassis.

4.4 PRELIMINARY CHECKS

- Check to ensure that all switches are undamaged and operating correctly and verify that connectors are securely mated.
- (2) Check that fuses are of the correct rating and type, and fit correctly in the holders.

4.5 TEST EQUIPMENT

The test equipment required for maintenance and repair of the instrument is listed in Table 4.1.



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Fig. 4.1 Layout, top view with boards A12 and A17 hinged back

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Table 4.1

Test equipment

ltem	Description	Recommended model
a	50 MHz counter	MI TF 2410
b	500 MHz converter	MI TM 8334
с	'N' type 50 Ω load	MI TM 7967
d	Digital voltmeter, up to 50 V with resolution of 1%	MI TF 2670
e	30 MHz oscilloscope	MI TF 2201
f	Wave analyser	MI TF 2330A
g	RF millivoltmeter	MI TF 2603
h	'T' connector	MI TM 7948
i	RF milliwattmeter	M1-Sanders type 6460
j	22 cm air spaced line	General Radio type 874-LO
k	DC microvoltmeter	MI TF 2655
1	Distortion factor meter	MI TF 2331
m	AM/FM modulation meter	MI TF 2300B
n	AF oscillator	MI TF 1101
о	Psophometer	Hatfield Instruments type MBC 1000
р	RF detector	MI TM 9650
ą	Short circuit current monitor	- See Fig. 4.5
r	Spectrum analyser	Hewlett Packard type 8554L/8552A/140T
s	Multimeter	•. GEC Selectest
t	Variable voltage transformer	Variac
	· · · · · · · · · · · · · · · · · · ·	



Fig. 4.5 Short circuit load

4.6 PERFORMANCE CHECKS

Many of the methods in this section are simplified and of restricted range compared with those which would be needed to demonstrate complete compliance with the specification. They should be regarded only as providing a check procedure, for Ę

use during routine maintenance. to determine whether adjustment or repair is necessary. Any figures given are for guidance only and should not be taken as guaranteed performance figures unless they are quoted in the Data Summary Sect. 1.2.

> NOTE. If the results quoted in the following sections are not obtainable. refer to the related section in Chapter 5.

4.6.1 Power supply unit A20

Test equipment : items d. f, t

1. With TF 2015 connected to the a.c. supply and switched on, connect the digital voltmeter between tag 17 (positive) and tag 15 (negative) and check that the meter indicates $22 \text{ V} \pm 100 \text{ mV}$.¹ If necessary adjust R4 to obtain this requirement.

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2. Connect the digital voltmeter between tag 12 (positive) and tag 15 (negative) and check that the meter indicates 19 V ± 100 mV. If necessary, select values for R14 and R17 to obtain this requirement.

3. Apply the a.c. input through a variable voltage transformer and check, using the digital voltmeter, that with inputs to the power supply of between 190 and 260 V the voltage between tag 12 and tag 15 is maintained to within ± 40 mV of that obtained in (2).

4. Disconnect the digital voltmeter and with the a.c. input at 230 V connect the wave analyser set for battery operation, in turn to the 22 V and 19 V outputs. The 100 Hz ripple should not be greater than 100 μ V r.m.s. and 1 μ V r.m.s. at the respective outputs.

5. Disconnect the a.c. supply and connect a 24 V d.c. supply to the battery terminals and check that the 19 V output at tag 12 is within ± 100 mV.

Disconnect the d.c. supply and re-connect TF 2015 directly to the a.c. supply.

4.6.2 Frequency calibration unit A12 (subassemblies A1 to A11)

Test equipment : items a, b.

1. Set the FUNCTION selector at CW and CAR-RIER switch at ON then ensure that the pointer of the MOD/CARRIER meter is within the white box on e meter scale.

2. Connect the counter to the COUNTER OUT socket (rear of instrument). Set the CARRIER RANGE selector in turn to ranges 1 to 11 and check, using the TUNE, FINE TUNE and EXTRA FINE TUNE controls, that the scale calibrations at three points on each range i.e. low and high frequency ends and at a centre frequency are within $\pm 1.5\%$ of counter indication.

If necessary, for ranges 1 to 9 inclusive adjust the appropriate core of T1 for low frequency correction and the appropriate capacitor C4 for high frequency correction. Since the core and C4 are interdependent ensure that after the second adjustment the first adjustment remains correct. For ranges 10 or 11 only C4 can be adjusted for high frequency correction. The oscillators are positioned as shown in Fig 4.1.

4.6.3 RF output (normal)

Test equipment : item i.

1. With the FUNCTION selector and CARRIER switch set as in 4.6.2 (1) connect the power meter to the output of TF 2015.

2. Set the RF OUTPUT controls at 0 and -7 dBm respectively.

3. Verify that the pointer of the MOD/CARRIER LEVEL meter is within the white box. Using the TUNE control to step through each frequency range, verify that the output level is held at -7 dBm ± 1 dB up to 100 MHz and within ± 2 dB from 100 MHz to 520 MHz.

4. Ensure that when making each check the pointer of the MOD/CARRIER meter remains within the white box.

4.6.4 RF output (+6 dB)

Test equipment : item i.

1. With the FUNCTION selector, CARRIER switch, RF OUTPUT controls, set as in 4.6.2 and with the power meter connected to the RF OUTPUT socket set the frequency of TF 2015 at 100 MHz.

Note the indicated output then set the CARRIER switch at ± 6 dB and check that the output level has been increased by 6 dB ± 1 dB.

2. Repeat the check with TF 2015 tuned to 500 MHz and check that with the CARRIER switch at +6 dB the output level has been increased by 6 dB $\pm 1 \text{ dB}$.

4.6.5 Checking v.s.w.r. (frequencies 350 to 500 MHz)

Test equipment : items j, k, q.

1. Connect the test equipment as in Fig. 4.6.

2. Set the FUNCTION selector at CW, the CAR-RIER switch at +6 dB, the RF OUTPUT controls at 10 dBm and 7 dBm respectively and the output frequency at 500 MHz.

3. Ensure that the pointer of the MOD/CARRIER meter is within the white box.

4. Set the adjustable line to $\frac{1}{2}$. λ (half wavelength) of the signal by applying the formula :

$$\lambda = \frac{300}{f}$$
where λ = wavelength in metres
and f = frequency in MHz

Adjust the line length for a maximum indication on the microvoltmeter and note its reading. Adjust the line length for minimum indication on the microvoltmeter, at $\frac{1}{4}$ wavelength away from the maximum response. Again note the voltmeter reading.

By definition, v.s.w.r. is the ratio of the maximum to minimum line voltage. The r.f. voltage developed across the virtual short circuit is detected by the diode and the corresponding d.c. maximum and minimum (e max and e min) are measured on the microvoltmeter. As the diode is being used on the lowest part of its characteristic a square law voltage applies, the v.s.w.r. being equal to

v.s.w.r. =
$$\sqrt{\frac{e \max}{e \min}}$$

The v.s. w.r. should be better than 1.3:1



Fig. 4.6 Test gear arrangement to check v.s.w.r.

4.6.6 Checking v.s.w.r. (frequencies 10 to 350 MHz)

Test equipment : items c, g, h.

1. With the TF 2015 controls set as in 4.6.2, connect the voltmeter to the RFOUTPUT of TF 2015 using the T connector and note the indicated output level.

2. Connect the 50 Ω load to the T connector and again note the indicated output level.

3. Compute the impedance Z using the following formula :

$$Z = \frac{50E}{V} - 50$$

where E = the open circuit output level and V = the output across the 50 Ω load.

From the above,

$$s.w.r. = \frac{Z}{50}$$

v

The v.s. w.r. should be better than 1.3:1.

4.6.7 Modulation oscillator performance

Test equipment : items a, g, l.

1. Set the FUNCTION selector at INT MOD and the CARRIER switch at OFF.

2. Connect the counter to the INT MOD OUT socket (rear of instrument) and check that the indicated frequency is between 920 Hz and 1080 Hz.

3. Disconnect the counter and connect the distortion factor meter (d. f. m.) to the INT MOD OUT socket.

The measured distortion should not be greater than 0.25%.

Set the d.f.m. to measure voltage and check that the measured voltage closely approximates 1.5 V r.m.s.

4.6.8 Checking f.m. deviation

Test equipment : item m.

1. Set the FUNCTION switch at INT FM, set the CARRIER switch at ON, set the frequency of TF 2015 at 20 MHz then adjust SET MOD control to position the pointer of MOD/CARRIER meter at centre of white box on meter scale.

2. Connect the modulation meter tuned to 20 MHz to the RF OUTPUT on TF 2015.

3. Set the DEVIATION controls at x10 and 10 kHz respectively. The measured deviation should be within $\pm 15\%$ of 100 kHz. If necessary adjust SET

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MOD to obtain this requirement then reset A18 R5 to position the pointer of the meter at centre of white box.

4. Set the DEVIATION multiplier at x1. The measured deviation should now be within $\pm 15\%$ of 10 kHz. If necessary adjust A19 R12 to obtain this requirement.

4.6.9 Checking a.m. depth

Test equipment : item m or e.

1. Set the FUNCTION selector at INT AM, set the CARRIER switch at ON. Set the frequency of TF 2015 at 20-MHz then adjust the SET MOD control to position the pointer of the MOD/CARRIER ter at the centre of the white box.

2. Set the AM DEPTH control at 80%.

3. Connect the modulation meter to the RF OUT-PUT on TF 2015. Set the RF OUTPUT controls to provide a suitable input to the modulation meter then tune the modulation meter to TF 2015.

4. Check that the measured a.m. depth is between 70% and 90%.

5. Repeat with the AM DEPTH control at 30% and check that the measured a.m. depth is between 25% and 35%.

If necessary adjust A18 R3 to obtain the best results for both 80% and 30% depths.

NOTE. If a modulation meter is not available the a.m. depth can be assessed by using the oscilloscope to measure the peak and trough values of the modulation envelope. The a.m. depth is then determined by

AM depth
$$\frac{C}{C} = \frac{Vp - Vt}{Vp + Vt} \times 100$$

where Vp and Vt are the measured peak-to-peak and trough-to-trough amplitudes respectively.

4.6.10 Checking f.m. tracking

Test equipment : item m.

1. Connect the modulation meter to the RF OUT-PUT of TF 2015. 2. Set the FUNCTION selector at INT FM and the CARRIER switch at ON.

3. Adjust the SET MOD control to position the pointer of the MOD/CARRIER meter at the centre of the white box. Then set the DEVLATION multiplier at x10 and the deviation control at 10 kHz.

4. Set the CARRIER RANGE selector at 10-14.3 and check that at frequencies 10, 12 and 14 MHz the measured deviation is 100 kHz $\pm 15\%$.

4.6.11 Checking a.l.c. system

Correct operation of the a.l.c. system can be quickly proved as follows :

1. Set the FUNCTION selector at CW and the CARRIER switch at ON then check that the pointer of the MOD/CARRIER meter is within the white box.

2. Switch to each CARRIER RANGE in turn then use the TUNE control to check that over each frequency range the pointer of the MOD/CARRIER meter remains within the white box.

4.6.12 Counter output level

Test equipment : item i.

1. Set the FUNCTION selector at CW and the CARRIER switch at ON.

2. Connect the power meter to COUNTER OUT socket (rear of instrument) then check that with TF 2015 tuned to 100 MHz the output level is not less than -7 dBm.

4.6.13 External modulation sensitivity

Test equipment : item n.

1. Set the FUNCTION selector at EXT AM and the CARRIER switch at ON:

2. Using the external a.f. generator connected to the EXT MOD IN socket (rear of instrument) check that, with an input signal between 1.2 V and 1.8 V r.m.s. at frequencies of 30 Hz, 1.5 kHz, 10 kHz and 50 kHz, the pointer of the MOD/CARRIER meter can be positioned at the centre of the white box on scale.



Fig. 4.7 Test gear arrangement to check f.m. and a.m. distortion and f.m. on a.m.

3. Repeat the above with FUNCTION selector at EXT FM.

NOTE. The TF 2015 meter can be overloaded if the applied a.f. input exceeds 10 V.

4.6.14 AM distortion

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Test Equipment : items 1, m.

1. Connect the test equipment as shown in Fig. 4.7.

2. Set the FUNCTION selector at CW, the CARRIER switch at ON and the frequency of TF 2015 at 100 MHz. Then check that the pointer of the MOD/ CARRIER meter is within the white box on meter scale.

3. Set the FUNCTION selector at INT AM. •. Then adjust SET MOD control to position the pointer of the MOD/CARRIER meter at the centre of the white box on scale.

4. Set the AM DEPTH control at 30%. Then tune the modulation meter to TF 2015 and check that the distortion indicated on the distortion factor meter is not greater than 5%.

5. If required, repeat the check using the external a.f. generator for modulation frequencies at 50 Hz and 20 kHz.

4.6.15 FM distortion

Test equipment : items l, m.

1. With the test equipment connected as in Fig. 4.7 repeat 4.6.14 (2). Then set the FUNCTION selector at INT FM and the SET MOD control to position the pointer of the MOD/CARRIER meter at the centre of the white box.

2. Set the DEVIATION multiplier at x10 and the DEVIATION control at 10 kHz.

3. Ensure that the modulation meter is tuned to TF 2015. Then check that the distortion indicated on the distortion factor meter is not greater than 5%.

4. If desired, repeat the check using the external a.f. generator for modulation frequencies at 50 Hz and 20 kHz.

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4.6.16 Spurious f.m. on c.w.

Test equipment : items m, o.

1. Connect the test equipment as in Fig. 4.8.



Maintenance 🗅

2. Set the FUNCTION selector at CW and the CARRIER switch at ON. Then tune TF 2015 to 174.5 MHz and ensure that the pointer of the MOD/ CARRIER meter is within the white box.

Set the FUNCTION selector at INT FM. Then 3. adjust the SET MOD control to position the pointer of the MOD/CARRIER meter at the centre of the white box.

4. Set the DEVIATION multiplier at x1 and the DEVIATION control at 5 kHz.

Set the psophometer controls as follows : 5.

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INPUT SELECTOR :	Terminated
WEIGHTING :	FILTER NO.1
ATTENUATOR :	0 dB

6. Using 44 MHz crystal, tune and crystal lock the modulation meter to TF 2015 (modulation meter at 176.0 MHz). Then set the ADJUST CALIBRA-TION control on the psophometer for a meter indication of 0 dB.

7. Set the FUNCTION selector at CW. Then adjust the psophometer attenuator to restore the meter pointer at 0 dB, and check that the change is greater than 38 dB (less than 61 Hz).

8. If desired repeat the check at a carrier frequency of 481.5 MHz and check that the change is greater than 34 dB (less than 100 Hz).

'7 Spurious a.m. on c.w.

Test equipment : items p, o.

1. Connect the test equipment as shown in Fig. 4.9. Then with the FUNCTION selector at CW. the CARRIER switch at ON and TF 2015 tuned to 100 MHz check that the pointer of the MOD/CAR-RIER meter is within the white box.

Set the FUNCTION selector at INT AM. 2. Then adjust the SET MOD control to position the pointer of the MOD/CARRIER meter at the centre of the white box on meter scale.

3. Set the AM DEPTH control at 20% and the TF 2015 output at maximum.

4. Set the Psophometer controls as follows:-

INPUT SELECTOR :	THRO
WEIGHTING :	FILTER NO.1
	(Telephone)

VOLTAGE/dBm RANGE

10 mV

e.

5. Switch the psophometer to INTERNAL SUPPLY. Then adjust the VOLTAGE/dBm RANGE switch and the ADJUST CALIBRATION control for a convenient indication on the meter.

Note the setting of the VOLTAGE/dBm RANGE switch and the meter indication.

Set the TF 2015 FUNCTION selector at CW. 6. Then increase the VOLTAGE/dBm RANGE switch on psophometer by 40 dBm. The meter should indicate a level which is less than that noted in 5.



Fig. 4.9 Test gear arrangement to check spurious a.m. on c.w. and a.m. on f.m.

4.6.18 Spurious a.m. on f.m.

Test equipment : items p. o.

1. With the test equipment connected as in Fig. 4.9 repeat 4.6.17 (1) and (2)

2. Set the AM DEPTH control at 30% and the TF 2015 output at maximum.

3. Set the psophometer controls as follows:-

INPUT SELECTOR :	THRO
WEIGHTING :	FILTER NO. 1
VOLTAGE/dBm RANGE ;	10 mV

Switch the psophometer to INTERNAL SUPPLY. 4. Then adjust the VOLTAGE/dBm RANGE switch and the ADJUST CALIBRATION control for a convenient indication on the meter. Note the setting of the VOLTAGE/dBm RANGE switch and the meter indication.

5. Set the TF 2015 FUNCTION selector at INT FM. Adjust the SET MOD control to position the pointer of the MOD/CARRIER Meter at the centre of the white box. Then set DEVIATION Multiplier at x10 and DEVIATION control at 3(i.e. 30 kHz).

6. Increase the VOLTAGE/dBm RANGE switch on psophometer by 20 dB. The meter should indicate a level which is not greater than that noted in 4.

4.6.19 Spurious f.m. on a.m.

Test equipment : items 1, m.

1. Connect the test equipment as in Fig. 4.7.

2. Set the FUNCTION selector at INT FM, the CARRIER switch at ON, the TF 2015 frequency at 100 MHz and its output at maximum. Then adjust the SET MOD control to position the pointer of the MOD/CARRIER meter at the centre of the white box on meter scale.

3. Set the DEVIATION multiplier at x1 and the DEVIATION control at 1 kHz then tune the modulation meter (preferably using crystal lock) to TF 2015 and ensure that the indicated deviation is 1 kHz.

4. Set the FUNCTION selector on the distortion factor meter (d. f. m.) at SET REF LEVEL, the INPUT switch at HI Z, the INPUT range at 1 V -10 V and the METER RANGE switch at 0.1 V. Then adjust the SET REF (d. f. m.) level control to give a meter indication of 0 dB.

5. Set the TF 2015 FUNCTION selector at INT AM and adjust the SET MOD control to position the pointer of the MOD/CARRIER meter at the centre of the white box. Then set the AM DEPTH control at 30%.

6. Note the difference in level shown on the d.f.m., using a more sensitive range if necessary, and ensure that it is greater than 6 dB, i.e. deviation less than 500 Hz.

(Deviation in Hz = $\frac{1000}{d. f.m. \text{ voltage ratio}}$)

4.6.20 Carrier harmonics

Test equipment : item r.

1. Connect the RF OUTPUT of TF 2015 to the r.f. input of the spectrum analyser.

2. Set the FUNCTION selector at CW, the CAR-RIER switch at ON and the RF OUTPUT controls fully clockwise. Then ensure that the pointer of the MOD/CARRIER meter is within the white box.

3. Tune TF 2015 through each r.f. range and

check that the amplitude of any harmonic is greater than 26 dB down on the fundamental.

4.6.21 Coarse r.f. output control

Test equipment : items i, s.

Provided the attenuator pads have not been damaged (see Sect. 2.4) it is only necessary to prove correct operation of the associated microswitches and this can be satisfied by making a series of resistance measurements as follows :

1. With the SUPPLY switch at OFF, connect the multimeter set at low ohms range to the RF OUT-PUT socket.

2. Set both RF OUTPUT controls fully clockwise and check that the multimeter indicates a resistance of 1050 Ω +2% (resistor tolerance).

3. Position the COARSE control in turn to each dB setting and check that the measured resistance at the 10 dB position is 60 Ω and at 20 dB position 51 Ω and each following position 50 Ω ±2% (resistor tolerance).

4.6.22 Fine r.f. output control

Test equipment : items i, s.

1. Connect the power meter to the RF OUTPUT socket.

2. Set the SUPPLY switch at ON, the FUNCTION selection at CW, the CARRIER switch at ON. Tune TF 2015 to 100 MHz and check that the pointer of the MOD/CARRIER meter is within the white box.

3. Set the COARSE control at 0 and the FINE control at -7 dBm then note the reading obtained on the power meter.

4. Repeat check 3 with the FINE control in turn at 9, 11, 13, 15, 17 dBm.

4.7 CLEANING ROTARY SWITCHES

If it is necessary to clean the contacts of any rotary switches, this should be done with benzine or white spirit (not carbon tetrachloride) and the contacts should afterwards be wiped with a suitable lubricant such as a 1% solution of petroleum jelly in white spirit. Avoid lubricants containing soap or solid materials.



Repair

5.1 INTRODUCTION

Since the functional checks given in Chapter 4 serve to localize the cause of incorrect performance of the instrument, this chapter is intended to assist in tracing the fault to the particular part of the suspect circuit.

The information given, should be read with ference to Chapter 3 and 4 and the circuit diacrams contained within this manual.

CARRIER FREQUENCY : at 14 MHz RANGE 1 CARRIER switch : at ON FUNCTION selector : at INT AM RF OUTPUT controls : at 0 and -7 dBm

Prior to making measurements ensure that screws on Unit A12 are tight.

Voltage levels at A12 pin 12 are dictated by the a.l.c. voltage at A17 pin 5 and should approximate those given in Table 5.1.

Table 5.1

5.2 FAULT LOCATION

Some aid to fault finding is provided by the typical d.c. voltage and signal levels given in Sections 5.2.1, 5.2.2, 5.2.3 and by the fault tables included in each of the following sections. The tables are not extensive but are intended as a pointer to further investigation. It is to be emphasized that each fault table should be studied having regard for the others since incorrect operation of a circuit may be caused by malfunction of an associated circuit.

5.2.1. D.C. Voltages

Test equipment : item s.

Voltages given on the circuit diagrams and in Tables 5.1 and 5.2 approximate those which can be expected using a 20 k Ω/V meter on a typical TF 2015 connected to an a.c. supply of 240 V, 50 Hz and unless stated otherwise were measured with the controls of TF 2015 positioned as follows :

	rable 5.1	
Range TF 2015	Frequency MHz	Voltage (V d.c.) at A12 pin 12
1 –7	10 - 124	9-11
8	124	8.6
	177	9.7
9	177	9.2
	250	9,8
10	250	4.3
	320	8.0
	360	6.5
11	360	7.0
	410	9,2
	520	8.3

Voltage levels on Unit A23 which were measured with TF 2015 set in turn at 10 MHz and 14 MHz (RANGE 1) with CARRIER switch at ON and f.m. at 100 kHz deviation should approximate those given in Table 5.2.

Test point	Voltage	(V d.c.)	Meter range	Remarks
	10 MHz	14 MHz	(V)	
Junction R1/D1	4.7	7.5	100	
Junction C2/D1	0	7,5	100	
Junction R4/D2	0.2	0,25	2.5	negative with respect to chassis
Junction R2/D2	0.03	0.03	2.5	

Table 5.2

5.2.2. R.F. Signal levels

Test equipment : item g.

The r.f. signal levels listed in Table 5.3 approximate those which can be expected using the specified meter on a typical TF 2015 connected to Measurements were made with the TF 2015 FUNCTION selector at CW, the RF OUTPUT controls at 0 and -7 dBm and with the earth probe of TF 2603 connected to a position on the earth track on the printed board close to the test point.

an a.c. supply of 240 V, 50 Hz.

		Ą	lange 1	R	ange 11	
Unit	Test point	Signal level (mV) 14 MHz		Signal level (mV) 520 MHz		
		Carrie	er switch	Carrier switch		
		On	+ 6dB	On	+6dB	
A12	Pin 14	54	-	69	-	
A16	Pin 1	53	_	24	-	
	Junction R2/R4	9	-	28	· _	
	TR1e	2	-	1.8	_	
	TR1c	20	-	28	-	
	TR2e	1.40	-	145	-	
	TR3e	7	_	46	-	
	TR3c	37	-	38	-	
	TR4c	330	-	230	-	
	Junction D3/D5	175	180	94	65	
	Junction D3/D4	23	67	27	125	
	Junction D4/D6	170	180	10:3	158	
	Junction D5/D6	24 .	68	31	118	
	Junction C19/T2	45	125	15.5	77	
	TR5b	5.5	11	20	9e	-
	TR5c	80	260	84	310	
	TR7b	6.2	13.5	37	81	Ξ.
	TR7c	105	210		205	
	TR8b	9.4	19.5	9 6	200	
	TR8c	205*	400	265*	490	
	TR6b	67*	-	38*	-	
	TR6c	690*	_	155*	-	
	Pin 7	2 [¢]	2.50	8.5%	24¢	

Table 5.3

* SKF (COUNTER OUTPUT) loaded with 50 Ω .

1.4

Repair

5.2.3. A.F. Signal levels

Test equipment : item c.

The a.f. signal levels listed in Table 5.4. approximate those which can be expected on a

typical TF 2015 connected to an a.c. supply of 3 240 V, 50 Hz.

Unless stated otherwise, measurements were made with TF 2015 set at 14 MHz RANGE 1 and CARRIER switch at ON.

Probable cause

Table 5.4 Remarks Signal level Control settings Test point Unit FUNCTION - INT FM 6.7 V p-p Pin 3 A19 DEVIATION - 100 kHz negative peak compressed 3.2 V p-p Pin 7 6.4 V p-p Pin 18 negative peak compressed Pin 17 1 V p-p 3 V p-p Pin 9 FUNCTION - INT FM 0.5 V p-p Pin 15 A21 DEVIATION - 100 kHz 205 mV p-p carrier frequency 10 MHz Pin 13 FUNCTION - INT AM 850 mV p-p Pin 6 A21 negative peak compressed AM DEPTH - 80% 58 mV p-p Pin 8 negative peak compressed 70 mV p-p Pin 9

Table 5.5

	\$ 13143 ·	
(a)	Fuse FS1 blows when Instru- ment is switched (0).	Short in a.c. input circuit. A0 C1 or C2 short-or partial short-circuit.
(b)	No 22 V output	A0 TR1 open-circuit (check 22 V line for short) A0 C1, C2 or C5 short-or partial short-circuit.
(c)	No 19 V output	TR4 open-circuit (check 19 V line for short).
(d)	Inability to set 22 V +100 mV.	TR2, TR3 or TR5 faulty. Resistor open-circuit or value changed. C3 or C4 short-or partial short-circuit.
(e)	Incorrect 19 V ± 100 mV $output$.	Faulty TR5, TR6, TR8 Resistor open circuit or value changed especially R9, R14 or R17.
(f)	100 Hz supply on 33 V line greater than 100 µV r.m. H.	A0 C1, C2 or C3 open-circuit or low capacitance. D1, D2, D3 or D4 faulty.
(g)	10 V 1109	C8 open circuit or low capacitance.

5.2.4 Power supply

Circuit diagram Fig. 7.4 (Jult A2G

t.a.dt

5.2.5 RF calibrations

Circuit diagram Fig. 7.1 Unit A12 (sub-assemblies A1 to A11)

	Fault	Probable cause	
(a)	Calibrations incorrect all ranges.	Incorrect d.c. voltage at A19 pin 8 (Fig. 7.3). Check d.c. path to switch SG (Fig. 7.3).	
(b)	Inability to set scale calibra- tions on any selected range.	Incorrect d. c. voltage to R4 (A1 to A9) R5 (A10 and A11). Check output A21 pin 13. Faulty varactor D1 or D2.	

Table 5.7

5.2.6 RF output (normal)

Circuit diagram Fig. 7.2 Units A16, A17, A18

Fault

Meter pointer not within white box

Table 5.6

Probable cause

Incorrect d. c. outputs A18 pins 12 or 13. Check circuit A18 TR2. Faulty resistor or diode in circuit A18 R13 to R21, D4 or D5. Incorrect operation of circuit A17, R1 to R7, TR1, TR2 or TR3. DC offsets to A16 modulator from A17 TR7 and TR8 at incorrect level. Check circuit TR7 and TR8. Fault in amplifier circuit A16 TR5, TR7 and TR8. D7 or D8 faulty.

Fault in a d. c. circuit of A17. Chepk TR9, TR10, IC1 and A16 D9 and D10.

5.2.7 RF output (+6 dB)

Table 5.8

Probable cause

Fault

+6dB output not obtainable

Faulty carrier switch. Faulty diode A18 D6 or R25. Recheck operation of circuits A17 TR7, TR8 for increased d. c. offsets to A16 modulator and operation of a.l.c. loop with carrier switch at +6dB.

5.2.8 VSWR

Check that A16 R39 is 50 Ω ±1% and recheck A22 as given in Sect. 4.6.21.

Repair

5.2.9 Modulation oscillator

Circuit diagram Fig. 7.3 Unit A19

Table 5.9

Fault

(a) No output or incorrect output

(b) Frequency incorrect

(c) Distortion greater than 0.25%

Probable cause

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No 22 V at pin 1: check switch SA1F TR1 or TR2 faulty.

R1, R2, C1 or C2 faulty; check values.

Faulty thermistor R7. Check C3, C4.

5.2.10 Amplitude modulation

Circuit diagram Figs. 7.2 and 7.3 Units A16, A17, A18

Table 5.10

Fault

 (a) Inability to set meter pointer at centre of white box by adjusting SET MOD control.

- (b) Inability to obtain required modulation depth.
- (c) Envelope distortion

Probable cause

A0R2 faulty. Fault in monitor circuit A18 TR1.

Fault in a.m. driver circuit A17 TR4, TR5 or TR6. Incorrect operation of modulator : check A16 D3 to D6, L1 and L2. Check A16 C17, C18 and a.f. path from SET MOD control to driver.

Fault in modulation oscillator. Fault in a.m. driver circuit.

5.2.11 Frequency modulation

Circuit diagram Figs. 7.2 and 7.3 Units A19 and A23

Foult

Table 5.11

Probable cause

Check a.f. path from SET MOD control to A23. Incorrect operation of circuits A19 or A23. Switch SG faulty.

Faulty resistor A0 R5 to R15, A19 R18 to R22 or R17.

Incorrect operation of circuits A19 or A23.

No frequency modulation.

.

Incorrect deviation

FM distortion

Repair

5.2.12 ALC system

Circuit diagram Fig. 7.2 Units A16, A17, A18

Table 5.12

١

oscillator output,

Fault	Probable cause	
Meter pointer at zero or full	Fault in r.f. signal path. Check outputs from	
scale - all ranges.	circuits on A16. Fault in circuit A17 TR7. TR8.	
	Fault in circuit A17 TR9, TR10, IC1.	
	Fault in circuit A18 TR2 : check outputs.	
	Check diodes D4, D5 and associated resistors.	
- on one range only.	No input to A16 pin 1 : suspect appropriate	

Table 5.13

5.2.13 Carrier harmonics

Circuit diagram Fig. 7.2 Units A16 and A17

Fault

Amplitude too high - all ranges.

5.2.14 Spurious modulation

Fault

Spurious f. m. or a. m. too high.

5.3 REPLACING TUNING DRIVE CORD

Drive cord : 1 metre nylon cord MI code No. 16410 - 604.Tension spring : MI code No. 31119-017. To gain access to the drive cord assembly proceed as follows.

- Remove the case from the instrument and (1)remove the attenuator unit (see Sect. 4.3).
- (2)Remove the scale plates for the three highest frequency ranges.

Probable cause

Incorrect operation of first a.l.c. loop, causing high oscillator outputs to overload pre-amplifier. Amplifier distortion: check circuits TR1 to TR4 and TR5, TR7 and TR8.

Excess ripple on 22 V or 19 V d.c. supplies or increased circuit noise. Check transistors, diodes, and if necessary resistors especially those incorporated in a.l.c. loops.

Probable cause

- Set the RANGE switch to the 14.3 20.5 MHz (3)position.
- (4)Stand the instrument upside-down on the bench.
- (5) Referring to Fig. 5.1, slacken the grub screws and unscrew the nut to pull aside R38. With the TUNE control three steps away from the fully count- clockwise position, slacken one of the grub we we in the cord-drum boss.

Table 5.14


Fig. 5.1 Replacing cord drive

Turn the TUNE control fully clockwise and slacken the other grub screw. Remove the TUNE and EXTRA FINE TUNE knobs. Unscrew the nut and pull away the switch assembly SG from the drum.

- (6) Remove the drum and attach the drive cord as shown in Fig. 5.1, temporarily holding it in place with adhesive tape. Seal the cord knots with bakelite varnish.
- (7) Leplace the drum and completely refit the switch SG and R38 assemblies.
- (8) Fit the loop of cord around the three pulleys. Check that the cord spring is under tension. If the cord has been properly dimensioned, the expanded length of the spring will be about 12 mm (¹/₂ inch). Refit the knobs.
- (9) Attach the pointer to the drive cord and adjust its position so that it reaches (or just passes) the 60 MHz and 124 MHz calibration marks :

(a) On 60 - 86 MHz range with both TUNE and FINE TUNE controls fully counterclockwise.

(b) On 86 - 124 MHz range with both TUNE and FINE TUNE controls fully clockwise.

(10) Refit scale plates, attenuator unit and case.

5.4 REPLACING TRANSFORMERS A16 T1 T2

If modulator transformers T1 or T2 are required to be replaced it is important that the replacement be connected, in the same way as the removed component. The required connections are shown in Fig. 5.2



Fig. 5.2

5.5 ADDITIONAL INFORMATION

If further information is required please write or telephone Marconi Instruments Limited, Service Division - see address on back cover - or contact nearest representative, quoting the type and serial number on the data plate on rear of instrument.

If the instrument is being returned for repair please indicate clearly the nature of the fault or the work you require to be done. £



Introduction

Each sub-assembly or printed circuit board in this instrument has been allocated a unit identification in the sequence A0 to A12 and A16 to A23.

The complete component reference carries its unit number as a prefix e.g. A1C1 but for convenience in the text and on circuit diagrams the prefix is not used.

However, when ordering replacements or in correspondence the complete component reference must be quoted.

One or more of the components fitted in this instrument may differ from those listed in this chapter for any of the following reasons:

- (a) Components indicated by ⁺ have their value selected during test to achieve particular performance limits.
- (b) Owing to supply difficulties, components of different value or type may be substituted provided the overall performance of the instrument is maintained.
- (c) As part of a policy of continuous development, components may be changed in value or type to obtain detail improvements in performance.

When there is a difference between the component fitted and the one listed, always use as a replacement the same type and value as found in the instrument.

Ordering

When ordering replacements, address the order to our Service Division (address on rear cover) or nearest agent and specify the following for each component required.

- (1) Type* and serial number of instrument
- (2) Complete circuit reference
- (3) Description
- (4) MI code number
- as given on the serial number label at the rear of the instrument; if this is superseded by a model number label, quote the model number instead of the type number.

Component references

The components are listed in alpha-numerical order and the following abbreviations are used:

С	:	capacitor
Carb	:	carbon
Cer	:	ceramic
Cerm	:	cermet
D	:	semiconductor diode
Elec	:	electrolytic
\mathbf{FS}	:	fuse
IC	:	integrated circuit (package)
L	:	inductor
Max	:	maximum
ME	:	meter
Met	:	metal
Mic	:	mica
Min	:	minimum value
Ox	:	oxide
\mathbf{PL}	:	plug
Plas	:	plastic dielectric
R	:	resistor
S	:	switch
SK	:	socket
Т	:	transformer
ТΡ	:	terminal
\mathbf{TR}	:	transistor
WW	:	wirewound
х	:	crystal oscillator
+	:	value selected during test;
•		nominal value listed
Ø	:	feed-through component_
w	:	watts at 70°C

V : watts at 70°C

TUNING DIODES

When ordering replacement diodes D1 and D2 for oscillator units A1 to A11 it is necessary to ensure that diodes of the correct voltage/capacitance characteristic are supplied by following the instructions given below.

For instruments with serial numbers 116903/01 and onwards there is a label on Unit A12 bearing a letter and a number to identify the particular group of diodes used. To obtain the correct replacement you order should call for :

- (i) Type and serial number of instrument.
- (ii) Complete circuit reference e.g. A1D2.

- Description, i.e. Diode BB105A. (iii)
- MI code number 44529-003 plus Group (iv) identification code e.g. 44529-003/D3.

DETECTOR DIODES

Instruments with serial numbers up to 116904/100 are fitted with type CG92H diodes to units A16 and A17. A16 D1, D2 and A16 D9, D10 are matched pairs (d.c. and r.f.) and each of these pairs has a corresponding pair in A17 : D1, D2 and D4, D5 which has been selected to have a matching d.c. characteristic. Consequently when replacing a diode pair in A16 it is also necessary to replace the corresponding pair in A17. To obtain a suitable group of four diodes your order should specify :

- (i) Type and serial number of instrument.
- Complete circuit reference e.g. A16D1D2 + (ii) A17D1D2.
- (iii) Description : Matched and selected CG92H.
- (iv) MI code number : 44529-001 plus two selected 28321-161.

When replacing a diode pair in A17 it is sufficient to replace just these two diodes provided that the d.c. characteristic of the corresponding pair in A16 is ascertained, to allow a suitable pair to be selected. To do this :

Temporarily remove the connection to pins (1)4 and 3 (for D1, D2) or to pins 12 and 13 (for D4, D5) on A16.

Using only a GEC Selectest or Avometer (2)model 8 set at Ω range, measure the forward resistance of each diode by connecting the meter, in turn, between each pin and the junction of the two diodes.

Add the two resistance values so found, (3)e.g. 450 Ω + 450 Ω = 900 Ω_{\star}

Specify in your order :

- Type and serial number of instrument. (i)
- Complete circuit reference e.g. A17D1D2. (ii)
- Description : Selected CG92H. (iii)
- MI code number : two 28321-161 selected for (iv) combined resistance of (the value measured as described above).

Instruments with serial numbers 116905/001 and above are fitted with hot carrier diodes type 5082-2835, MI code 28349-006 which do not require selection and may be ordered straight forwardly.

Circuit	
reference	

Description

M.I. code

Unit (

L15

When ordering, prefix circuit reference with AO

C1	Elec 2200µF +50-10% 40V	26426-086
C2 to C25	Cer Ø 0.001µF +80-20% 300V	26373-733
C26	Cer Ø 50pF 10% 300V	26333-229
C27	Cer Ø 50pF 10% 300V	26333-229
C28	Cer Ø 50pF 10% 300V	26333-229
C29 to C35	Cer Ø 500pF +80-20% 300V	26373-732
C36	Cer Ø 50pF 10% 300V	26333-229
C37	Cer Ø 500pF +80-20% 300V	26373-732
C38	† Elec $22\mu F \pm 20\% 15V$	26486-583
D1	LED (On-Off indicator)	28624-103
FS1	160mA slow-blow (190-264 V)	23411-054
	or 250mA slow-blow (95-132 V)	
	Unider C. Dor	23416-191
FS2	500m A 1 1	23411-056
	Holdon for Egg	23416-191
L1	Choke 100µH	
L2	Chalas 100 tr	23642-561
L3	Cholto 100-44	23642-561
 L4	Obeles 100 It	23642-561
L5	Chalter 100, II	23642-561
 L6	Chalas 100 H	23642-561
L7	Chalse 100 H	3642-561
L8		3642-561
L9	Chalma 100 M	3642~561
L10	Challes 100 tr	3642-561
L11	Chalter 10.1	3642-561
L12	Chaire 10-11	3642-555
L13	Chalas 100 yr	3642-555
L14	Chalte 1000	8642-561
L15	Choke 10-11	8642-561
	23 23	642-555

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	rcuit rence Description		Circ		
L16		M.I. cod e	refere	nce Description	M.I. code
L17		23642-555		Met film 2001) 2% ‡W	24773-256
L18	100µ11	49116-909		Met film 180Ω 2% ‡W	24773-255
		49116-909) R33	Met film $160\Omega \ 2\% \ W$	24773-254
ME	l Meter		R34	Met film 130Ω 2% ¦W	24773-252
		44559-011	R35	Met film 1209 2% 4W	24773-251
R1	Cerm 1kΩ 10% 2W		R36	Met film $100\Omega 2\% \frac{1}{4}W$	24773-249
R2	Cerm 1kΩ 10% 2W	25725-403	R17 t	o R36 (inclusive) mounted o	
R3	Met film 1kß $2\% \frac{1}{4}W$	25725-403	R37	WW 100ks 10% 10-turn	
R 1	Cerm $1k\Omega \ 10\% \ 2W$	24773-273	R38	WW 10k Ω 10% $\frac{1}{2}$ W	25812-696
R5	Met film 2.2k Ω 2%	25725-403	$\mathbf{R}39$	Met film 500 1% 1/8W	24616-321
R6		24773-281	R40	Met film 200Ω 2% ‡W	24773-256
R7	Met film $1.5k\Omega 2\% \frac{1}{4}W$	24773-277	R41	Met film 8200 2% 4W	24773-271
R8	Met film 1.1kG 2 $\%$ $\frac{1}{4}$ W	24773-274	R42 -	† Met film 2.2kΩ 2% ‡W	24773-281
R9	Met film 750 $\Im \frac{1}{4}W$	24773-270	R43	Met film $470\Omega~2\%$ $rac{1}{4}$ W	24773-265
R10	Met film 510 Ω 2% $\frac{1}{4}$ W	24773-266	SA	Function switch	
R11	Met film $360 \Im 2\% \frac{1}{4} W$	24773-262	SB		44340-022
R11	Met film $240\Omega \ 2\% \ \frac{1}{4}W$	24773-258	SC	Carrier switch 4-pole 3-	
R12	Met film 1805 2% tw	24773 - 255	00	Deviation multiplier swit SPDT	ch 23462-252
R14	Met film 120 G 2% 4W	24773-251	SD1	Front wafer of carrier ra	
R14	Met film 130 $\Im 2\% \frac{1}{4}W$	24773-252		switch	44340-031
	Met film 150\$ 2% 4W	24773-253	SD2	Rear wafer of carrier rar	lge
	R15 (inclusive) mounted on RA	NGE switch SD	SE	switch	44340-030
R16	Cerm 10k5 10% 2W	25725 - 404	SF	Supply switch DPDT	23462-258
R17	Met film 1.1k Ω 2% $\frac{1}{4}W$	24773 - 274	01	Voltage selector switch 230V/115V	- 23467-155
R18	Met film 750 $\Im 2\% \frac{1}{4}$ W	24773-270	SG	Tune switch	
R19	Met film $680 \Omega \ 2\% \ \frac{1}{4}W$	24773-269	SKA	Bulgin P580	- 44340-015 22122 150
R20	Met film 560 $\Omega 2\% \frac{1}{4}$ W	24773-267	SKB	BNC 50 Ω (Ext mod in)	23423-159
R21	Met film 510 Ω 2% $\frac{1}{4}W$.	24773-266	SKC	BNC 50 Ω (Int mod out)	23443-443
R22	Met film $430 \Omega 2\% \frac{1}{4} W$	24773-264	SKD	BNC 50Ω (Sync in)	23443-443
R23	Met film $430\Omega \ 2\% \ \frac{1}{4}W$	24773-264	SKE	DIN speaker type (IF probe	23443-443
R24	Met film 390 Ω 2% $\frac{1}{4}$ W	24773-263		supply)	23435-252
R25	Met film 360Ω 2% 4 W	24773-262	SKF	TNC 50 Ω (counter out)	23444-741
R26	Met film $330\Omega \ 2\% \ \frac{1}{4}W$	24773-261	SKH	BNC 50Ω (RF out)	43129-029
R27	Met film $300\Omega \ 2\% \ \frac{1}{4}W$	24773-260			-
R28	Met film $300\Omega 2\% \frac{1}{4}W$	24773-260	T1	Mains transformer	43490-010
R29	Met film $240\Omega \ 2\% \ \frac{1}{4}W$	0	TP1	Battery terminal + input	
R30	Met film 2205 2% ‡W	0.1		Battery terminal - input	23235-176
-	For symbol.	s and abbreviations see		to this chapter	23235-176

For symbols and abbreviations see Introduction to this chapter

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Circ			Circuit		
refere TR1		M.I. code	reference	e Description	M.I. code
1.1.1	Transistor 2N3055	28456-567	R3	Met film 1kΩ 2% 4W	24773-273
X1 to			R4	Met film $10k\Omega \ 2\% \ \frac{1}{4}W$	24773-297
X40	Ferrite bead	41372-006	R5	Met film 100Ω 2% ‡W	24773-249
			R6 †	Met film $22k\Omega \ 2\% \ \frac{1}{4}W$	24773-305
Unit	(AI) oscillator range 1				
When	ordering, prefix circuit reference with A1		_ T1	Coil assembly	44290-080
C1	Cer 0.01µF +80-20% 100V	26383-055	TR1	Transistor WN428	28459-025
C2 -	Cer 10pF ±0.5pF 500V	26343-121			
C3	Cer 220pF 20% 500V	26383-134	Unit (A3) oscillator range 3	
C4	Cer vār. 0.5 - 5pF	26845-141	When ord	lering, prefix circuit reference with A3	
	-		C1 ·	Cer 0.01µF +80-20% 100V	26385-055
D1	Tuning varactor BB105A	See p.41	C2 †	Cer 15pF ±0.5pF 500V	26343-127
D2	Tuning varactor BB105A	tuning diodes	C3	Cer 220pF 20% 500V	26383-134
			C4	Cer var. 0.5 - 5pF	26845-141
R1	Met film 300kΩ 2% ¼W	24773-321			
R2	Met film 300k Ω 2% $\frac{1}{4}$ W	24773-321	D1	Tuning varactor BB105A)	
R3	Met film 1kΩ 2% ¼W	24773-273	D2	Tuning varactor BB105A	See p.41 tuning diodes
$\mathbf{R4}$	Met film $10k\Omega \ 2\% \ \frac{1}{4}W$	24773-297			0
R5	Met film $100\Omega \ 2\% \ \frac{1}{4}W$	24773-249	R1	Met film 300kΩ 2% ¼W	24773-332
R6 +	Met film $22k\Omega \ 2\% \ \frac{1}{4}W$	24773-305	R2	Met film 300 k $\Omega 2\% \frac{1}{4}$ W	24773-332
			R3	Met film 1kΩ 2% JW	24773-273
Т1	Coil assembly	44290-079	$\mathbf{R4}$	Met film $10k\Omega 2\% \frac{1}{4}W$	24773-297
TR1	Transistor WN428	28459-025	R5	Met film $100\Omega 2 \frac{9}{6} \frac{1}{4} W$	24773-249
Unit	2) oscillator range 2		T1	Coil assembly	44290-081
When ord	ering, prefix circuit reference with A2		111 To #		
C1	Cer 0.01µF +80-20% 100V	26383-055	TR1	Transistor WN428	28459-025
C2 †	12pF 10% 500V	26343-125	Unit (A) oscillator range 4	
C3	Cer 220pF 20% 500V	26383-134		ing, prefix circuit reference with A4	
C4	Cer var. 0.5 - 5pF	26845-141		Cer 0.001µF +80-20% 500V	26383-242
5.	_			Cer 10pF ±0.5pF 500V	26343-121
DI	Tuning varactor BB105A	See p.41		Cer 220pF 20% 500V	26383-134
D2	Tuning varactor BB105A) t	uning diodes		Cer var. 0.5 - 5pF	26845-141
D 1				-	AIL
R1	Met film $300k\Omega \ 2\% \ \frac{1}{4}W$	24773-332	D1	Tuning varactor BB105A	See p.41
R2	Met film $300k\Omega \ 2\% \ \frac{1}{4}W$	24773-332	D2	Funing varactor BB105A	tuning diodes
	For symbo	ls and abbreviations re	المالة المسادية	A 111 A	

For symbols and abbreviations see introduction to this chapter

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	Circuit Ference Description	M.I. code	Circu referei		.
R1	Met film 300kΩ 2% ¼W			- ····	M.I. code
R2	_	24773-332	C3	Cer 220pF 20% 500V	26383-134
R3	-	24773-332	C4	Cer var. 0.5 - 5pF	26845-141
R4		24773-273 24773-297			
R5		24773-249	D1	Tuning varactor BB105A	See p.41
		24113-249	D2	Tuning varactor BB105A) tuning diodes
T1	Coil assembly	44290-082		_	
		11230-082	R1	Met film 300 k $\Omega 2\% \frac{1}{4}$ W	24773-332
TR	1 Transistor WN428	28459-025	R2	Met film $300k\Omega 2\% \frac{1}{4}W$	24773-332
		20409-023	R3	Met film 1kΩ 2% ¼W	24773-273
Uni	it (A5) oscillator range 5		R4	Met film $10k\Omega 2\% \frac{1}{4}W$	24773-297
	n ordering, prefix circuit reference with A5		R5	Met film $100\Omega \ 2\% \ \frac{1}{4}W$	24773-249
	the string profix circuit reference with A5				
C1	Cer 0.001µF +80-20% 5001	V 26383-242	T 1	Coil assembly	44290-084
C2	† Cer 10pF ±0.5pF 500V	26343-121			
C3	Cer 220pF 20% 500V	26383-134	TR1	Transistor WN428	28459-025
C4	Cer var. 0.5 - 5pF	26845-141		~	
			Unit	(7) oscillator range 7	
D1	Tuning varactor BB105A)	See p.41	When ord	ering, prefix circuit reference with A7	
D2	Tuning varactor BB105A	tuning diodes	C1	Cer 680pF +40-20% 500V	26383-141
			C2 †	Cer 7.5pF ±0.5pF 500V	26343-117
R1	Met film 300kΩ 2% ‡W	24773-332	C3	Cer 220pF 20% 500V	26383-134
$\mathbf{R2}$	Met film 300 k Ω 2 $\%$ $\frac{1}{4}$ W	24773-332	C4	Cer var. 0.5 - 5pF	26845-141
R3	Met film 1kΩ 2% ‡W	24773-273			
R4	Met film 10kΩ 2% ¼W	24773-297	D1	Tuning varactor BB105A]	See
R5	Met film $100\Omega \ 2\% \ \frac{1}{4}W$	24773-249	D2	Tuning varactor BB105A	_See p.41 tuning diodes
T1	Coil assembly	44290-083	R1	Met film 300k£ 2% 4 W	24773-332
			R2	Met film 300kΩ 2% ±W	24773-332
TRI	Transistor WN428	28459-025	R3	Met film $1k\Omega 2\% \frac{1}{4}W$	24773-273
			R4	Met film 10kΩ 2% ±W	24773-297
Unit (A6) oscillator range 6		R5	Met film $100\Omega 2\% \frac{1}{6}W$	24773-249
	dering. prefix circuit reference with A6				
			T1	Coil assembly	44290-085
C1	Cer 0.001µF +80-20% 500V	26383-242		;	11230-000
C2 -	† Cer 7.5pF ±0.5pF 500V	26343-117	TR1	Transistor WN429	28459-025

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Circui referenc		M.I. code	Circuit reference	e Description	M.I. code
Unit (A8) oscillator range 8		T1	Coil assembly	44290-087
When or	dering, prefix circuit reference with A8			y	
C1	Cer 470pF +40-20% 500V	26383-139	TR1	Transistor WN428	28459-025
C2 †	•	26343-125			
C3	Cer 220pF 20% 500V	26383-134	Unit (A		
C4	Cer var. 0.5 - 5pF	26845-141	When ord	ering, prefix circuit reference with A10	
D1	. Tuning Paractor BB105A)	See p. 41	C1	Cer 220pF 20% 500V	26383-134
Pa	Tuning varactor BB105A	tuning diodes	C2	Cer 2.2pF ±0.5% 500V	26343-110
	-		C3	Cer 100pF 20% 500V	26343-167
R1	Met film 300k Ω 2% $\frac{1}{4}W$	24773-332	C4	Cer var. 0.5 - 5pF	26845-141
R2	Met film 300k $\Omega \ 2\% \ \frac{1}{4}W$	24773-332	C5 .	Cer 68pF 20% 500V	26343-163
R3	Met film $1k\Omega \ 2\% \ \frac{1}{4}W$	24773-273	C6	Cer 2.2pF ±0.25% 750 V	26324-011
R4	Met film $10k\Omega 2\% \frac{1}{4}W$	24773-297		•	
R5	Met film $100\Omega \ 2\% \ \frac{1}{4}W$	24773-249	D1	Tuning varactor BB105A }	See p.41
			D2	Tuning varactor BB105A)	tuning diodes
T1	Coil assembly	44290-086	D3	Diode 1N4148	28336-076
TR1	Transistor WN428	28459-025	R1	Met film 4.7k Ω 2% $\frac{1}{4}W$	24773-289
		20100 020	R2	Met film $10k\Omega 2\% \frac{1}{4}W$	24773-297
Unit ()	A9) oscillator range 9		R3	Met film $10k\Omega \ 2\% \ \frac{1}{4}W$	24773-297
	ering, prefix circuit reference with A9		$\mathbf{R4}$	Met film 2.2k Ω 2% $\frac{1}{4}W$	24773-281
			R5	Met film $10k\Omega \ 2\% \ W$	24773 - 297
C1	Cer 330pF 20% 500V	26383-136	R6	Met film $100\Omega 2\% \frac{1}{4}W$	24773 - 249
C2 †	^	26343-121	R7	Carb 22Ω 5% ¼W	24312-788
C3	Cer 220pF 20% 500V	26383-134	L1	Oscillator inductor	
C4	Cer var. 0.5 - 5pF	26845-141			
			TR1	Transistor BFR90	28452-167
D1	Tuning varactor BB105A	See p. 41			
D2	Tuning varactor BB105A)	tuning diodes			
			When orde	ring, prefix circuit reference with A11	
R1	Met film $300k\Omega 2\% \frac{1}{4}W$	24773-332	C1	Cer 100pF 20% 500V	26343-167
R2	Met film $300k\Omega \ 2\% \ \frac{1}{4}W$	24773-332	C2 †	Cer 47pF 20% 500V	26343-113
R3	Met film $1k\Omega \ 2\% \ \frac{1}{4}W$	24773-273	C3	Cer 100pF 20% 500V	26343-167
R4	Met film $10k\Omega 2\% \frac{1}{4}W$	24773-297	C4	Cer var. 0.5 - 5pF	26845-141
R5	Met film 100Ω 2% ¼W	24773-249	C5	Cer 68pF 20% 500V	26343-163

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For symbols and abbreviations see introduction to this chapter

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refe	rcuit rence Description	M.I. code	Circu refere		M.I. code
D1	Tuning varactor BB105A	See p. 41	DI	Diode 1N3062	28336-248
D2	Tuning varactor BB105A \int	tuning diodes	D2	Diode 1N3062	28336-248
R1			D3	Diode 1N3062	28336~248
R2	Met film $4.7k\Omega 2\% \frac{1}{4}W$	24773-289	D4	Diode 1N3062	28336-248
R3	Met film $10k\Omega 2\% \frac{1}{4}W$	24773-297	D5	Diode 1N3062	28336-248
R4	Met film $10k\Omega 2\% \frac{1}{4}W$	24773-297	D6	Diode 1N3062	28336-248
R5	Met film $1k\Omega \ 2\% \ \frac{1}{4}W$	24773-273	D7	Diode 1N3062	28336-248
R6	Met film $10k\Omega 2\% \frac{1}{4}W$	24773-297	D8	Diode 1N3062	28336-248
R7	Met film $100\Omega \ 2\% \ \frac{1}{4}W$	24773-249	D9	Diode 1N3062	28336-248
R/	Carb 22 Ω 5% $\frac{1}{4}$ W	24312-788	D10	Diode 1N3062	28336-248
L1	Oggilleteret		D11	Diode 1N3062	28336-248
	Oscillator inductor	-	D12	Diode 1N3062	28336-248
TR1	Transistor BFR90	28452-167	D13	Diode 1N4148	28336-076
			R1	Met film 100k Ω 2% $\frac{1}{4}$ W	24773-321
Unit	(A12) mother board		R2	Met film $1.8k\Omega 2\% \frac{1}{4}W$	24773-279
	ordering, prefix circuit reference with A12		R3	Met film $100\Omega \ 2\% \ \frac{1}{4}W$	24773-249
			R4	Met film $100\Omega 2\% \frac{1}{4}W$	24773-249
C1	Cer 0.01µF +80-20% 100V	26383-055	R5	Met film 100Ω 27 4W	24773-249
C2	Cer 0.01µF +80-20% 100V	26383-055	R6	Met film $100\Omega 2\% \frac{1}{4}W$	24773-249
C3	Cer 0.01µF +80-20% 100V	26383-055	$\mathbf{R7}$	Met film $100\Omega 2\% \frac{1}{6}W$	24773-249
C4	Cer 0.001µF +80-20% 500V	26383-242	R8	Met film $100\Omega 2\% \frac{1}{4}W$	24773-249
C5	Cer 0.001µF +80-20% 500V	26383-242	R9	Met film 100 2 2 1 W	24773-249
C6	Cer 0.001µF +80-20% 500V	26383-242	R10	Met film 100Ω 2% W	24773 - 249
C7	Cer 0.001µF +80-20% 500V	26383-242	R11	Met film $100\Omega \ 2\% \frac{1}{4}W$	24773-249
C8	Cer 100pF 20% 500V	26343-167	R12	Met film $100\Omega 2\% \frac{1}{3}W$	_
C9	Cer 68pF 20% 500V	26343-163	R13	Met film $100\Omega 2\% \frac{1}{4}W$	- 24773-249
C10	Cer 470pF +40-20% 500V	26383-139			24773-249
C11	Cer 100pF 20% 500V	26343-167	X1	Ferrite bead	41950 000
C12	Cer 0.01µF +80-20% 100V	26383-055	X2	Ferrite bead	41372-006
C13	Cer 0.01µF +80-20% 100V	26383-055			41372-006
C14	Cer 0.001µF +80-20% 500V	26383-242	Unit (A	b pre-amplifier, modulato	r counter
C15	Cer 0.001µF +80-20% 500V	26383-242	amplifie	r, wide band amplifier	, councer
C16	Cer 0.001µF +80-20% 500V	26383-242		ing, prefix circuit reference with A16	
C17	Cer 0.001µF +80-20% 500V	26383-242	Cl	Cer 0.001µF +80-20% 500V	96999 040
C18	Cer 0.001µF +80-20% 500V	26383-242		Cer 0.01µF +80-20% 100V	
C19	Cer 470pF +40-20% 500V	26383-139		Cer 0.001 μ F +80-20% 500V	26383-055
C20	Cer 470pF +40-20% 500V	26383-139		Cer 0.001 μ F +80-20% 500V	26383-242
	For symbol	s and abbreviations s		n to this chapter	26383-242
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Circu			Circuit
referer C5	Cer 10pF 20% 500V	M.I. cod e 26343-121	reference Description M.I. code C42 Cer 0.01µF +80-20% 100V 26383-055
C6	Cer 47pF 20% 500V	26343-113	
C7	Cer 0.01µF +80-20% 100V	26383-055	
C8	Cer 0.01µF +80-20% 100V	26383-055	
C9	Cer 0.001µF +80-20% 500V	26383-242	C45 T Cer 2.2pF 10% 63V 26582-418 C46
C10	Cer 0.01µF +80-20% 100V	26383-055	to \uparrow Cer 0.001 μ F +80-20% 500V 26383-242
C11	Cer 0.01µF +80-20% 100V	26383-055	C49
C12	Cer 1pF ±0.5pF 500V	26343-101	C50 Cer 2.7pF ±0.5pF 500V 26343-109
C13	Cer 0.01µF +80-20% 100V	26383-055	C51 † Cer 3.3pF ±0.5% 500V 26343-111
C14	Cer 0. <u>0</u> 01µF +80~20% 500V	26383-242	
C15	Cer 0.001µF +80-20% 500V	26383-242	D1, D2 See p. 42, detector diodes
C16	Cer 0.001µF +80-20% 500V	26383-242	D3-D6 Matched set of four diodes 44529-007
C17	Cer 0.001µF +80-20% 500V	26383-242	D7, D8 PIN diode 28383-999
C18	Cer 0.001µF +80-20% 500V	26383-242	D9, D10 See p.42, detector diodes
C19	Cer $0.001 \mu F$ +80-20% 500V	26383-242	D11-D14 Matched sct of four diodes 44529-007
C20	Cer 0.1 μ F +50-25% 30V	26383-031	
C21 -	+ Cer 3.3pF ±0.5pF 63V	26343-459	L1 Inductor 100µH 10% 23642-561
C22	Cer 0.001µF +80-20% 500V	26383-242	L2 Inductor 100µH 10% 23642-561
C23	Cer 0.01 μ F +80-20% 100V	26383-055	L4 Inductor 33µH 10% 23642-558
C24	Cer 0.01µF +80-20% 100V	26383-055	R1 Met film $82\Omega 2\% \frac{1}{4}W$ 24773-247
C25	Cer 0.001µF +80-20% 500V	26383-242	
C26	Cer 0.01µF +80-20% 100V	26383-055	
C27	Cer 0.001 μ F +80-20% 500V	26383-242	
620	Cer 220pF 20% 500V	26383-134	
C29	Cer 0.001µF +80-20% 500V	26383-242	
C30	Cer 0.01µF +80-20% 100V	26383-055	
C31	Cer 0.001µF +80-20% 500V	26383-242	R7Met film $250\Omega 2\% \frac{1}{4}W$ $24773-259$ R8Met film $200\Omega 2\% \frac{1}{4}W$ $24773-256$
C32	Cer 0.01µF +80-20% 100V	26383-055	R9 $+$ Met film $39\Omega \ 2\% \ \frac{1}{4}W$ 24773-239
C33	Cèr 0.001µF +80-20% 500V	26383-242	R10Met film 100Ω 2 $\% \frac{1}{4}$ W24773-249
C34	Cer 0.001µF +80-20% 500V	26383-242	R11 Met film $22\Omega 2\% \frac{1}{4}W$ 24773-233
C35	Cer 0.001µF +80-20% 500V	26383-242	R12 Met film $270\Omega \ 2\% \ \frac{1}{4}W$ 24773-259
C36	Cer 68pF 20% 500V	26343-163	R13 Met film $120\Omega 2\% \frac{1}{4}W$ 24773-251
C37	Cer 82pF 20% 500V	26383-134	R14 Met film $270\Omega \ 2\% \frac{1}{4}W$ 24773-259
C38	Cer 0.001µF +80-20% 500V	26383-242	R15 Met film $47\Omega 2\% \frac{1}{4}W$ 24773-241
C39	Cer 18pF 20% 500V	26343-129	R16 Met film $150\Omega 2\% \frac{1}{4}W$ 24773-253
C40	Cer 18pF 20% 500V	26343-129	R17 Met film $15k\Omega 2\% \frac{1}{4}W$ 24773-301
C41	Cer 0.001µF +80-20% 500V	26383-242	R18 Met film $15k\Omega 2\% \frac{1}{4}W$ 24773-301
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For symbols and abbreviations see introduction to this chapter

Ciri refer	cuit ence	Description	M.I. code		rcuit erence	Description	M.I. code
R19	Met	film $15\Omega \ 2\% \ \frac{1}{4}W$	24773-229	Un	it (A17)	ALC and AM driver	
R20	Met	film $82\Omega \ 2\% \ \frac{1}{4}W$	24773-247		\cup		
R21	Met	film 200 Ω 2% $\frac{1}{4}$ W	24773-256			g. Prefix circuit reference with A17	
R22	† Met	film 68k Ω 2% $\frac{1}{4}$ W	24773-317	. C1	C	Cer 0.01µF +80-20% 100V	
R23	Met	film 1k Ω 2% $\frac{1}{4}W$	24773-273	C2		Cer 0.1 μ F +50-20% 30V	26383-055
R24	Met	film $470\Omega \ 2\% \ \frac{1}{4}W$	24773-265	C3		Elec 22μ F +100-20% 25V	26383-031
R25	Met	film 5.6kΩ 2% ‡ W	24773-291	C4		Elec 47μ F +100-20% 40V	26415-805
R26	Met	film 6.8k Ω 2% $\frac{1}{4}$ W	24773-293	C5		lec 47μ F +100-20% 40V	26415-810
R27	Met	film $200\Omega 2\% \frac{1}{4}W$	24773-256	C6		lec 10μ F +100-20% 63V	26415-810
R28	Met d	film 390Ω 2% ‡ W	24773-263	C7		las 2.2 μ F 10% 63V	26415-802
R29	Met f	film 15Ω 2% ‡ W	24773-229	C8		er 100pF 20% 500V	26582-418
R30	† Met f	film 56k Ω 2% $\frac{1}{4}$ W	24773-315	C 9		er 100pF 20% 100V	26343-167
R31	Met f	ilm 1kΩ 2% 4 W	24773-273	C10		er 0.01µF +80-20% 100V	26343-167
R32	Met f	ilm 82 Ω 2% $\frac{1}{4}$ W	24773-247	C11		lec 22μ F 20% 15V	26383-055
R33	Met f	ilm 200 Ω 2 $\%$ $\frac{1}{4}$ W	24773-256	C12		er 100pF 20% 500V	26486-583
R34	† Met f	ilm $12k\Omega 2\% \frac{1}{4}W$	24773-299	C13		er 100pF 20% 500V	26343-167
R35	Met fi	ilm 430Ω 2% 1 W	24773-264	C14		as 5500pF 2% 125V	26343-167
R36	Met fi	ilm $82\Omega \ 1\% \ \frac{1}{4}W$	24762-569		·	1 0 0 0 0 0 1 0 0 1 2 0 V	26516-666
R37	Met fi	ilm 1kΩ 2% ¼W	24773-273				
R38	Met fi	Im 560Ω 2% ‡W	24773-267	D1 }	Se	e p.42, detector diodes	
R40	Met fi	$\operatorname{Im}\ 22\Omega\ 2\%\ \frac{1}{4}\mathrm{W}$	24773-233.	D2)			
R41 -	† Met fi	lm 68Ω 2% ∃W	24773-245	D3	Di	ode 1N4148	28336-676
R42	Met fi	$\lim 390 \Omega 2\% \frac{1}{4} W$	24773-263	D4 }	See	p.42, detector diodes	
T1	Transf	lormer	43590-009	D5)			
T2	Transf	ormer	43590-009	101			•
TR1	Transi	stor BFR90	28452-167	IC1	$\mu \Lambda$	741 (operational amplifier)	28461-304
TR2	Transi	stor BFR90	28452-167	R1	Ма		
TR3	Transi	stor BFY90	28452-157	R2		t film $1k\Omega 2\% \frac{1}{4}W$	24773-273
TR4	Transis	stor BFYJ0	28452-157	R3		t film $1k\Omega 2\%$ W	24773-273
TR5	Transis	stor BFR90	28452-167	R4		film 4.7k Ω 2% $\frac{1}{4}W$	24773-289
TR6	Transis	stor BFY90	28452-157	R5		film $10k\Omega 2\% \frac{1}{4}W$	24773-297
TR7	Transis	stor BFY90	28452-157	R6 +		b var. $220\Omega \ 10\% \ \frac{1}{2}W$	25711-546
TR8		tor BFY90	28452-157	R6 R7 +		film $560\Omega 2\% \frac{1}{4}W$	24773-267
				R7 T R9		film $390\Omega 2\% \frac{1}{4}W$.	24773-261
X3	Cerami	c bead	23213-146	R10			24773-305
		•		1/10	met	film $10k\Omega 2\% \frac{1}{4}W$	24773-297

For symbols and abbreviations see introduction to this chapter

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Circuit reference	Description	M.I. cod e	Circuit reference	Description	M.I. code
R12	Met film 5.6k Ω 2% $\frac{1}{4}$ W	24773-291	TR8	Transistor BCY70	28434-857
R13	Met film $15k\Omega \ 2\% \ \frac{1}{4}W$	24773-301	TR9	Transistor BCY70	28434-857
R14	Met film $1k\Omega 2\% \frac{1}{4}W$	24773-273	TR10	Transistor BCY70	28434-857
R15	Met film $430\Omega \ 2\% \ \frac{1}{4}W$	24773-264			
R16	Met film 2.7k Ω 2% $\frac{1}{4}$ W	24773-283	-		
R17	Met film $1k\Omega \ 2\% \ \frac{1}{4}W$	24773-273	Unit (All	8) Monitors	
R18	Met film $430\Omega \ 2\% \ \frac{1}{4}W$	24773-264	When order	ing, prefix circuit reference with A18	
R19	Met film 15kΩ 2% ‡ W	24773-301	C1	$D_{10} = 0.47 \times T.10^{10}$ (201)	0.0500 410
R20	Met film 2.2k Ω 2% $\frac{1}{4}$ W	24773-281	C1	Plas 0.47µF 10% 63V	26582-410
- ?1	Met film 4.3k $\Omega \ 2\% \ \frac{1}{4}W$	24773-288	C2	Elec 22μ F +100-20% 25V	26415-805
R22	Met film 3.3k Ω 2% $\frac{1}{4}$ W	24773-285	C3	Plas 0.47µF 10% 63V	26582-410
R23	Met film $36k\Omega \ 2\% \ \frac{1}{4}W$	24773-310	DI	Diada 1N4140	00000 070
R24	Met film $36k\Omega \ 2\% \ \frac{1}{4}W$	24773-310	D1	Diode 1N4148	28336-676
R25	Met film 8.2k Ω 2% $\frac{1}{4}$ W	24773-295	D2 D3	Diode 1N4148 Diode 1N4148	28336-676
R26	Met film 8.2k Ω 2% $\frac{1}{4}$ W	24773-295	D3 D4	Diode 1N4148	28336-676
R27	Met film $120k\Omega \ 2\% \ \frac{1}{4}W$	24773-323	D 1 D5	Diode 1N4148	28336-676
R28	Met film $12k\Omega \ 2\% \ \frac{1}{4}W$	24773-299	D5 D6	Diode 1N4148	28336-676
R29	Met film $27k\Omega \ 2\% \ \frac{1}{4}W$	24773-307	D0	Didde 114146	28336-676
R30	Met film $47k\Omega \ 2\% \ \frac{1}{4}W$	24773-313			
R31	Met film $1.5 \mathrm{k}\Omega \ 2\% \ \frac{1}{4} \mathrm{W}$	24773-277	RI	Met film $1.1 \mathrm{k}\Omega \ 2\% \ \frac{1}{4} \mathrm{W}$	24773-274
R32	Met film 6.2k Ω 2% $\frac{1}{4}$ W	24773-292	R2 +	Met film $3.3k\Omega 2\% \frac{1}{4}W$	24773-285
R33	Met film $680\Omega \ 2\% \ \frac{1}{4}W$	24773-269	R3	Carb var 2.2kΩ 20% 0.21W	25711-547
)4	Met film $2k\Omega \ 2\% \ \frac{1}{4}W$	24773-280	R4	Met film 4.7k Ω 2% $\frac{1}{4}$ W	24773-289
R35	Met film 5.6k Ω 2% $\frac{1}{4}$ W	24773-291	R5	Carb var 10k Ω 10 $\% \frac{1}{2}W$	25711-543
R36	Met film 15k Ω 2% $\frac{1}{4}$ W	24773-301	R6	Met film $27k\Omega \ 2\% \frac{1}{4}W$	24773-307
R37	Met film $27k\Omega \ 2\% \ \frac{1}{4}W$	24773-307	R7	Met film $220k\Omega 2\% \frac{1}{4}W$	24773-329
R38	Met film $27k\Omega \ 2\% \ \frac{1}{4}W$	24773-307	R8	Met film 2.2k Ω 2% $\frac{1}{4}W$	24773-281
R39	Carb 470kΩ 5% 1/8W	24311-937	R9	Thermistor 1.5k Ω (CZ3)	25683-644
R40	Carb 470kΩ 5% 1/8W	24311-937	R10	Met film $680\Omega 2\% \frac{1}{4}W$	24773-269
(UD 1			R11	Met film $1.5k\Omega 2\% \frac{1}{4}W$	24773-277
TR1	Transistor BCY71	28435-235	R12	Met film 10ks2 % 4W	24773-297
TR2	Transistor BCY71	28435-235	R13	Met film $3.3k\Omega \ 2\% \ \frac{1}{4}W$	24773-285
TR3	Transistor BC109	28452-777	R14	Met film 4.7k Ω 2% $\frac{1}{4}$ W	24773-289
TR4	Transistor BC107	28455-437	R15	Met film 4.3k Ω 2% $\frac{1}{4}W$	24773-288
TR5	Transistor BC107	28455-437	R16	Met film 3.9k $\Omega \ 2\% \ \frac{1}{4}W$	24773-287
TR6	Transistor BC107	28455-437	R17	Carb var $47k\Omega \ 10^{\prime\prime}_{\prime\prime0} \ 5W$	25711-549
TR7	Transistor BC107	28455-437	R18	Met film $5.1k\Omega 2\% \frac{1}{4}W$	24773-290

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For symbols and abbreviations see introduction to this chapter

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	Circuit ference Description	M.I. code	Circ refere		
R1	9 Met film 2.2k Ω 2% $\frac{1}{4}W$			Deteription	M.I. code
R2		24773 - 281 24773 - 295		Met film $82k\Omega$ $2\% \frac{1}{4}W$	24773 - 319
R2				Met film $220k\Omega \ 2\% \ \frac{1}{4}W$	24773-329
R2:		24773-297		Met film $10 \mathrm{k}\Omega \ 2\% \ \frac{1}{4} \mathrm{W}$	24773-297
R2		25711-546		Met film $100\Omega \ 2\% \ 4W$	24773-249
R2		25711-544		Met film 100요 2% ╁W	24773-249
R25		24773-301		Met film $100 \Omega \ 2\% \ 4W$	24773-249
R26		24773-297	R21	Met film $100\Omega \ 2\% \ W$	24773 - 249
	- /0 4 ***	24773-281	R22	5- mint 2103t 2.6 4 W	24773-258
TR1	Transistor BC108		R23 †	and a streng a to 1 th	24773-305
TR2		28452-787	R25	Met film $220\Omega \ 2\% \ \frac{1}{6} W$	24773-257
		28452 - 777	R26	Met film $9.1 \mathrm{k\Omega} \ 2\% \ \mathrm{IW}$	24773-296
Uni	t (AI9) modulation oscillator a	nd FMA 11	R27	Met film $10 \mathrm{k}\Omega \ 2\% \ \mathrm{4W}$	24773-297
When	ordering, prefix circuit reference with A19	nd FM drive	R28	Met film 2.2k Ω 2% $\frac{1}{2}W$	24773-281
			R29 +	Met film 220 Ω 2% $\frac{1}{4}W$	24773-257
C1	Plas 0.018µF 2% 125V	26516-786	R::0 +	Met film $270\Omega/2^{+}_{-6}$ $\frac{1}{4}W$	24773-259
C2	Plas 0.018µF 2% 125V	26516-786	TR1	Transistor BC108	
C3	Elec $22\mu F$ +100-20% 25V	26415-805	TR2	Transistor BCY71	28452-787
C4	Cer 0.047 μ F +50-20 $\%$ 30V	26383-018	TR3	Transistor BCY71	28435-235
C5	Elec 22µF +100-20% 25V	26415-805	TR4		28435-235
C6	Plas 0.47µF 10% 63V	26582-410	TR5	Transistor BC108	28452-787
C7	Elec 100µF 20% 10V	26486-606	110	Transistor BC103	28452-787
C8	Elec 100µF 20% 10V	26486-606			
39	Elec 22µF 20% 15V	26486-583	Unit A	//	
			vyhen orde.	ring, prefix circuit reference with A20	
R1	Met film $18k\Omega \ 2\% \ W$	24773-303	C2	Elec 100µF +100-20% 25V	26423-244
R2	Met film 18k Ω 2% $\frac{1}{4}$ W	24773-303	C3	Cer 0.0047µF +80-20% 100V	26383-321
R3	Met film 2.2k Ω 2% $\frac{1}{4}$ W	24773-281	C4	Tant 4.7µF 20% 35V	26486-219
$\mathbf{R4}$	Met film 680 Ω 2% $\frac{1}{4}$ W	24773-269	C5	Elec 100µF +100-20% 25V	26423-243
R5	Met film $22k\Omega$ 2% $\frac{1}{4}W$	24773-305		Cer 0.01µF +100-20% 100V	26383-055
$\mathbf{R6}$	Met film 9.1k Ω 2% $\frac{1}{4}W$	24773-296		Plas 1µF 10% 63V	26582-414
$\mathbf{R7}$	Thermistor $20 \mathrm{k}\Omega$ (RA54) 20%	25683-387		Plas 2.2µF 10% 63V	26582-418
RS	Met film $1k\Omega \ 2\% \ \frac{1}{4}W$	24773-273		Cer 0.1µF +50-20% 30V	26383-031
R9	Met film 100 k $\Omega 2\% \frac{1}{4}$ W	24773-321			
R10	Met film $560\Omega 2\% \frac{1}{4}W$	24773-267		Diode 1N4004	28357-028
R11	Met film $6.8k\Omega 2\% \frac{1}{4}W$	24773-293		Diode 1N4004	28357-028
R12	Carvar $2 5 + 0.00\%$ o com	-***3-293	D3 I	Diode 1N4004	28357-028

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Car var. $2.5k\Omega 20\% 0.21W$

For symbols and abbreviations see introduction to this chapter

 $\mathbf{D4}$

25541-322

Diode 1N4004

28357-028

R12

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Circ		Circuit		iit	
refere D5	ence Description Diode Zener Z5B8.2	M.I. code	referei	, · · ·	M.I. code
D7	Diode 1N4004	28371-673	R4	Met film 2kΩ 2% ¼W	24773-280
		28357-028	R5	Met film $2k\Omega \ 2\% \ \frac{1}{4}W$	24773-280
R1	Met film 2.2kΩ 2% ¼W		R6	Met film $10k\Omega \ 2\% \ \frac{1}{4}W$	24773-297
R2	Met film 2.2k Ω 2% $\frac{1}{4}W$	24773-281	R7	Met film 220k Ω 2% $\frac{1}{4}$ W	24773-329
R3	Met film $12k\Omega \ 2\% \frac{1}{4}W$	24773-281	R8	Met film $10k\Omega 2\% \frac{1}{4}W$	24773-297
R4	Preset 2.2k Ω 20% 0.21W	24773-299	R9	Met film 150k Ω 2% $\frac{1}{4}$ W	24773-325
R5		25711-547	R10	Carb 1MΩ 5% ¼W	24311-945
R6	Met film 7.5k Ω 2% $\frac{1}{4}W$	24773-294	R11	Carb 1MΩ 5% ¼W	24311 - 945
R7	Met film 2.2k Ω 2% $\frac{1}{4}W$	24773-281	TR1 Transistor BFW11		28459-012
	Met film $1.5k\Omega 2\% \frac{1}{4}W$	24773-277			20455-012
R8.	Met film $24k\Omega \ 2\% \ \frac{1}{4}W$	24773-306	Unit (A22) coarse RF output (att		
н9 †		24773-283		rdering, prefix circuit reference with A2	anuator) A22
R10 †		24773-293	САМ	10dB	
R11	Met film 2.2k Ω 2% $\frac{1}{4}$ W	24773-281	CAM	20/1dB	31359-022
R12	Met film $100\Omega \ 2\% \ \frac{1}{4}W$	24773-249	CAM	20/2dB	31359-021
R13	Met film 13kΩ 2% ¼W	24773-300	CAM	30/1dB	31359-019
R14 †	Met film $3k\Omega \ 2\% \ \frac{1}{4}W$	24773-284	CAM		31359-018
R15	Met film 100Ω 2% ¼W	24773-249	CAM	30/2dB	31359-020
R16	Met film 3.3k Ω 2% $\frac{1}{4}W$	24773-321	R1		
R18	WW 1.5kΩ 5% 1.5W	25123-084		Met film $790\Omega = 1\% \frac{1}{4}W$	24762-646
			R2	Met film 790Ω 1% ¼ W	24762-646
TR2	Transistor BFY51	28455-827	R3	Met film $247 \Omega = 1\% \frac{1}{4} W$	24762-631
TR3	Transistor BC108	28452-787	R4	Met film $247\Omega = 1\% \frac{1}{4}W$	24762-631
34	Transistor MM4001	28432-787	$\mathbf{R5}$	Met film 71.2 Ω 1% $\frac{1}{4}W$	24762-572
TR5	Transistor BCY71		R6	Met film 53.3 Ω 1% $\frac{1}{4}W$	24762 - 557
TR6 +		28435-235	R7	Met film 53.3 Ω 1% $\frac{1}{4}W$	24762-557
I	(complete with matched	46883-258	R8	Met film 53.3 Ω 1% $\frac{1}{4}W$	24762-557
	resistors R9 and R10)		R9	Met film 53.3 Ω 1% $\frac{1}{4}$ W	24762~557
TR7	Transistor BC239C	28452-771	R10	Met film 61.1 Ω 1% $\frac{1}{4}$ W	24762-571
TR8	Transistor BC239C	28452-771	R11	Met film 61.1Ω 1% ¼W	24762-571
			R12	Met film 61.1Ω $1\%\frac{1}{4}W$	24762-571
Unit (A			R13	Met film 61.1 Ω 1 $\%$ $\frac{1}{4}W$	24762-571
	ering, prefix circuit reference with A21		R14	Met film 96.3 Ω 1% $\frac{1}{4}$ W	24762-582
C1	Plas 0.47µF 10% 63V	26582-410	R15	Met film 96.3 Ω 1% $\frac{1}{2}W$	24762-582
C2	Elec 22µF 20% 15V	26486-583		~ *	
C3	Plas 0.1µF 10% 100V	26582-211	SA	Microswitch UHF	23483-131
ri †	Met film $6.8k\Omega 2\% \frac{1}{4}W$	24773-293	SB	Microswitch UHF	
R2	Met film 10k Ω 2% $\frac{1}{4}$ W	24773-297	SC	Microswitch UHF	23483-131
	For tumbo		-	OF ALCH OTT	23483-131

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For symbols and abbreviations see introduction to this chapter

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	ircuit erence Description	M.I. code	Circuit reference	Description	
SD SD	Microswitch UHF			Description	M.I. code
SE	Microswitch UHF	23483-131			
SF	Microswitch UHF	23483 - 131 23483 - 131	i uncu		41145 - 407
SG	Microswitch UHF	23483-131	Carrie	er range	41145-407
SH	Microswitch UHF	23483-131	1 1110 11	ine	41141-708
SJ	Microswitch UHF	23483-131			41149-014
SK	Microswitch UHF	23483-131	Extra fine tune		41149-016
		20403-131	Set mo		41149 - 002
	(A23) FM predistortion—ter	mperature	Deviati		41149-002
	pensation		AM dep		41149-002
¥¥ hen	ordering, prefix circuit reference with A	23		r.f. output	41145-407
C1	Elec 4.7µF 20% 35V	26486-219		f. output	41149-002
C2	Elec 4.7µF 20% 35V	26486-219	Cursor (pointer a	assembly)	41179-898
		20100-219	Drive cord		16410 - 604
D1	Diode 1N4148	28336-676	Drive cord spring	3	31119-017
D2	Diode 1N4148	28336-676	Drive cord pin		33900-110
		20030-010	Handle (strap)		37587-925
R1	Carb 680kΩ 5% 1/8W	24311-941	Handle (attachme		37588-110
R_2	Met film 10k Ω 2 $\%$ $\frac{1}{4}$ W		Handle (end caps) Scale 10-14.3 MHz		22315-573
R3	Met film $100k\Omega 2\% \frac{1}{4}W$	24773-297 24773-321			31759-004
R4	Met film $56k\Omega \ 2\% \ \frac{1}{4}W$	24773-321	14.3-20.5 N		31759-005
R5	Met ox $820k\Omega \ 2\% \ \frac{1}{2}W$	24573-143 .	20.5-29.9 N	IHz	31759-006
$\mathbf{R6}$	Met ox $330 k\Omega \ 2\% \ \frac{1}{2}W$	24573-133	29.4-42 MH:	z	31759-007
7	Resistor assembly 270Ω	44359-003	42-60 MHz		- 31759-008
R8	Met film $220\Omega \ 2\% \ \frac{1}{4}W$	24773-257	60-86 MHz		31759-009
R9	Met film 560 Ω 2% $\frac{1}{4}$ W		86-124 MHz		31759-010
R10	Thermistor 1.5k Ω (CZ3)	24773 - 321 25683 - 644	124-177 MHz		31759-011
R11	Met film $820\Omega \ 2\% \ \frac{1}{4}W$	24773-271	177-254 MHz		31759-012
		24113-211	254-363 MHz		31759-014
Misseller			363-520 MHz		31759-021
Miscellaneous			Scale window		37490-020
Cap and chain (TNC)		A A A A A	Spiral bevel gears (carrier range switch) / pair		∫ 31318-406
		23444-898			31318-407

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Index to Units

Unit number	M.I. code	Description	Parts list ‡age	Circuit diagram Fig.
A0	-	Chassis and panel components	42	All
A1	44827-137	Oscillator range 1	44	7.1
A2	44827-138	Oscillator range 2	44	7.1
A3	44827-139	Oscillator range 3	44	7.1
A4	44827-140	Oscillator range 4	44	7.1
A5	44827-141	Oscillator range 5	45	7.1
A6	44827-142	Oscillator range 6	45	7.1
A7	44827-143	Oscillator range 7	45	7.1
A8	44827-144	Oscillator range 8	46	7.1
A9	44827-145	Oscillator range 9	46	7.1
A10	44827-146	Oscillator range 10	46	7.1
A11	44827-147	Oscillator range 11	46	7.1
A12	44827-148	Mother board	47	7.1
A16	44827-152	Pre-amplifier, modulator, counter amplifier, wide band amplifier	47	7.2
A17	44827-153	ALC and AM driver circuits	49	7.2
A18	44827-154	Monitors	50	7.2
A19	44827-155	Modulation oscillator and FM drive	51	7.3
A20	44827-105	Power supply	51	7.4
A21	44827-222	Fine tuning circuits	52	7.3
A22	44429-004	Coarse RF output (attenuator)	52	7.4
A23	44827-366	Predistortion circuits	53	7.3



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Circuit diagrams

Circuit notes

1. COMPONENT VALUES

Resistors: No suffix = ohms, k = kilohms, M = megohms. Capacitors : No suffix = microfarads, p = picofarads. Inductors : No suffix = henrys, m = millihenrys, μ = microhenrys. + value selected during test, nominal value shown.

2. VOLTAGES

Shown in italics adjacent to the point to which the measurement refers. See Sect 5.2 for conditions.

These symbols indicate the method used for voltage measurements at various points.

- Use 4.7kΩ resistor in series with meter.
 Position resistor at test point.
- Use 100 V meter range
- see Table 5.2.

3. SYMBOLS

Symbols are based on the provisions of B.S. 3939

+ arrow indicates clockwise rotation of knob.

[RANGE] etc., external front or rear panel marking.

- -Ó- tag on printed board,
- O preset control.
- (A12) unit identification number.
- 4. CIRCUIT REFERENCES

These are, in general, given in abbreviated form. See also introduction to Chap. 6.

5. SWITCHES

Rotary switches are drawn schematically. Letters or figures indicate control knob settings. 1F = 1st section (front panel), front. 1B = 1st section, back 2F = 2nd section, front. etc. These symbols are used to identify branches of the power supply circuitry but have no particular physical reality on the printed boards.



point marked with this symbol is connected to and receives power from point marked with this symbol.



Fig. 7.3 Tuning, modulation and predistortion

2015 (2f)



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2015 (2e)





